

# wireless world



FEBRUARY 1979 40p

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**Audio preamplifier  
Ionosphere and h.f.**



# high quality

## Multicore solder helps Europe's OTS-2 satellite orbit at 23,000 miles

OTS-2, a forerunner of a European communications satellite, was launched from Cape Canaveral on May 11th 1978.

Now orbiting the equator at a height of 23,000 miles, OTS-2 was built for the European Space Agency by the MESH consortium, led by British Aerospace Dynamics Group, Stevenage.

In building OTS-2, complete precision was called for. Which is why Ersin Multicore solders were used for many of the soldering operations.

Quite simply, Multicore solders have *that* kind of reputation. For quality, toughness and reliability. That's why so many electronics manufacturers won't use anything else.

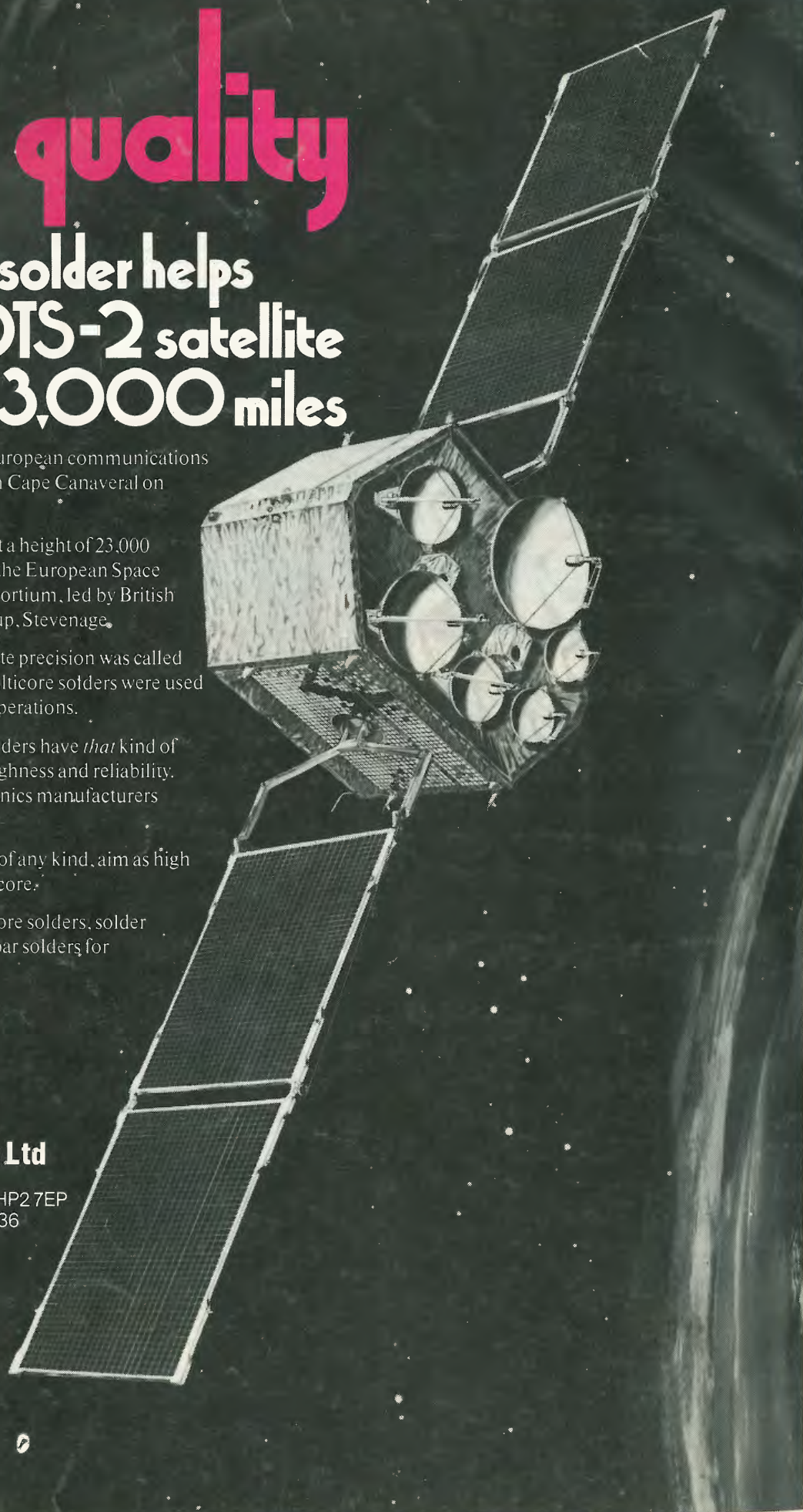
Next time you need solder of any kind, aim as high as OTS-2... and use Multicore.

Full details of Ersin Multicore solders, solder chemicals and high purity-bar solders for automatic soldering from



### Multicore Solders Ltd

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Hemel Hempstead Herts HP2 7EP  
Tel: Hemel Hempstead 3636  
Telex: 82363



WW-004 FOR FURTHER DETAILS

# wireless world

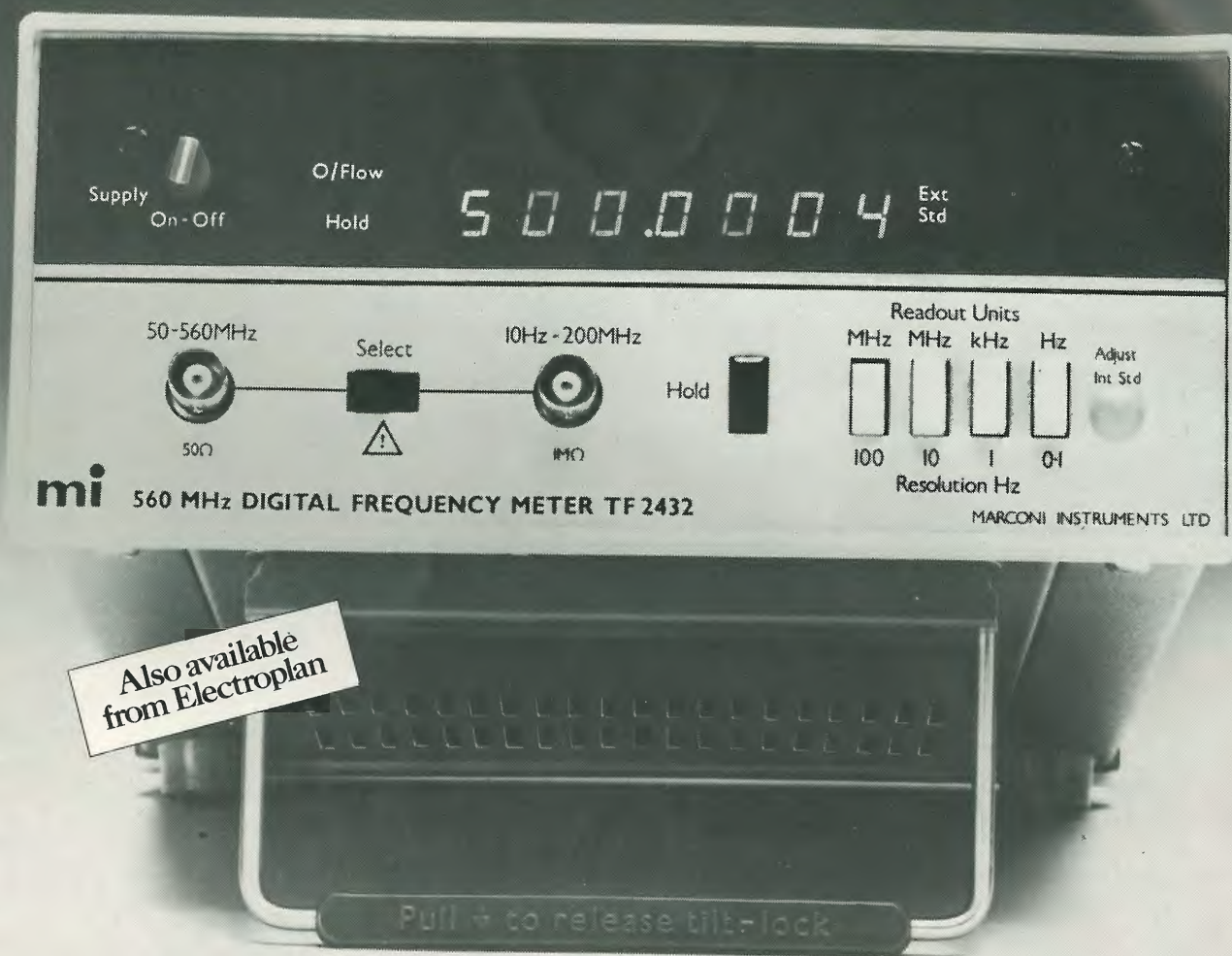
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WIRELESS WORLD FEBRUARY 1979 VOL 85 NO 1518

## Audio preamplifier Ionosphere and h.f.

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Greece Dr. 47.00  
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WIRELESS WORLD, FEBRUARY 1979



Front cover shows flexible printed circuit material made using Du Pont KAPTON by M. B. Metals of Portslade. Photo by Roles & Parker.

## IN OUR NEXT ISSUE

**Novatexts.** Start of a series in which tutorial information on electronics is presented in a new way. Devised by Peter Williams of the Circards team.

**Logic analyser.** Diagnostic instrument for construction monitors signal points and displays real time data on their logic states. On-line, it will detect faults.

**H.f. synthesizer** for amateur use covers 1.5 to 28.5MHz in 500 kHz bands. Construction design includes a digital p.l.l., oscillator and mixer.

Current issue price—40p, back issue (if available) 50p, at Retail and Trade Counter, Paris Garden, London SE1. Available on microfilm: please contact editor.

By post, current issue 55p, back issues (if available) 50p, order and payments to Room CP34, Dorset House, London SE1 9LU.

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**Telephones:** Editorial 01-261 8620. Advertising 01-261 8339.

**Telegrams/Telex:** Wlworld Bispres 25137 BISPRES G. Cables: Ethaworld, London SE1.

**Subscription rates:** 1 year: £7.00 UK and \$23.40 overseas (\$24 USA and Canada).

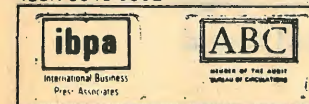
**Student rate:** 1 year, £3.50 UK and £4.50 overseas (\$11.70 USA and Canada).

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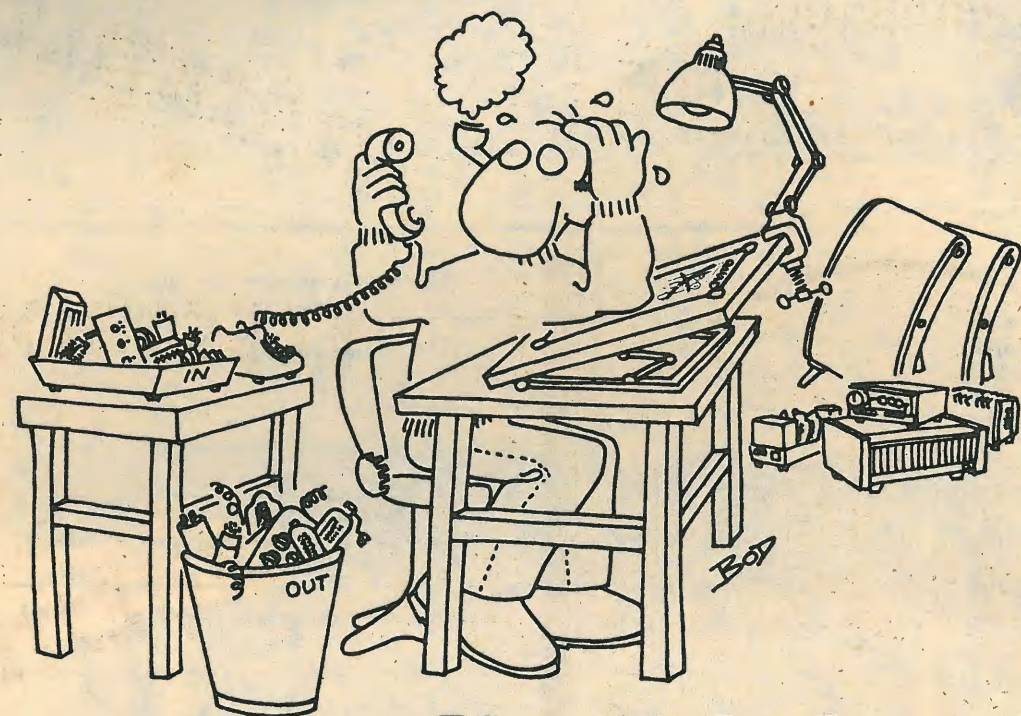
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## If walls had ears?

"We can't possibly use an output condenser, it must cut the bass mustn't it? And what about the damping?"

"And no output transformer, what with all that hysteresis and iron distortion."

"Pentodes? Tetrodes?"

"No, No, nothing but triodes will do."

"Triodes then, but wait, we can't have all that accumulated Miller effect."

"Transistors then?"

"Oh no, this year's crop are all hard and brittle."

"And that see-saw phase splitter, it's asymmetrical; if we fed a square wave..."

"But what have square waves to do with programme?"

"Shut up, that's irrelevant."

"Class B? But doesn't that always produce crossover distortion?"

"Ah! Feedback will cure all;"

"No, No, we've read that too much feedback causes TID or something."

Of course, these things have little or nothing to do with good or bad amplifier design, and are not at all what you might overhear in our laboratory..... zzzzzzzzzzz

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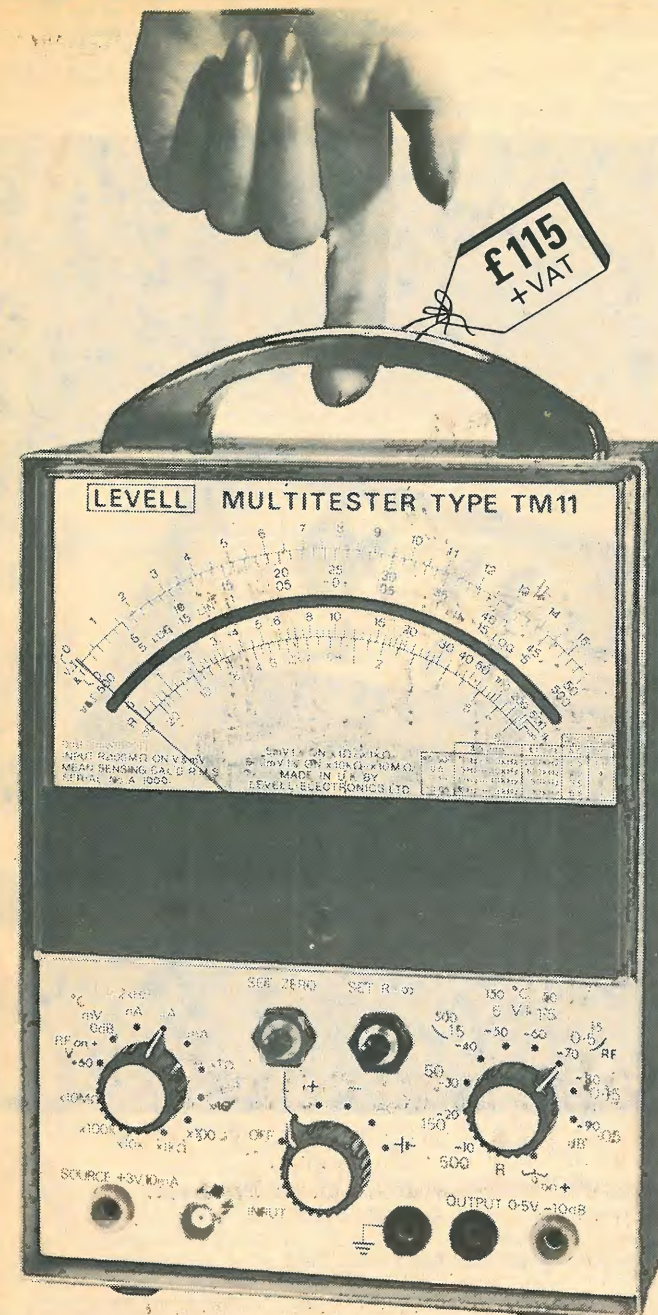
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- RESISTANCE : 0.2 $\Omega$ /10G $\Omega$  in 7 ranges, polarity reversible. Low test voltage for solid state circuits.
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The instrument operates from a 9 volt battery, life 1000 hrs., or, AC mains when optional Power Supply Unit is fitted. Size is 240mm x 150mm x 80mm. Weight is 1.75 kg. Meter scale length is 140mm. Leather case is available at £14 + VAT.

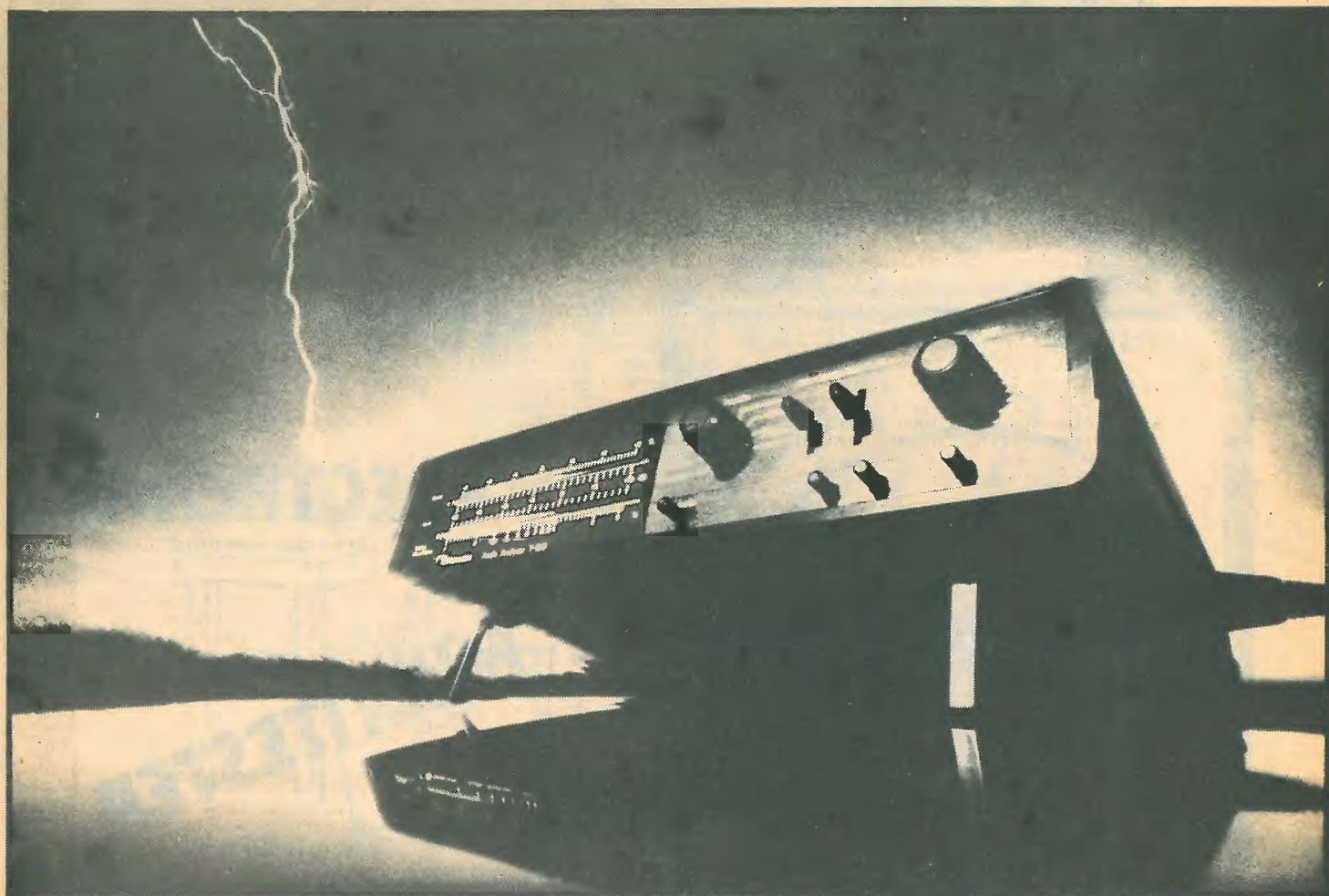
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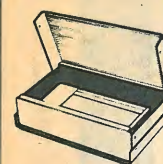
## YOUR COMPLETE RANGE OF ELECTRONIC HARDWARE...

### BIMENCLOSURES

#### ALL METAL BIMCASES

Red, Grey or Orange 14swg Aluminium removable top and bottom covers. 18 swg black mild steel chassis with fixing support brackets.

BIM 3000  
(250x167.5x68.5mm)  
£14.58



#### MINI DESK BIMCONSOLES

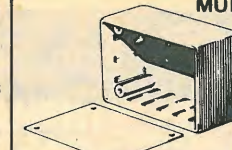
Orange, Blue, Black or Grey ABS body incorporates 1.8mm pcb guides, stand-off bosses in base with 4 BIMFEET supplied. 1mm Grey Aluminium panel sits recessed with fixing screws into integral brass bushes.



BIM 1005 (161 x 96 x 58mm) £2.18  
BIM 1006 (215 x 130 x 75mm) £3.05

#### MULTI PURPOSE BIMBOXES

Orange, Blue, Black or Grey ABS with 1mm Grey Aluminium recessed front cover held by screws into integral brass bushes. 1.8mm pcb guides incorporated and 4 BIMFEET supplied.



BIM 4003 (85x56x28.5mm) £1.18  
BIM 4004 (111x71x41.5mm) £1.62  
BIM 4005 (161x96x52.5mm) £2.19

#### ALL METAL BIMCONSOLES

All aluminium, 2 piece desk consoles with either 15° or 30° sloping fronts, sit on 4 self-adhesive non-slip rubber feet. Ventilation slots in base and rear panel for excellent cooling. See latest catalogue for new styles and sizes.



##### 15° Sloping Panel

BIM7151 (102x140x51[28] mm) £10.67  
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BIM7153 (165x216x51[28] mm) £12.61  
BIM7154 (165x211x76[33] mm) £13.82  
BIM7155 (254x211x76[33] mm) £15.36  
BIM7156 (254x287x76[33] mm) £16.67  
BIM7157 (356x211x76[33] mm) £17.58  
BIM7158 (356x287x76[33] mm) £18.55

##### 30° Sloping Panel

BIM7301 (102x140x76[28] mm) £10.67  
BIM7302 (165x140x76[28] mm) £11.44  
BIM7303 (165x183x102[28] mm) £12.61  
BIM7304 (254x140x76[28] mm) £13.82  
BIM7305 (254x183x102[28] mm) £15.36  
BIM7306 (254x259x102[28] mm) £16.67  
BIM7307 (356x183x102[28] mm) £17.58  
BIM7308 (356x259x102[28] mm) £18.55

#### ABS & DIECAST BIMBOXES

6 sizes in ABS or Diecast Aluminium. ABS moulded in Orange, Blue, Black or Grey. Diecast Aluminium in Grey Hammetone or Natural. All boxes incorporate 1.8mm pcb guides, stand-off supports in base and have close fitting flanged lids held by screws into integral brass bushes (ABS) or tapped holes (Diecast).

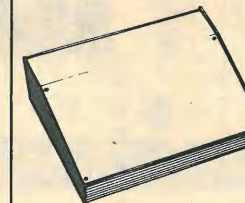


	ABS	Diecast	Hammetone	Natural
(50x60x31mm)	N/A	BIM5001/11	TBA	£1.02
(100x50x25mm)	BIM2002/12	£0.96	BIM5002/12	£1.46
(112x62x31mm)	BIM2003/13	£1.13	BIM5003/13	£1.78
(120x65x40mm)	BIM2004/14	£1.35	BIM5004/14	£2.24
(150x80x50mm)	BIM2005/15	£1.52	BIM5005/15	£2.84
(190x110x60mm)	BIM2006/16	£2.37	BIM5006/16	£3.94

Also available in Grey Polystyrene with no slots and self-tapping screws  
BIM 2007/17 (112x61x31mm) £1.00

#### LOW PROFILE BIMCONSOLES

Orange, Blue, Black or Grey ABS body has ventilation slots as well as 1.8mm pcb guides and stand-off bosses in base. Double angle recessed front panel with 4 fixing screws into integral brass bushes. 4 BIMFEET supplied.



BIM 6005 (143 x 105 x 55.5 [31.5] mm) £2.37  
BIM 6006 (143 x 170 x 55.5 [31.5] mm) £3.08  
BIM 6007 (214 x 170 x 82.0 [31.5] mm) £4.12

#### EUROCARD BIMCONSOLES

Orange, Blue, Black or Grey ABS body accepts full or 1/2 size Eurocards, with bosses in the base for direct fixing. 1.8mm wide pcb guides incorporated and 4 BIMFEET supplied. 1mm Grey aluminium lid sits flush with body top and held by 4 screws into integral brass bushes.



BIM 8005 (169x127x70[45] mm) £4.12  
BIM 8007 (243x187x103[66] mm) £6.10

### BIMTOOLS + BIMACCESSORIES

#### MAINS BIMDRILLS

Small, powerful 240V hand drill complete with 2 metres of cable and 2 pin DIN plug. Accepts all tools with 1mm, 2mm or .125" dia. shanks. Drills brass, steel, aluminium and pcb's. Under 250g, off load speed 7500 rpm. Orange ABS, high impact, fully insulated body with integral on/off switch £10.53

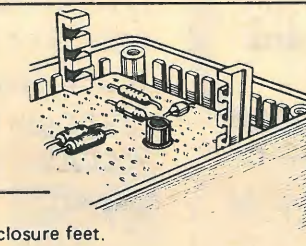


Mains Accessory Kit 1 includes 1mm, 2mm, .125" twist drills, 5 burrs and 2.4mm collet £2.48

Mains Kit 2 includes Mains BIMDRILL as above, 20 assorted drills, mops, burrs, grinding wheels and mounted points, 1mm, 2mm, 2.4mm and .125" collets. Complete in transparent case measuring 230x130x58mm £22.14

#### BIMDAPTORS

Allows pcb's to be flat mounted sandwich fashion in BIMBOXES, BIMCONSOLES, and all other enclosures having 1.5mm wide vertical guide slots. One plastic BIMDAPTOR on each corner of pcb(s) enables assembly to be simply slid into place. 54mm long, 10 slots on 5mm spacing and can be simply snapped off to length. £1.08 per pack of 25.



#### BIMFEET

11mm dia. 3mm high, grey rubber self-adhesive enclosure feet. £0.77 per pack of 24



#### 12 VOLT BIMDRILLS

2 small, powerful drills easily hand held or used with lathe/stand adaptor. Integral on/off switch and 1 metre cable.

Mini BIMDRILL with 3 collets up to 2.4mm dia. £ 8.10  
Major BIMDRILL with 4 collets up to 3mm dia. £13.60



Accessory Kits 1 have appropriate drills and collets as above plus 20 assorted tools. Mini Kit 1 - £15.12, Major Kit 1 - £19.44. Accessory Kits 2 have appropriate drills, collets plus 40 tools and mains-12V dc adaptor. Mini Kit 2 - £34.02, Major Kit 2 - £39.42. Accessory Kits 3 as appropriate Kits 2 plus stand/lathe unit. Mini Kit 3 - £45.36, Major Kit 3 - £50.76.

#### BIMPUMPS

2 all metal desoldering tools provide high suction power and have easily replaceable screw in Teflon tips. Primed and released by thumb operation with in-built safety guard and anti-recoil system.



BIMPUMP Major (180mm long) £7.99  
BIMPUMP Minor (150mm long) £6.80

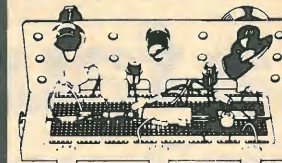
#### BIMIRONS

Type 30 General Purpose 27 watt iron with long life, rapid change element, screw on tip, stainless steel shaft and clip on hook. Styled handle with neon. £4.05



Type M3 Precision 17 watt iron, quick change tip, long life element, styled handle with clip on hook. £4.43

### BIMBOARDS



DIL COMPATIBLE BIMBOARDS

Accept all sizes (4-50 pin) of DIL IC packages as well as resistors, diodes, capacitors and LEDs. Integral Bus Strips up each side for power lines and Component Support Bracket for holding lamps, switches and fuses etc. Available as single or multiple

units, the latter mounted on 1.5mm thick black aluminium back plate which stand on non slip rubber feet and have 4 screw terminals for incoming power.

BIMBOARD 1 has 550 sockets, multiple units utilising 2, 3 and 4 BIMBOARDS incorporate 1100, 1650 and 2200 sockets, all on 2.5mm (0.1") matrix.

BIMBOARD 1 £ 8.83  
BIMBOARD 2 £21.01  
BIMBOARD 3 £29.84  
BIMBOARD 4 £38.79

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1, 2, or 3 BIMBOARDS mounted on BIM 6007 BIMCONSOLE with Integral Power Supply (±5 to ±15Vdc @ 100mA and fixed +5Vdc @ 1A) All O/P's fully isolated. Short circuit and fast fold back protection. Power rails brought out to cable clamps that accept stripped wire or 4mm plug.

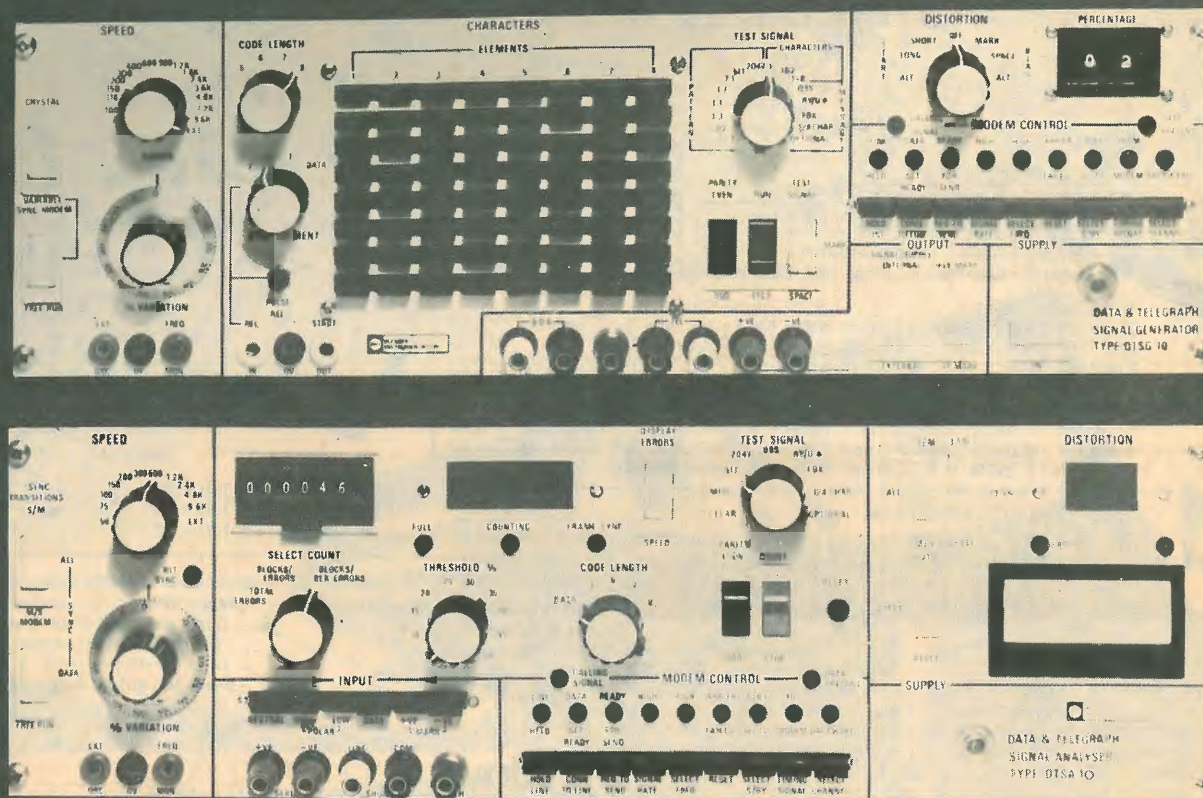
DESIGNER 1 £55.62  
DESIGNER 2 £61.02  
DESIGNER 3 £66.42

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## Data faults sorted-fast!

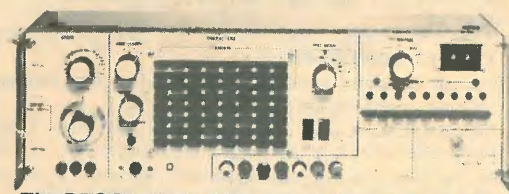
### using the Plessey Data & Telegraph Test Set DTTS10

This compact and highly versatile equipment enables you to locate and quickly isolate faults and problems by substituting the DTTS10 for your terminal (circuit looped) or using the units at either end of your channel.

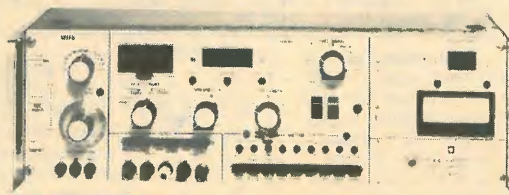
The generator DTSG10 provides a wide range of CCITT recommended test signals and patterns and also includes a push button matrix, useful for special pattern or polling addresses.

The associated signal analyser

DTSA10 offers facilities for measuring speed or bit, block/bit and block/block errors and also measures peak and bias distortion. Both units include modem and control facilities, valuable when testing end-to-end.



The DTSG10 Data & Telegraph Signal Generator



The DTSA10 Data & Telegraph Signal Analyser

**PLESSEY**  
CONTROLS

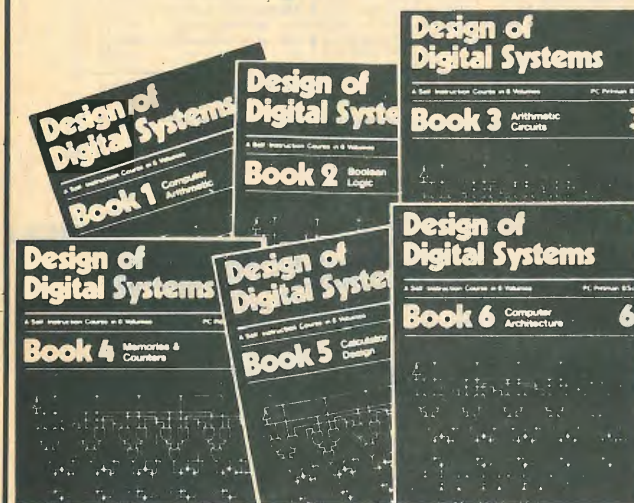
Plessey Controls Limited, Sopers Lane, Poole, Dorset, United Kingdom, BH17 7ER  
Telephone: Poole (020 13) 5161 Telex: 41272

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## Understanding Digital Electronics

### New teach-yourself courses



In the years ahead the products of digital electronics technology will play an important part in your life. Calculators and digital watches are already commonplace. Tomorrow a digital display could show your vehicle speed and petrol consumption; you could be calling people by entering their name into a telephone which would automatically look up their number and dial it for you.

These courses were written by experts in electronics and learning systems so that you could teach yourself the theory and application of digital logic. Learning by self-instruction has the advantages of being faster and more thorough than classroom learning. You work at your own pace and must respond by answering questions on each new piece of information before proceeding.

After completing these courses you will have broadened your career prospects and increased your fundamental understanding of the rapidly changing technological world around you.

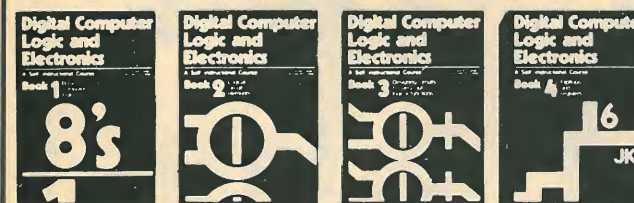
The six volumes of **Design of Digital Systems** cost only: **£8.10** + 90p post & packing  
And the four volumes of **Digital Computer Logic and Electronics** cost only: **£4.60** + £1 post & packing  
But if you buy both courses, the total cost is only: **£12.00** + £1 post & packing

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Design of Digital Systems is written for the engineer seeking to learn more about digital electronics. Its six volumes — each A4 size — are packed with information, diagrams and questions designed to lead you step-by-step through number systems and Boolean algebra to memories, counters and simple arithmetic circuits, and finally to a complete understanding of the design and operation of calculators and computers.

#### The contents of Design of Digital Systems include:

- Book 1** Octal, hexadecimal and binary number systems; conversion between number systems; representation of negative numbers; complementary systems; binary multiplication and division.
- Book 2** OR and AND functions; logic gates. NOT, exclusive OR, NAND, NOR and exclusive-NOR functions; multiple input gates; truth tables; De Morgan's Laws; canonical forms; logic conventions; Karnaugh mapping; three-state and wired logic.
- Book 3** Half adders and full adders; subtractors; serial and parallel adders; processors and arithmetic logic units (ALUs); multiplication and division systems.
- Book 4** Flip flops; shift registers; asynchronous and synchronous counters; ring, Johnson and exclusive-OR feedback counters; random access memories (RAMs) and read only memories (ROMs).
- Book 5** Structure of calculators; keyboard encoding; decoding display data; register systems; control unit; program ROM; address decoding; instruction sets; instruction decoding; control program structure.
- Book 6** Central processing unit (CPU); memory organisation; character representation; program storage; address modes; input/output systems; program interrupts; interrupt priorities; programming; assemblers; computers; executive programs; operating systems and time sharing.



**Digital Computer Logic and Electronics** is designed for the beginner. No mathematical knowledge other than simple arithmetic is assumed, though the student should have an aptitude for logical thought. It consists of four volumes — each A4 size — and serves as an introduction to the subject of digital electronics. Everyone can learn from it — designer, executive, scientist, student, engineer.

Contents include: Binary, octal and decimal number systems; conversion between number systems; AND, OR, NOR and NAND gates and inverters; Boolean algebra and truth tables; De Morgan's Laws; design of logic circuits using NOR gates; R-S and J-K flip flops; binary counters, shift registers and half adders.

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WW35



# ambit<sup>®</sup> international

Production of the new catalogue has been held up for a few weeks - since we have just been appointed as distributors for two of the most exciting ranges of radio components products yet - The Micrometals range of iron dust torroids cores and formers, and the OKI range of VLSI for digital frequency displays for receivers. We apologize for any inconvenience, but these two ranges are really worth the wait, and include some products you will find hard to believe, like the MSM5523 IC, an IC with less than ten external components that gives AM frequency readout to 1kHz from LW to 39.999MHz, FM frequency readout in 100kHz steps - (all usual IF offsets programmable by diodes), a 24 hour format clock with 12 hour display, independent on and off timers, time signals on the hours, stopwatch facility and a sleep timer. This costs £14 with its timebase crystal, and makes all that has gone before an expensive and time wasting exercise. Rather like the way the Intersil ICM7216 has revolutionized the instrument counter market. (See the OSTs ad.) And those of you familiar with Amidon and IG dust torroids, favoured in many new RF designs, will be pleased to know Ambit will be stocking a broad range of the Micrometals types for applications from EMI filters to RF PA stages.

OKI frequency counter ICs: details in cat2  
MSM5523 for CA LEDs with RHPD such as FND507 £14 inc xtal  
MSM5525 for 3 1/2 digit LCD AM/FM with direct segment drive, no clock or timers £11 inc xtal  
Other types for fluorescent displays etc OA

Other new semiconductor additions:  
KB4437 pilot cancel mpx decoder 4.35  
KB4438 muting stereo preamp 2.22  
HA1370 supercedes TDA2020 2.99  
TDA1090 HiFi AM/FM 3.35  
TDA1220 low cost AM/FM 1.45

PRICES DOWN ON VMOS: as expected, this new technology in power transistors is getting cheaper. 120v comp pairs /100W for £10.00

Price reduction on CA3189E - now £2.20  
New varicaps: to add to the biggest range...  
KV1211 2-9v bias to tune MW, like the KV1210, but a double diode £1.75

New pilot tone filters from TOKO...  
208BLR series, individual per channel with a 26/38kHz version for pilot cancel decoder applications. Flat to 15kHz £0.90

New crystal filter for amateur NBFM...  
TOYO 10M4B1 with over 90dB adjacent ch rejection for 2m NBFM. 10.7MHz £14

New ceramic IF filters for 455kHz...  
CFM455H 6kHz/6dB, 15kHz max /60dB - ideal for MC3357 etc. £10

TERMS etc: CWO please, VAT on Ambit items is generally 12%, except where marked (\*). Catalogue part 1:45p, part 2:50p all inclusive. Postage 25p per order, carriage on tuner kits £3. Phone Brentwood (0277) 216029/227050 9am-7pm. Callers welcome inc. Saturdays.

## At last, DIY Hi Fi which looks as if it isn't.

That's not to say it doesn't look like HiFi - just that it doesn't look like the usual sort of thing you have come to associate with DIY HiFi. The Mk3 outstrips and outperforms all British made HiFi tuners, and most imported ones too. Certainly at the price, there isn't one near it. But more than that, it looks superb - A small pic here would be an insult, so send an SAE for details on the kit that looks as if isn't. It's something else.....

- \* Exceptionally high performance - exceptionally straightforward assembly
- \* Baseboard and plug-in construction. Future circuit developments will readily plug in, to keep the MkIII at the forefront of technical achievement
- \* Various options and module line-ups possible to enable an installment approach to the system

and now previewing the matching 60W/channel VMOS amplifier:

- \* Matching both the style and design concepts of the MkIII HiFi FM tuner
- \* Hitachi VMOS power fets - characterized especially for HiFi applications
- \* Power output readily multiplied by the addition of further MOSFETs
- \* VU meters on the preamp - not simply dancing according to vol level
- \* Backed with the usual Ambit expertise and technical capacity in audio

## The PW Dorchester-LW,MW,SW,& FM stereo tuner



In much the same way as we have swept away the 'old technology' in frequency/timer counters - with the OKI and Intersil single IC counters, we now offer a single IC "All Band" radio tuner. Don't confuse this one chip radio with things like the ZN414 - for this is a genuine superhet receiver with a mechanical AM IF filter, and ceramic IF filters for FM.

The AM section employs a balanced input mixer section, covering all broadcast bands - plus a BFO and MOSFET product detector for SSB/CW - though at this price, the tuner is not intended as a "communications receiver" - although we know of many lesser designs that make that claim. The AM sensitivity is nevertheless better than 5uV, and FM sensitivity is 1.2uV for 30dB S/N. As a multiband broadcast superhet receiver, it is a unique constructor project that fulfills the requests we very frequently get for a general coverage circuit that isn't over complicated. The set has CA3089E FM performance, with mute etc., and a PLL stereo decoder with full pilot tone filtering.

The tuner board - with "on board" PCB mounted switching, all components etc : £33.00  
The case/cabinet with PSU, meter and mechanics etc £25.00  
An SAE for full details please. See the feature article in Practical Wireless (Dec/Jan)

**2 Gresham Road, Brentwood, Essex.**

WW-045 FOR FURTHER DETAILS

## OSTS

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## CD4000 CMOS

4000	17p	4059	563p	4522	149p
4001	17p	4060	115p	4527	157p
4002	17p	4063	109p	4528	102p
4006	109p	4066	53p	4529	141p
4007	18p	4067	400p	4530	90p
4008	80p	4068	25p	4531	141p
4009	58p	4069	20p	4532	125p
4010	58p	4070	20p	4534	614p
4011	17p	4071	20p	4536	380p
4012	17p	4072	20p	4538	150p
4013	55p	4073	20p	4539	110p
4016	52p	4075	20p	4541	141p
4017	80p	4076	90p	4543	174p
4018	80p	4077	20p	4549	399p
4019	80p	4078	20p	4553	440p
4020	93p	4081	20p	4554	153p
4021	82p	4082	20p	4556	77p
4022	90p	4085	82p	4557	386p
4023	17p	4086	82p	4558	117p
4024	76p	4089	150p	4559	388p
4025	17p	4093	50p	4560	218p
4026	180p	4094	190p	4561	65p
4027	55p	4096	105p	4562	530p
4028	72p	4097	372p	4566	159p
4029	100p	4098	110p	4568	281p
4030	58p	4099	122p	4569	303p
4031	250p	4160	90p	4572	25p
4032	100p	4161	90p	4580	600p
4033	145p	4162	90p	4581	319p
4034	200p	4163	90p	4582	164p
4035	120p	4174	104p	4583	84p
4036	250p	4175	95p	4584	63p
4037	100p	4194	95p	4585	100p
4038	105p	4501	23p		
4039	250p	4502	91p		
4040	83p	4503	69p		
4041	90p	4506	51p		
4042	85p	4507	55p		
4043	85p	4508	248p		
4044	80p	4510	99p		
4045	150p	4511	149p		
4046	130p	4512	98p		
4047	99p	4513	206p		
4048	60p	4514	260p		
4049	55p	4515	300p		
4050	55p	4516	125p		
4051	65p	4517	362p		
4052	65p	4518	103p		
4053	65p	4519	57p		
4054	120p	4520	109p		
4055	135p	4521	236p		

## micromarket

6800 series	8216	1.35
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6810P	8255	5.40
6852		3.65

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8212	2112	£3.40
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	4027	£5.78

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6810P £4	<b>MEMORIES</b>		21
6852 <b>3.65</b>	2102	£1.70	T1, In
<b>8080 series</b>	2112	£3.40	Harris
8080	2513	£7.54	
8212 <b>2.30</b>	4027	£5.78	

## Voltage Regs

### NEW LOW PRICES

7800 series UC TO220 package 1A	all 95p
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78LCP series TO92 100mA	all 35p
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78MGT2C %amp adjustable volts	175p
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723C precision controller	65p

### MAINS FILTERS FOR NOISE/RFI etc

1 amp in IEC connector	£4.83
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NE550A 73p	

## LINEARS

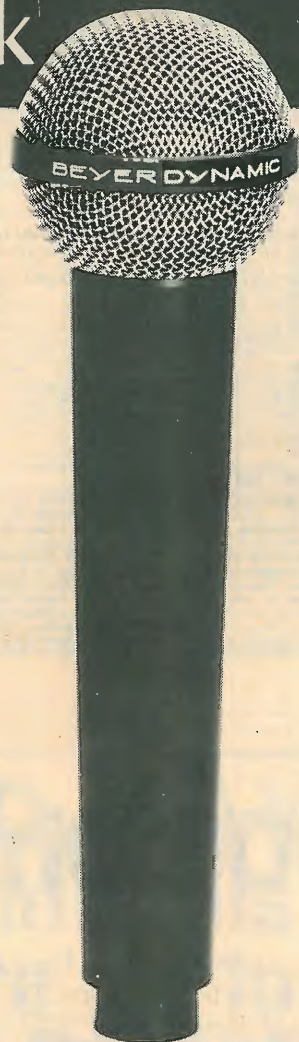
Op amps	LM301AH	67p	LM301AN	30p	LM308H	121p	LM308N	70p	LM318H	279p	LM318N	224p
	CA3130E	84p	LM339N	66p	709HC	105	709PC	36p	710HC	105	710PC	59p
	CA3130T	90p	LM348N	186p	710HC	105	710PC	36p	710HC	105	710PC	59p
	CA3140E	35p	LM3900N	60p	710HC	105	710PC	36p	710HC	105	710PC	59p
	CA3140T	72p	709HC	105	64p	710HC	105	65p	710HC	105	710PC	59p
	CA3160E	90p	709PC	36p		710HC	105	65p	710HC	105	710PC	59p
	CA3160T	99p	710HC	105	65p	710HC	105	65p	710HC	105	710PC	59p
	Op amps		710PC	59p		710PC	59p		710PC	59p		
			723CN	65p		723CN	65p		723CN	65p		
			741CH	105	66p	741CH	105	66p	741CH	105	66p	
			741CN	80p	27p	741CN	80p	27p	741CN	80p	27p	
			747CN	70p		747CN	70p		747CN	70p		
			748CN	36p		748CN	36p		748CN	36p		
			NE531T	12p		NE531T	12p		NE531T	12p		
			NE531N	105p		NE531N	105p		NE531N	105p		

## OPTO

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# The New Most Important Link



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## DYNAMIC RIBBON MICROPHONE

### Specifications:

Frequency Response: 50 — 18 000 Hz  
Polar Pattern: Hypercardioid  
Output Level: 0.9mV/PA @ —60 dbm  
EIA Sensitivity Rating: —153 dbm  
Electrical Impedance: 200 ohms  
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Excellent anti-feedback characteristic over the whole frequency range.

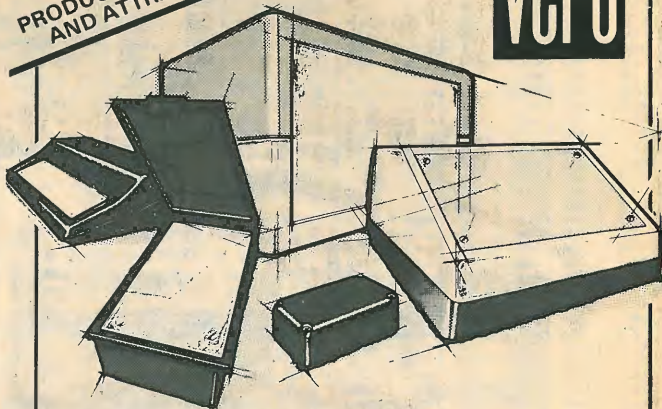
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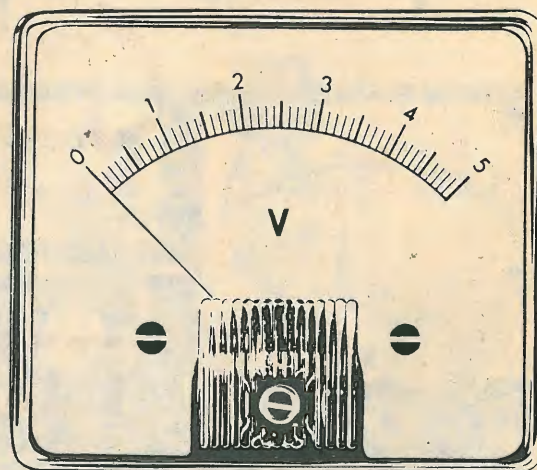
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Our new 1978 catalogue lists a whole range of plastic boxes to house all your projects. And we've got circuit boards, accessories, module systems, and metal cases — everything you need to give your equipment the quality you demand. Send 25p to cover post and packing, and the catalogue's yours.

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**HARRIS ELECTRONICS (London)**

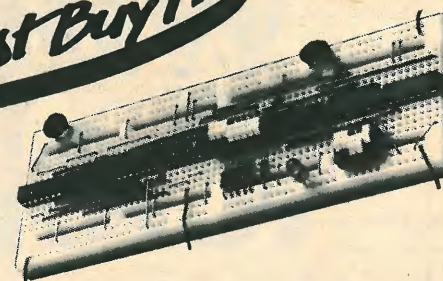
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FAST AND EASY FAST AND EASY FAST AND EASY FAST AND EASY

# Britain's Best Breadboard Buys!

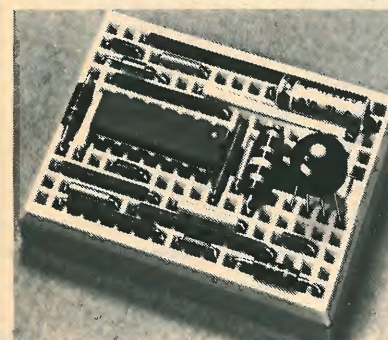
Best Buy 1.



## Lektrokit Super Strip SS2

Only £11.05 inc. p & p and VAT

Super Strip accepts *all DIP's*—as many as nine 14-pin at a time—and/or TO-5's and discrete components. With interconnections of any solid wire up to 20 AWG. *And no soldering.* Super Strip has 840 contact points, combining a power/signal distribution system with a matrix of 640 contacts in groups of 5. Distribution system has 8 bus-bars, each with 25 contact points.



Best Buy 2-8

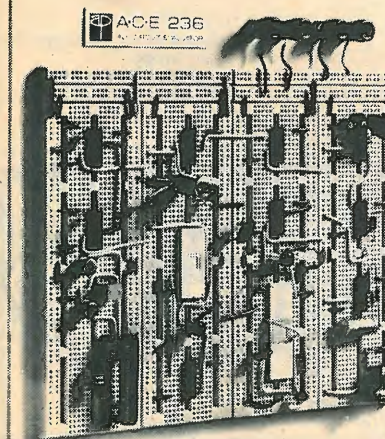
## Lektrokit Breadboards and Bus Strips

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BREADBOARDS			BUS STRIPS	
Model	Contacts	Price, each	Model	Price
264L	640	£8.32	212R	£1.78
248L	480	£6.65	209R	£1.62
234L	340	£5.75	206R	£1.45
217L	170	£3.25		

(All prices include p & p and VAT)



Best Buy 9-15

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# MHz electronics

2111 WEST CAMELBACK SUITE B PHOENIX, ARIZONA USA

(602) 242-3037

## R F Transistors

Type	Price	Type	Price	Type	Price
2N1561	\$15.00	2N5637	\$20.70	HEPS3006	\$19.90
2N1562	15.00	2N5641	4.60	HEPS3007	24.95
2N1692	15.00	2N5643	20.70	HEPS3010	11.34
2N1693	15.00	2N5764	27.00	HEPS5026	25.56
2N2857JAN	2.45	2N5862	50.00	MMCN918	1.00
2N2876	12.35	2N5913	3.25	MMT72	.61
2N2880	25.00	2N5922	10.00	MMT74	.94
2N2927	7.00	2N5942	49.50	MMT857	1.43
2N2947	17.25	2N5943	1.75	MMT960A	6.25
2N2948	15.50	2N5945	10.90	PT4186B	3.00
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2N3307	10.50	2N6094	5.75	JO4030	15.60
2N3309	3.90	2N6095	10.35	40281	10.90
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2N3824	3.20	2N6166	36.80		
2N3866	1.09	2N6439	43.45	FETs	
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2N3866JANTX	4.85	MM1550	10.00	40673	1.35 or 10/10.00
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2N3927	11.50	MM1601	5.50	MEM631	.63 or 10/ 5.30
2N3950	26.25	MM1602	7.50		
2N3961	6.60	MM1607	8.65		
2N4072	1.70	MM1620	17.50		
2N4073	2.00	MM1661	15.00		
2N4135	1.24	MM1669	17.50		
2N4427	20.00	MM1943	3.00		
2N4430	8.60	MM2605	3.00		
2N4440	6.30	MM2608	5.00		
2N4957	4.35	MM8002	2.05		
2N4958	4.35	MM8006	2.12		
2N4959	2.12	MRF245	31.05		
2N4976	19.00	MRF304	43.45		
2N5070	13.80	MRF501	.49		
2N5090	6.90	MRF504	6.95		
2N5108	3.90	MRF509	4.90		
2N5109	1.55	MRF511	8.60		
2N5160	3.34	MRF517	20.70		
2N5177	20.00	MRF577	20.70		
2N5179	.68	MRF8004	1.90		
2N5184	2.00	HEP76/S3014	4.95		
2N5216	47.50	HEPS3002	11.30		
2N5589	4.60	HEPS3003	29.88		
2N5590	6.30	HEPS3005	9.55		
2N5591	10.35				

## MHz Electronics Kits :

## Kit # 1

Motorola MC14410CP CMOS Tone Generator. CMOS Tone Generator uses 1 MHz crystal to produce standard dual frequency dialing signal. Directly compatible with 12 key Chomeric Touch Tone Pads. Kit includes the following:

- 1 Motorola MC14410CP Chip
- 1 1 MHz Crystal
- 1 PC Board

And all other parts for assembly. NOW ONLY \$15.70

## Kit # 2

Fairchild 95H90DC Prescaler 350MHz. 95H90DC Prescaler divides by 10 to 350 MHz. This kit will take any 35MHz Counter to 350 MHz. Kit includes the following:

- 1 Fairchild 95H90DC Chip
- 1 2N5179 Transistor
- 2 UG-88/U BNC Connectors
- 1 PC Board

And all other parts for assembly. Now Only \$19.95

## Kit # 3

Fairchild 11C90DC Prescaler 650MHz. 11C90DC Prescaler divides by 10/100 to 650 MHz. This counter will take any 65 MHz Counter to 650MHz. or with a 82590 it will take a 6.5 MHz Counter to 650MHz. Kit includes the following:

- 1 Fairchild 11C90DC Chip
- 1 2N5179 Transistor
- 2 UG-88/U BNC Connectors
- 1 LM/MC7805 Voltage Regulator
- 1 50volt Lamp Bridge
- 1 LED Indicator
- 1 PC Board

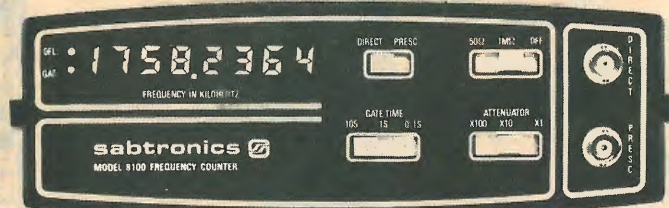
And all other parts for assembly. Now Only \$29.95

## FAIRCHILD VHF and UHF Prescaler Chips

95H90DC	350MHz Prescaler Divide by 10/11	\$ 8.95
95H91DC	350MHz Prescaler Divide By 5/6	8.95
11C90DC	650MHz Prescaler Divide By 10/11	15.95
11C91DC	650MHz Prescaler Divide By 5/6	15.95
11C83DC	1GHz Divide by 248/256 Prescaler	29.90
11C70DC	600MHz Flip/Flop with reset	12.30
11C58DC	ECL VCM	4.53
11C44DC	Phase Frequency Detector (MC4044P/L)	3.82
11C24DC	Dual TTL VCM (MC4024P/L)	3.82
11C06DC	UHF Prescaler 750MHz D Type Flip/Flop	12.30
11C05DC	1GHz Counter Divide by 4	74.35
11C01FC	High Speed Dual 5-4 Input NO/NOR Gate	15.40

Crystal Filters. Tyco 001-19880 same as 2194F  
10.7MHz Narrow Band Crystal Filter  
3 db bandwidth 15khz minimum 20 db bandwidth 60khz minimum 40 db bandwidth 150khz minimum  
Ultimate 50 db : Insertion loss 1.0db Max. Ripple 1.0db Max. Ct. 0+ ~ 5pf. Rt. 3600 Ohms.  
Now Only \$ 5.95

WW-027 FOR FURTHER DETAILS



Model 8100 Frequency counter  
Kit £ 69.95  
assembled tested: £ 84.95  
(plus p.p. £ 3.50 and VAT at 8%)

## The Winners

These two products are our best sellers!

The two products shown above from Sabtronics are our best selling products. Both these products compare with similar equipment selling for at least £ 150.00. Is there more to these products than value? Let's take a closer look.

## The Frequency Counter Model 8100

It employs LSI Technology, has the performance and characteristics you demand, guaranteed frequency range of 20 Hz to 100 MHz; selectable hi/lo impedance; superior sensitivity; selectable resolution and selectable attenuation. Plus an accurate time base with excellent stability. An 8 digit LED Display features floating decimal point, leading zero suppression and overflow indicator.

## Brief specifications:

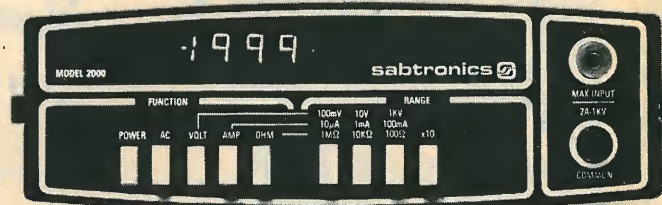
Frequency Range: 20 Hz to 100 MHz guaranteed, (10 Hz to 130 MHz typical) - Sensitivity: 10 mV RMS,

20 Hz to 50 MHz (5 mV typical); 15 mV RMS, 50 MHz to 100 MHz (10 mV typical) - Selectable impedance: 1 MΩ/25 pF or 50Ω - Attenuation: X1, X10 or X100 - Accuracy: ± 1 Hz plus time base accuracy - Aging Rate: ± 5 ppm/yr - Temperature Stability: ± 10 ppm, 0° to 50° C - Resolution: 0.1 Hz, 1 Hz, 10 Hz selectable - Display: 8-digit LED, floating DP, overflow indicator - Overload Protection - Power Requirement: 9-15 VDC. Optional prescaler will be available from around March 1979.

## The DMM Model 2000

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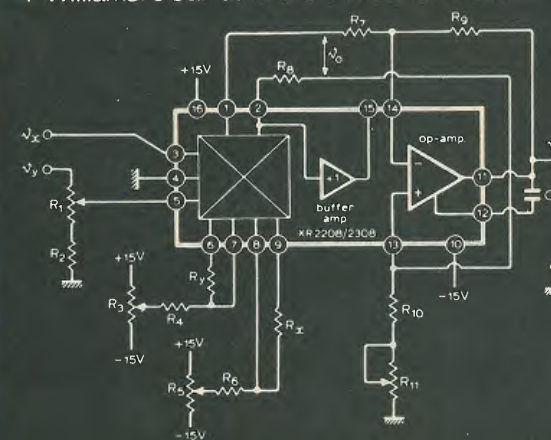
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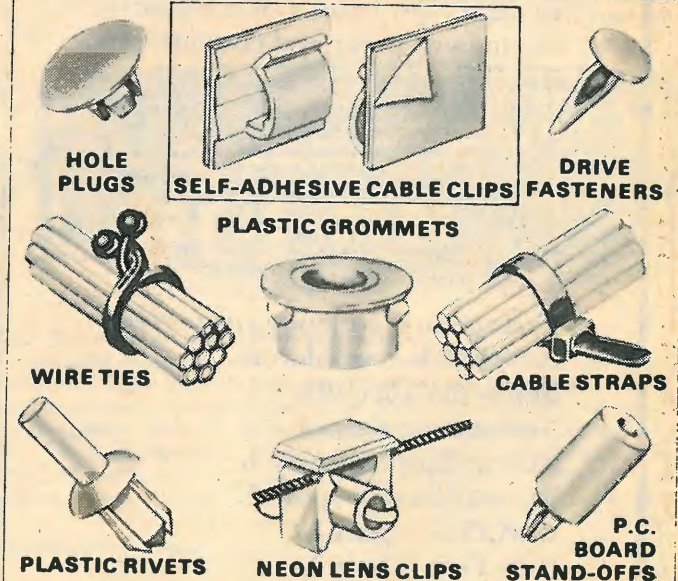


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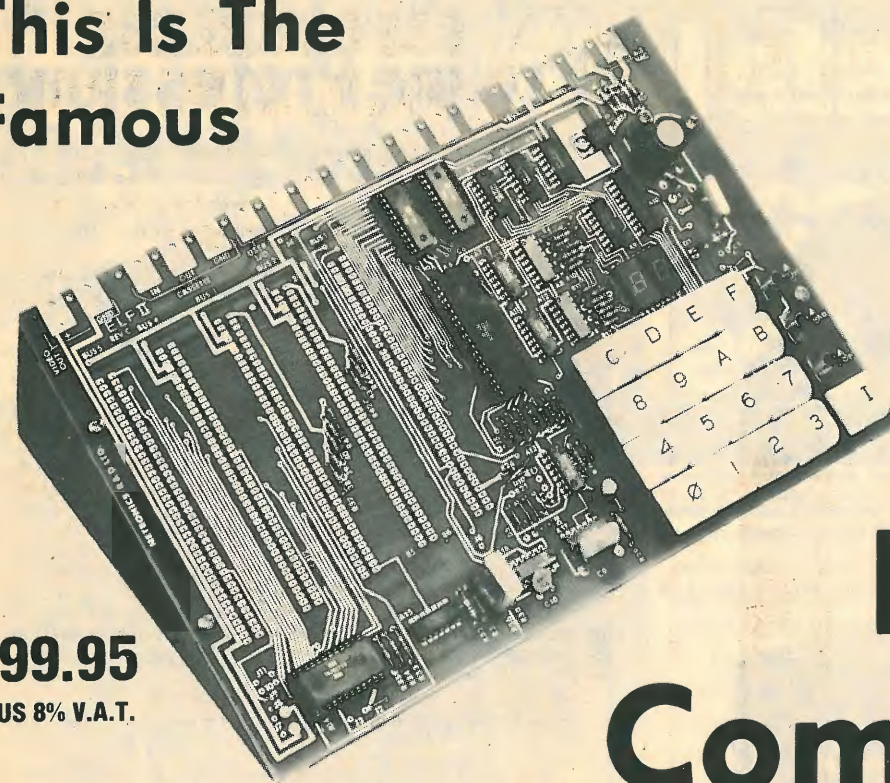


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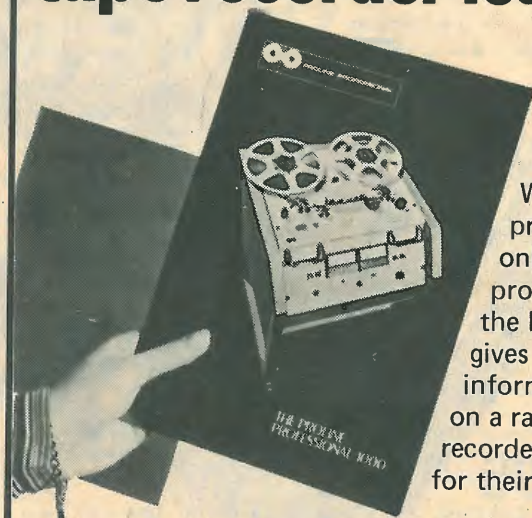
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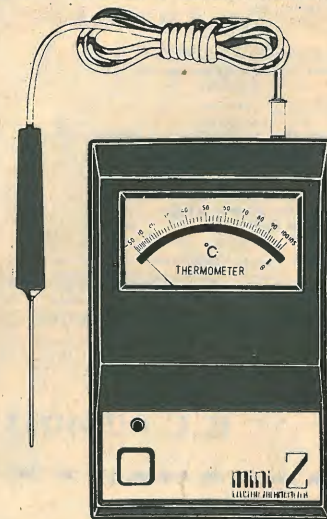
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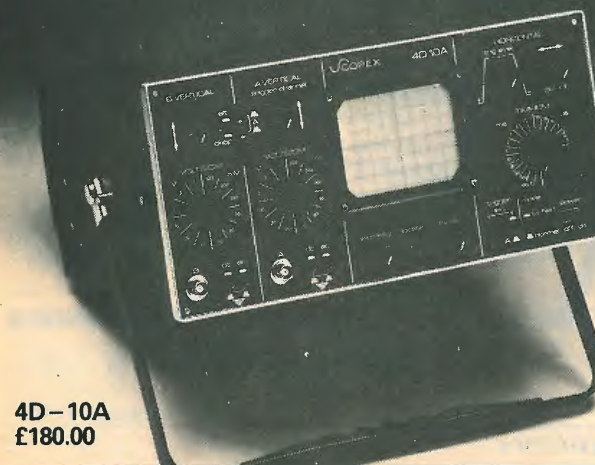
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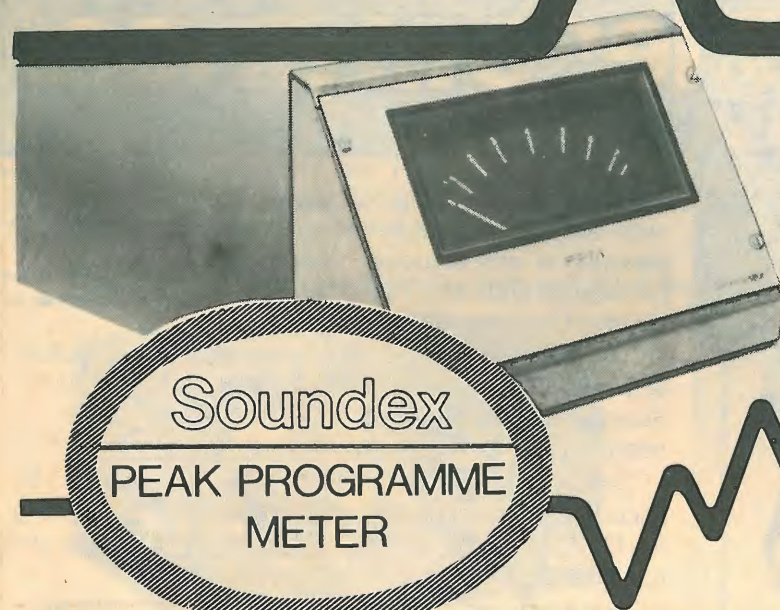
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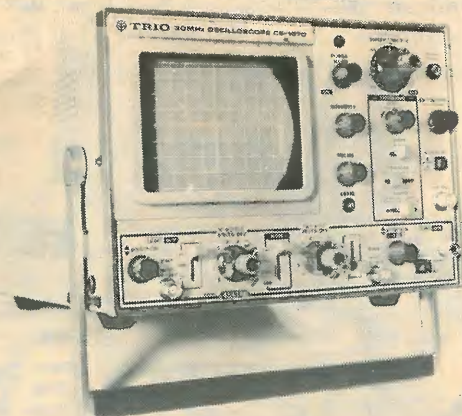
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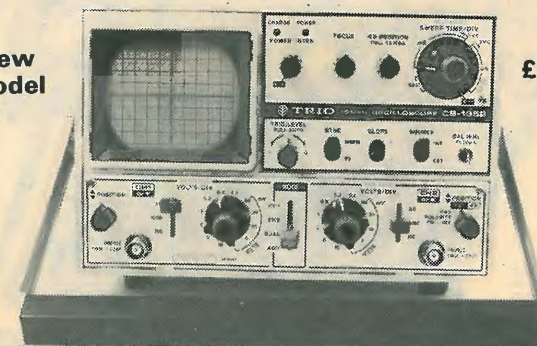
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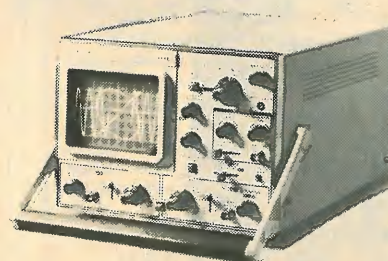
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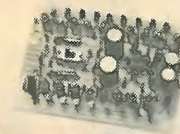
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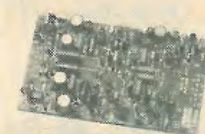
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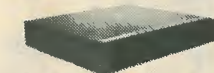
**MC 1 — PRE-AMPLIFIER.** Suitable for nearly all moving-coil cartridges. Sensitivity 70/170μV switchable on the p.c.b. This module brings signals from the now popular low output moving-coil cartridges up to 3.5mV (typical signal required by most pre-amp disc inputs). Can be powered from a 9V battery or from our REG 1 regulator board.

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**POWER AMPLIFIERS.** It would be pointless to list in so small a space the number of recording studios, educational and government establishments, etc. who have been using CRIMSON amps satisfactorily for quite some time. We have a reputation for the highest quality at the lowest prices. The power amp is available in five types, they all have the same specification: T.H.D. typically .01% any power 1kHz 8 ohms; T.I.D. insignificant; slew rate limit 25V/μs; signal to noise ratio 110dB; frequency response 10Hz-35kHz, —3dB; stability unconditional; protection drives any load safely; sensitivity 775mV (250mV or 100mV on request), size 120x80-25mm.

**POWER SUPPLIES.** We produce suitable power supplied which use our superb TOROIDAL transformers only 50mm high with a 120-240 primary and single bolt fixing (includes capacitors/bridge rectifier).

**POWER AMPLIFIER KIT.** The kit includes all metalwork, heatsinks and hardware to house any two of our power amp modules plus a power supply. It is contemporarily styled and its quality is consistent with that of our other products. Comprehensive instructions and full back-up services enables a novice to build it with confidence in a few hours.



### POWER AMPLIFIER MODULES

CE 608 60W/8 ohms 35-0-35v	£16.30
CE 1004 100W/4 ohms 35-0-35v	£19.22
CE 1008 100W/8 ohms 45-0-45v	£23.22
CE 1704 170W/4 ohms 45-0-45v	£29.12
CE 1708 170W/8 ohms 60-0-60v	£31.90

### TOROIDAL POWER SUPPLIES

CPS 1 for 2xCE 608 or 1xCE 1004	£14.47
CPS 2 for 2xCE 1004 or 2xCE 608	£16.82
CPS 3 for 2xCE 1008 or 1xCE 1704	£17.66
CPS 4 for 1xCE 1008	£15.31
CPS 5 for 1xCE 1708	£22.68
CPS 6 for 2xCE 1704 or 2xCE 1708	£23.98

### HEATSINKS

Light duty, 50mm, 2 C/W	£1.30
Medium power, 100mm, 1.4 C/W	£2.20
Disco group, 150mm, 1.1 C/W	£2.85
Fan, 80mm, state 120 or 240v	£18.50
Fan mounted on two drilled 100mm heatsinks	
2x.4 C/W, 65 C max. with two 170W modules	£29.16

### THERMAL CUT-OUT, 70 C

<b>POWER AMP KIT</b> £32.40	
<b>PRE-AMPS:</b>	
These are available in two versions — one uses standard components, and the other (the S) uses MO resistors where necessary and tantalum capacitors.	
CPR 1	£29.49
MC1	£18.50
CPRIS	£39.98
MC1S	£29.49
<b>POWER SUPPLY:</b>	
REG1	£6.75
TR6	£1.75
<b>BRIDGE DRIVER, 8DI</b>	
Obtain up to 340W using 2x170W amps and this module.	
B01	£5.40

## CRIMSON ELEKTRIK

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All prices shown are UK only and include VAT and post. COD 90p extra, £100 limit. Export is no problem, please write for specific quote. Send large SAE or 3 International Reply Coupons for detailed information. Distributor: Minic Teleprodukt, Box 12035, S-750 12 Uppsala 12, Sweden.

WW—032 FOR FURTHER DETAILS

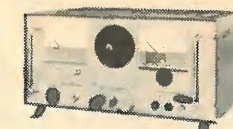
### PRODUCTION TESTING

### DEVELOPMENT

### SERVICING

### POWER UNITS

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3 OUTPUTS



Type 250VRU/30/25

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counters we have produced a

NEW 200MHz COUNTER KIT specially for home constructors

Our new K200 counter, although small, is a no-compromise design. It offers:

- A full 8-digit LED display.
- A frequency range of 10Hz to 200MHz.
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- 5/6 volt operation from batteries of mains PSU.
- Power consumption of only 1W maximum at maximum frequency.
- A crystal oscillator at 5MHz which doesn't need any special setting up equipment.
- Small size 4" x 2" x 1" — Uses only 4 i.c.s.
- Assembly time of about two hours.
- Full illustrated assembly instructions.

The K200 consists of 2 PCB assemblies, one being the complete input and counter unit, the other, the display unit. Both units are available in kit or assembled/tested module form. Prices, INCLUDING VAT:

Input/Counter Kit	£59.00	Display Kit	£12.96
Input/Counter Module	£68.50	Module	£16.64

(Add 75p for Post and Packing)

This NEW CATRONICS model K200 complements our DFM5, a 250 MHz, 7 DIGIT MAINS/12v HIGH QUALITY Frequency Counter, and the DFM 500 — a REAL 500 MHz — try some of the others actually at this frequency.

Both are absolute value for money and are available now with better than 1 in 10 reference oscillators as / S models.

SPECIAL PRICES INCLUDING VAT

DFM5	£148.50	DFM5/S	£191.70
DRM500	£177.12	DFM500/S	£220.86

Add £5 for Special Packing & Securicor

**Catronics** (Dept. 922)  
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Type 130 is a compact 2/3 inch camera with an unusually high specification for a camera of this size. It is a sturdy, reliable camera of great versatility.

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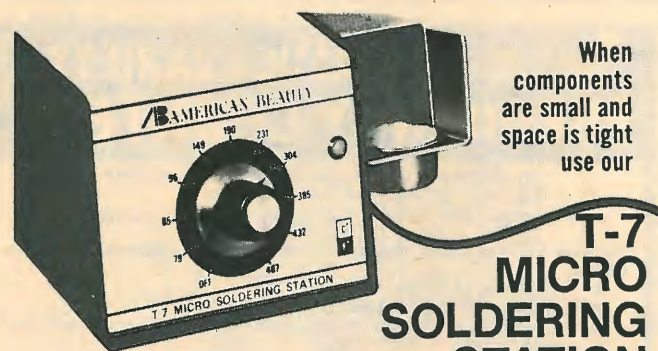
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Slip-on tip in use.  
2/3 actual size.

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S 500D — dual channel  
up to 500W/RMS per  
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The **H||H** 'S' range is designed to handle heavy industrial usage in the fields of vibrator driving, variable frequency power supplies and servo motor systems.

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Dual Channel  
19" rack mount 3½" high  
500w r.m.s. into 2.5 ohms per channel  
900w r.m.s. in bridge mode  
DC-20 KHZ at full power  
0.005% harmonic distortion (typical) at  
300w r.m.s. into 4 ohms at 1 KHZ  
3KW dissipation from in-built force cooled  
dissipators

#### S 250D

Single Channel  
19" rack mount 3½" high  
500w r.m.s. into 2.5 ohms  
Retro-convertible to dual channel  
DC-20 KHZ at full power  
Full short and open circuit protection  
Drives totally reactive loads with no  
adverse effects

A complete range of matching transformers and peripheral equipment for closed loop, constant current and voltage use are available.

Alternative input and output termination to order. Rack case for bench use built to specifications. For complete data write or call.



**Kirkham Electronics**

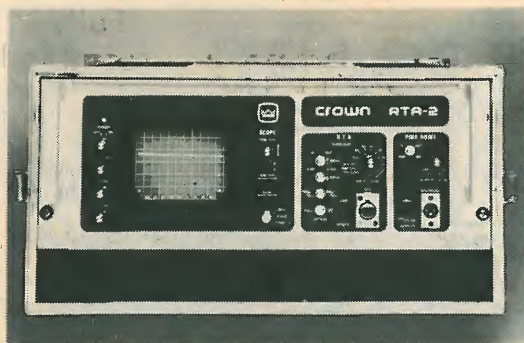
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FRANCHISED COMMERCIAL AND INDUSTRIAL AGENTS FOR **H||H** ELECTRONIC  
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## New from AMCRON

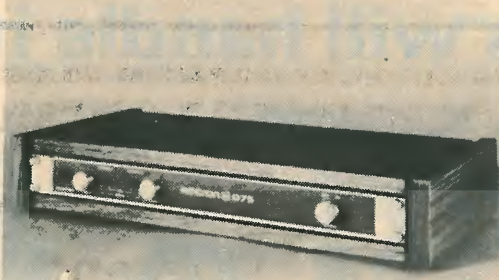
### Real Time Analyser RTA2



The Amcron RTA2 Real Time Analyser is designed as much for use as a production tool as it is for on-site audio analysis of theatres, and recording studios. A flight case is available.

- ★ 5" CRT Display
- ★ Internal Pink Noise Source
- ★ 1/3 or 1 octave Display
- ★ Frequency range 20 - 20kHz
- ★ Outputs for X-Y Recorders
- ★ Compatible with any microphone
- ★ Price £1,960 ex. VAT

### POWER AMPLIFIER D75



The AMCRON D75 power amplifier replaces the previous model D60. Employing completely new type circuitry it offers also many new features but without any increase in the price.

- ★ New Amcron IOC comparator.
- ★ Balanced XLR input connectors.
- ★ Signal Presence indicators.
- ★ Separate Signal / chassis earth.
- ★ 45 watts into 8 ohms per channel.
- ★ Price £230 ex. VAT.

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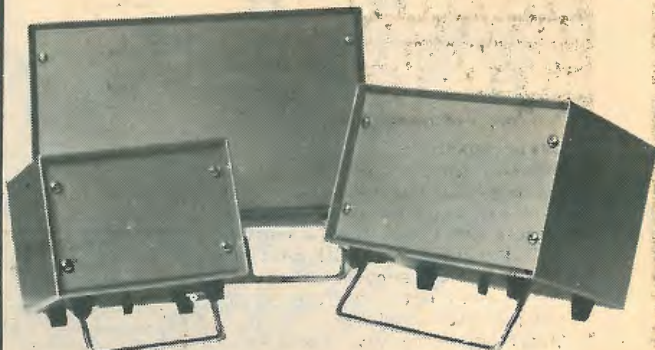
DC300A amplifier 500 watts / chan.	£550
D150A amplifier 200 watts / chan.	£350
VFX2A Electronic Variable Filter	£270
EQ2 Equaliser	£599
IC150A Pre-amplifier	£260
IMA Intermodulation Distortion Analyser unit	£610

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## OLSON

### MINICASES



Type	Overall Dimension			Case no vents	Case with vents	Chrome leg
	Width	Height	Depth			
21	6½"	4½"	4½"	—	5.88	1.15
22	8½"	5½"	5½"	—	6.50	1.15
23	10½"	6½"	6½"	—	7.70	1.30
24	12½"	7½"	7½"	—	8.45	1.30
25A	6½"	4½"	4½"	5.60	6.20	1.30
25B	6½"	4½"	6¼"	6.00	6.60	1.30
26A	8½"	5½"	6¼"	7.90	8.50	1.30
26B	8½"	5½"	8¼"	8.25	8.85	1.30
27A	12½"	7½"	5½"	8.65	9.35	1.30
27B	12½"	7½"	8"	9.35	10.05	1.30
28A	14"	10½"	6½"	10.20	10.90	—
28B	14"	10½"	8½"	11.15	11.85	—
29A	10"	4"	6"	7.15	7.75	1.30
29B	10"	4"	8"	7.53	8.13	1.30
30A	12"	5"	6"	7.80	8.50	1.30
30B	12"	5"	8"	8.15	8.85	1.30
31A	14"	6"	6"	8.50	9.20	1.30
31B	14"	6"	8"	8.93	9.63	1.30
61	15½"	7½"	9½"	—	13.00	—
62	17½"	8½"	9½"	—	15.00	—
63	16½"	9½"	9½"	—	15.00	—
64	15½"	7½"	12½"	—	15.00	—
65	17½"	8½"	12½"	—	17.15	—
66	16½"	9½"	12½"	—	17.15	—

Postage & VAT extra

### INSTANT TRUNKING SYSTEM!

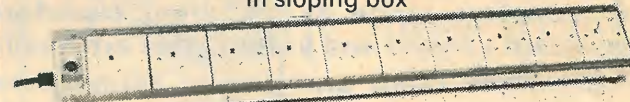


Ready to use. Internal wiring suitable for 30 amp

TR6 — 6 sockets switched	£21.50
TR9 — 9 sockets switched	£25.50
P&P £1.85	+VAT

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NEW! 10 sockets switched in sloping box



Type 13A/10SW £27.50. P&P £1.85 + VAT



COMPLETE WITH 6FT. CABLE AND 13 AMP FUSED PLUG

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TEL: 01-739 2343

WW—046 FOR FURTHER DETAILS

## MODEL 756 FULL ASCII KEYBOARD

LOW COST!

Fully Assembled



NUMERIC KEYPAD

- 756 KEYBOARD
- Intended for professional micro-processor applications.
- This one Keyboard will meet most present and future requirements.
- Full 128-character ASCII 8-bit code
- Tri-mode MOS encoding.
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- Upper and lower case characters generated by keyboard with latching shift-lock.
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- Size 305 X 140 X 32mm (12¼ X 5½ X 1¼ in)
- MOS/DTL/TTL compatible outputs.
- New guaranteed OEM grade components.
- Needs +5 and -12V supply

- Board has space for small low cost DC/DC converter so that entire unit operates off single 5V rail.

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### pocket-sized 500 MHz FREQUENCY COUNTER

(CM 7044)

for UHF/VHF or CB  
mobile radio

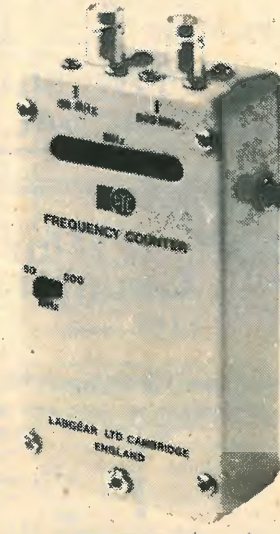
- \* ±2ppm accuracy (0-0002%)
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- \* Exceptional value

Frequency Ranges  
(Switched)  
(a) 10MHz — 50MHz  
(b) 40 MHz — 500 MHz

Sensitivity  
50 MHz range, 30 mV  
500 MHz range, 50 mV

Display  
7 digit, 7 segment LED display

Dimensions  
4.75" 2.6" 1.25"  
(121 67 32mm)



**Labgear**

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a happy  
event...



... a new addition to our family of amplifiers. The VTN 30. The new baby weighs in at 30 watts, and has a maximum of 3 inputs.

The other members of the Vortexion family are the system 2000, 50/70 watt and CP50 mains/battery amplifiers.

Contact Jennifer Hall — VORTEXION DIVISION.

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Telex Casint G 22574; Telegrams: Electronic Wallington.

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## Transformers

Efficient design and manufacture ensures that all our transformers have a low heat-rise at full load, making a more reliable and safer product. Safety is also the reason behind our clip-on terminal insulators, which can make our transformers 'touch-proof' in use.

Lascar Components  
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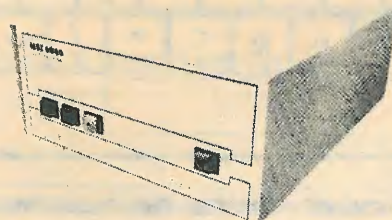


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# seed

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- NEW** Period Measurement  
1 us to 1 sec.
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10 mv @ 150 MHz
- The New Model CTR-2A Series Counters are designed and built to the highest standards to fulfil the needs of commercial communications, engineering labs and serious experimenters. With an accuracy of +.00005%, the CTR-2A can handle the most critical measurements and is about half the cost of other commercial counters.
- If you need a reliable counter at an affordable price, the CTR-2A is the answer.
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  - Built-in VHF-UHF Prescaler
  - Automatic Dip Placement
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7410	12p	74118	82p
7411	19p	74119	140p
7412	17p	74121	25p
7413	25p	74123	40p
7414	48p	74125	35p
7415	24p	74126	35p
7417	24p	74132	50p
7420	12p	74141	50p
7421	22p	74142	200p
7422	18p	74145	58p
7423	22p	74147	110p
7425	22p	74148	90p
7426	24p	74150	70p
7427	24p	74151	50p
7428	24p	74153	50p
7430	12p	74154	85p
7432	22p	74155	52p
7433	32p	74156	52p
7437	22p	74157	53p
7438	22p	74160	80p
7440	13p	74161	65p
7441	52p	74162	65p
7442	43p	74163	65p
7443	75p	74164	70p
7444	75p	74165	70p
7446	70p	74166	80p
7447	55p	74167	180p
7448	58p	74170	125p
7451	14p	74171	400p
7452	13p	74172	400p
7454	14p	74174	65p
7455	14p	74175	65p
7456	14p	74176	58p
7457	14p	74177	58p
7458	14p	74178	58p
7459	14p	74179	58p
7460	14p	74180	58p
7461	14p	74181	145p
7462	14p	74182	60p
7463	14p	74183	60p
7464	14p	74184	60p
7465	14p	74185	60p
7466	14p	74186	60p
7467	14p	74187	60p
7468	14p	74188	60p
7469	14p	74189	60p
7470	14p	74190	72p
7471	14p	74191	72p
7472	14p	74192	60p
7473	14p	74193	60p
7474	14p	74194	60p
7475	14p	74195	60p
7476	14p	74196	60p
7477	14p	74197	60p
7478	14p	74198	60p
7479	14p	74199	60p
7480	14p	74200	60p
7481	14p	74201	60p
7482	14p	74202	60p
7483	14p	74203	60p
7484	14p	74204	60p
7485	14p	74205	60p
7486	14p	74206	60p
7487	14p	74207	60p
7488	14p	74208	60p
7489	14p	74209	60p
7490	14p	74210	60p
7491	14p	74211	60p
7492	14p	74212	60p
7493	14p	74213	60p
7494	14p	74214	60p
7495	14p	74215	60p

## Transistors

AC125	18p	BC558	12p
AC126	18p	BC559	12p
AC127	18p	BC560	12p
AC128	18p	BC561	12p
AC129	18p	BC562	12p
AC130	18p	BC563	12p
AC131	18p	BC564	12p
AC132	18p	BC565	12p
AC133	18p	BC566	12p
AC134	18p	BC567	12p
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AC136	18p	BC569	12p
AC137	18p	BC570	12p
AC138	18p	BC571	12p
AC139	18p	BC572	12p
AC140	18p	BC573	12p
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AC143	18p	BC576	12p
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AC146	18p	BC579	12p
AC147	18p	BC580	12p
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AC149	18p	BC582	12p
AC150	18p	BC583	12p
AC151	18p	BC584	12p
AC152	18p	BC585	12p
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AC154	18p	BC587	12p
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AC168	18p	BC601	12p
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AC184	18p	BC617	12p
AC185	18p	BC618	12p
AC186	18p	BC619	12p
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AC285	18p	BC718	12p
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AC291	18p	BC724	12p
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AC294	18p	BC727	12p
AC295	18p	BC728	12p
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AC297	18p	BC730	12p
AC298	18p	BC731	12p
AC299	18p	BC732	12p
AC300	18p	BC733	12p
AC301	18p	BC734	12p
AC302	18p	BC735	12p
AC303	18p	BC736	12p
AC304	18p	BC737	12p
AC305	18p	BC738	12p
AC306	18p	BC739	12p
AC307	18p	BC740	12p
AC308	18p	BC741	12p
AC309	18p	BC742	12p
AC310	18p	BC743	12p
AC311	18p	BC744	12p
AC312	18p	BC745	1

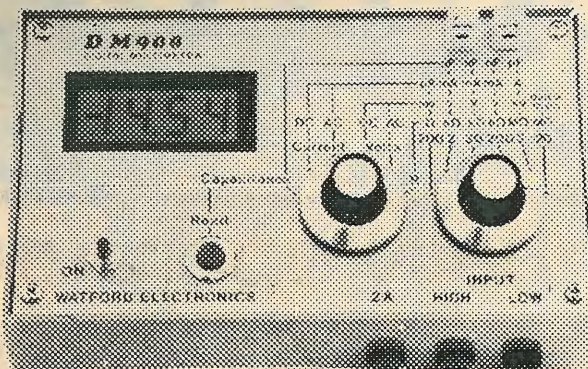


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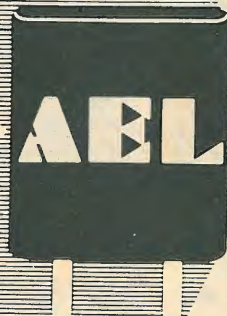
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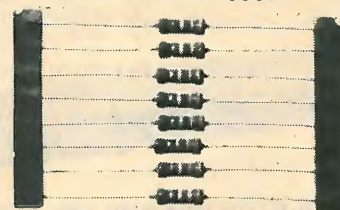
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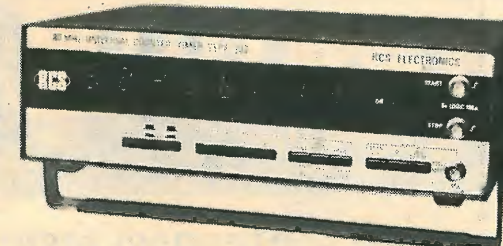


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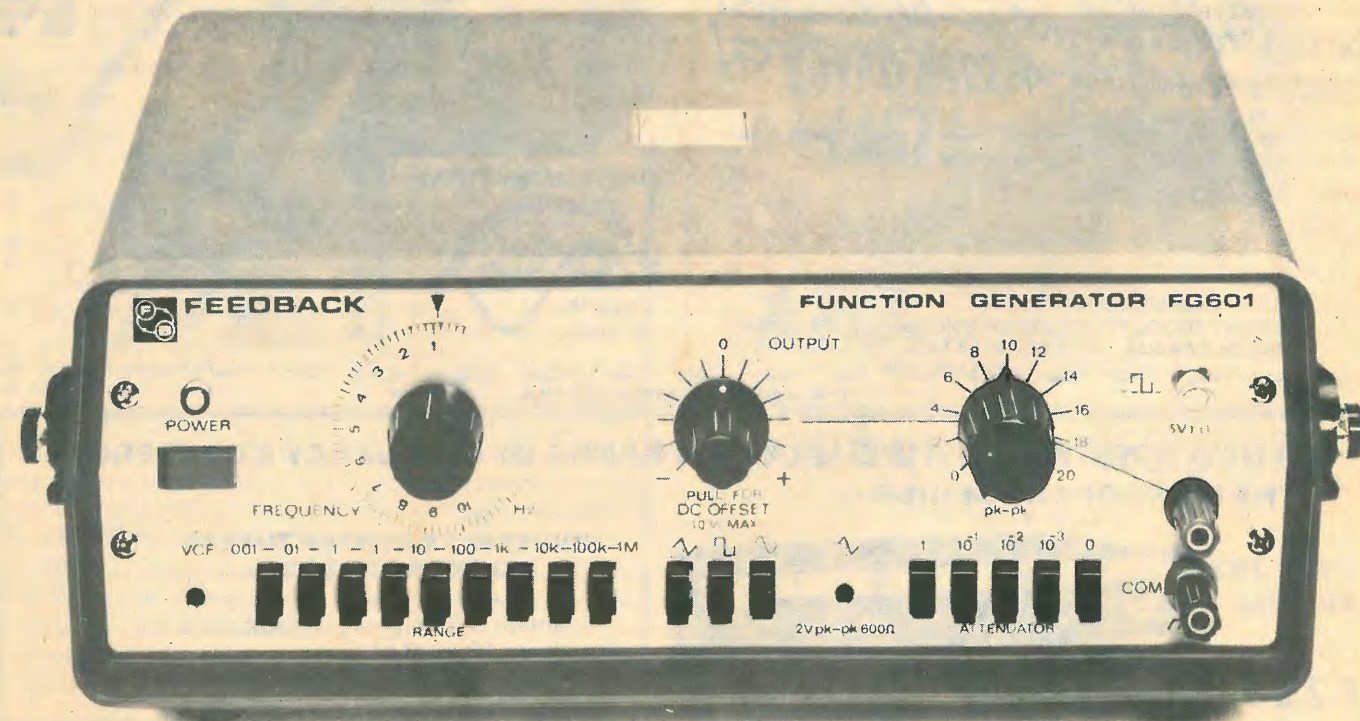
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WW — 005 FOR FURTHER DETAILS

## wireless world

### The map is not the territory

One can't help wondering why some of the controversies in our correspondence columns about physical realities in electronics seem to get so involved. The disputants seem to be men of intelligence and probity, so why can't they get at the truth, the reality, which must be lurking in there somewhere? Ah, but here's the rub. The truth is one thing, the reality another. As the semanticist Korzybski observed, the map is not the territory. In electronics, as in all of science, what carries the truth (or its antithesis) is a statement or expression in words, mathematics or diagrams, a product of the mind alone; the reality is something out there beyond the mind which we can only approach in a restricted way through the senses, either directly or indirectly by using them to observe meter readings, oscilloscope traces and so on. It does seem that some of our correspondents get confused between these two very different things, especially when they use the word "truth" for rhetorical purposes.

Unfortunately "truth" is an extremely woolly word. Its meanings can range from the spiritual ("I am the way, the truth . . ." of Jesus; "Beauty is truth, truth beauty" of Keats) to the strictly formal (as in deductive logic and the truth tables of our electronic logic systems). Somewhere between, perhaps, comes the everyday idea of truth, as used in law and described technically as the "correspondence theory" of truth. In our scientific terms this is the correspondence between the conceptual model or hypothesis (the map) and the set of sometimes disconnected experiences or observations we get from the real world (the territory). Aristotle's version of truth (in the *Metaphysics*): "To say of what is that it is not, or of what is not that it is, is false; while to say of what is that it is, or of what is

not that it is not, is true" doesn't seem to get us very far and is apparently tautologous. But it does have the virtue of distinguishing between statements and reality. And in fact it has been used by the 20th-century philosopher Tarski as the basis of a precisely defined correspondence theory of truth.

Some thinkers have claimed that human beings, their consciousness and thoughts, don't have to come into the matter of truth at all. Frege, the logician, for example, said " . . . the thought, for example, which we expressed in the Pythagorean theorem is timelessly true, true independently of whether anyone takes it to be true." He seemed to have forgotten that logic itself is a human invention. With the phenomena of reality, however, we certainly have only ourselves to depend on. Bishop Berkeley went so far as to suggest that there might not be anything material around beyond the phenomena we experience. How can we be sure that material objects actually exist when all we have is our notoriously unreliable perceptions to go by? Kant said this was absurd because it implied that there can be appearance without anything that appears, and he went on to postulate "noumena" or things-in-themselves (the real stuff), as distinct from mere phenomena.

One of our great difficulties may be the fact that the very principle of objectivity, so dear to science, is itself really an article of faith, a product of the human mind and therefore of subjectivity. The rigidly "scientific" point of view is both naive and dishonest because it takes for granted, without explicitly mentioning it, the point of view of human consciousness, by which the world must be *my* world. As such objectivity has its limitations. We still don't know, for example, whether the electron is a particle or a packet of waves.

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# Computer buses

Conventions of communication between the components of a computer

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Subsystems of a computer are usually connected together by a common bus, instead of by connecting each subsystem to every other one individually. A bus is a passive device — just a bundle of wires. However, to make it work, a logical superstructure of protocol and convention must be adopted, and observed correctly by every device connected to the bus. If any one device fails to observe the protocol, it can disrupt all communications along the bus. This is the price of a common bus.

THE COMPONENTS of a computer system include a processing unit (often a microprocessor chip), a store, and input-output interfaces to enable it to communicate with the world outside. The diagram of Fig. 1, where the processor is at the centre, communicating with the store on one side and input-output interfaces on the other, illustrates the way that early computers were organized. However, it is apparent that in this configuration the processor constitutes a bottleneck, for usually the input-output data is placed into or taken out of store rather than being processed immediately. This leads to the inclusion of the dashed line in the figure, as a "direct memory access" channel. Early minicomputers were built in this way.

This kind of interconnection, where every device is connected to every other, quickly becomes cumbersome as more devices are added to the system. Recent minicomputers, and almost all microprocessor systems, use the alternative structure of Fig. 2, where a common bus is used to connect the various devices. "Bus" is a contraction of the Latin "omnibus", which means "for all" — indicating that the bus is used for all data transfers between subsystems.

It may seem that the bus structure has re-introduced the bottleneck of the processor-centred computer model, only now it is the bus and not the processor which is responsible for the blockage. This is perfectly true, but since the bus is such a simple device — just a bundle of wires — the bottleneck can be relieved by making data transfers on it go very fast, whereas transfers through the processor are much slower.

It is obviously important that transfers along the bus are received

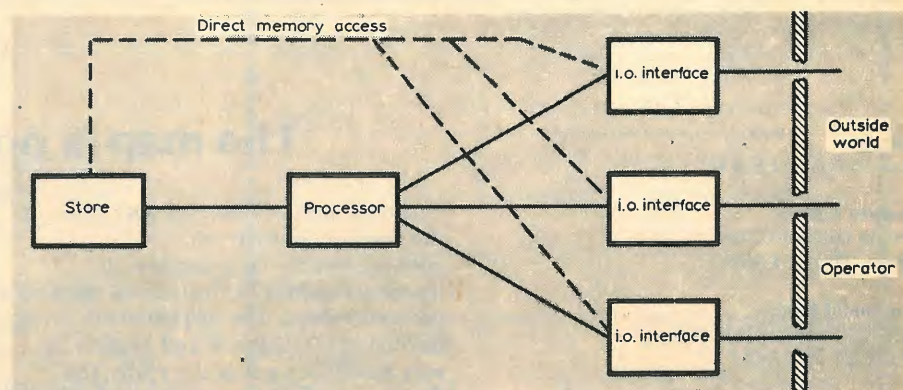


Fig. 1. Processor-centred computer model.

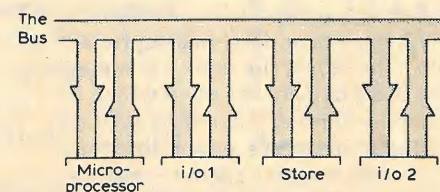


Fig. 2. Bus-centred computer model.

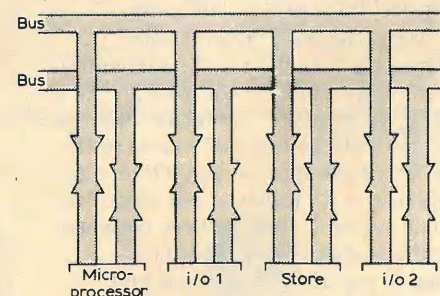


Fig. 3. Address and data buses.

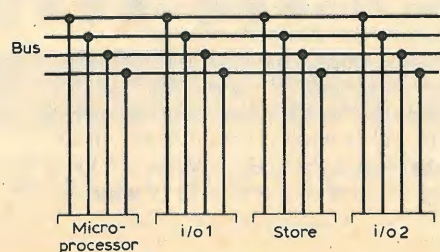


Fig. 4. Individual lines of a 4-bit bus.

only by the device for which they are intended. If i/o device 1 transfers data to the store, for example, it should not be received by i/o device 2 as well (or instead!). This is usually accomplished by dividing the bus into two parts: an address bus and a data bus, as in Fig. 3. A sending device puts the address of the intended receiver on the address bus, and the data for it on the data bus. Devices ensure that they only accept the data from the bus when the address bus indicates that they are the intended recipient. Addresses, like data, are simply binary numbers with a specified number of bits. Typically, the address bus is 16 bits wide and the data bus is 8 bits wide. This allows for addressing up to  $2^{16} = 65,536$  different devices. There is hardly likely to be this number of separate devices on the bus, of course: the number is so large because some devices — notably stores — often incorporate many different addresses. A single store may easily contain  $2^{15} = 32,768$  separate locations, each of which is addressable individually on the bus and would use up half of the "address space" of a 16-bit address bus.

The bus of Fig. 2 is redrawn in Fig. 4 to show the individual lines, assuming that it is 4 bits wide. Notice that although two channels are used in Fig. 2 to show the information coming off the bus separately from that going onto it, both channels are implemented on the same piece of wire in Fig. 4. The information on the wire is bi-directional. Of course, there is nothing special about the wire itself — any piece of wire can be driven from either end. It is the way that it is driven that makes it bi-directional. The bi-directional nature of the lines makes ordinary t.t.l. gates unsuitable for putting information on to buses, and there

are two commonly-used alternative methods: open-collector gates and tri-state gates.

## Driving a bus

Ordinary transistor-transistor logic (t.t.l.) suffers from the disadvantage that the outputs of two logic gates cannot be connected together. Thus if two devices are connected to the bus as in Fig. 5 and each may at some time put a signal on to it, ordinary t.t.l. gates cannot be used to drive the bus.

The reason is as follows. T.t.l. gates can only assume logic states 0 or 1. Suppose that device 2 does not want to put data on to the bus during a certain period of time. If it tries to put logic 0 on the bus through its bus driver, and if device 1 — which does want to put data on to the bus — assumes logic 1, the two devices will be fighting for the bus, one trying to pull it up to 1 while the other tries to pull it down to 0. Similarly, if device 2 tries to detach itself from the bus by driving it with logic 1, it will interfere with a device which wishes to drive the bus to logic 0.

T.t.l. gates are not designed for this situation, and it is not possible to forecast the result of such a fight by considering the gates as logic elements. One must look at the detailed circuitry of the output stage of a gate, shown in Fig. 6. Two transistors are arranged in a "totem-pole" configuration, one providing active pull-up through the top transistor when the output is to be high, and the other providing active pull-down through the bottom one when it is to be low. This arrangement is used to increase the switching speed of the device from one logic level to the other, and to ensure that the drive capability of the device is the same whether it is in

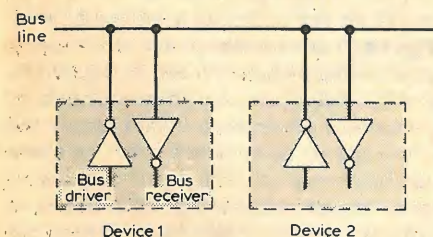
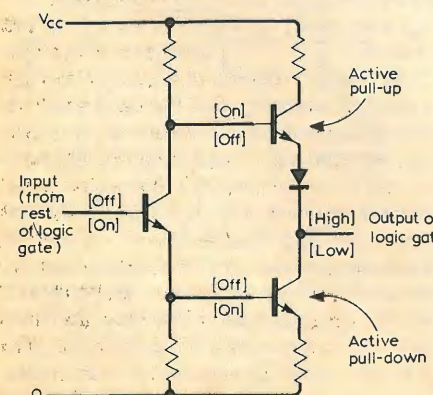


Fig. 5. Two devices connected to a bus line.



a high or a low state. The figure is annotated to show the input and output levels and the states of the transistors, when the output is high (upper annotations) and low (lower annotations).

Fig. 7 sketches the totem-poles of two gates whose outputs are connected together by a bus line, annotated with the states that occur when device 1 is putting a high level on to the bus and device 2 is driving it low. The result will be a large current flow, as shown by the arrow. Its value can be calculated as  $V_{cc}$  minus twice the collector-emitter drop minus one diode drop, divided by the resistor value — about 30mA with normal component values. The bus line will probably end up at a low logic level, but this will depend on how it is loaded by other devices. Both output stages will get hot and may fail — with 30mA between a 5 volt  $V_{cc}$  and ground, 150mW must be dissipated. T.t.l. is just not designed to operate in this way.

**Open-collector gates.** One solution to the problem is to use "open-collector" gates to drive the bus. These are t.t.l. gates which do not have active pull-up. Instead, the output is taken from the collector of the pull-down transistor, which is otherwise left open, as shown in Fig. 8.

Somewhere on the bus line, a bus termination resistor is supplied to pull the line up to  $V_{cc}$ . Then, if any of the open-collector gates which drive the bus has its output transistor ON, this active pull-down will overcome the pull-up resistor to bring the line to a low state. Only if all the gates which drive the bus have their output transistor OFF will the line float up to a high level. Thus a device can (logically) detach itself from the bus by turning its output transistor OFF, when it will not interfere with a device which wishes to drive the bus.

The table in Fig. 9 shows that if A is high, B's level is transferred faithfully to the bus, and vice versa. The table is that of the logical AND function, if the levels are interpreted in positive logic (LOW = logic 0, HIGH = logic 1), and the configuration is often called a "wired-AND" since the AND operation has been accomplished simply by a piece of wire. (The term "wired-OR" refers to exactly the same phenomenon, with a "negative logic" interpretation.)

**Tri-state gates.** There are two drawbacks to the use of open-collector gates to drive buses, both of which stem from the lack of active pull-up. Firstly, they are slow; and secondly, they are susceptible to noise. A simple alternative, which has only appeared recently with the advent of c.m.o.s. (complementary metal-oxide semiconductor) logic, is the tri-state gate. The c.m.o.s. logic family is often used for microprocessors, rather than t.t.l., and it turns out to be natural in c.m.o.s. to give gates a third, "disabled", state in which they do not affect the bus. The idea of a tri-state gate

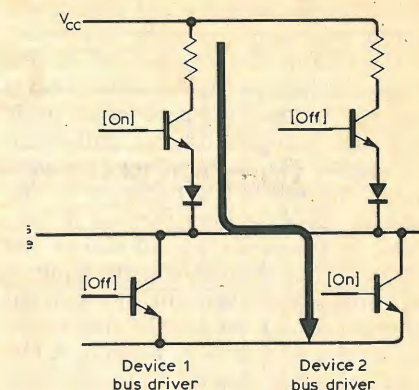


Fig. 6. T.t.l. "totem-poles" fighting for the bus.

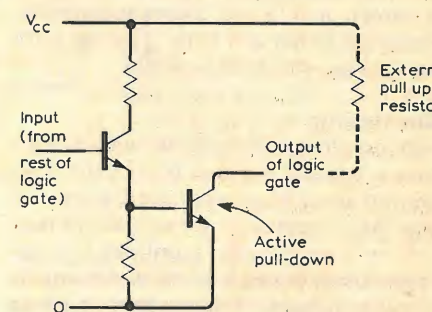
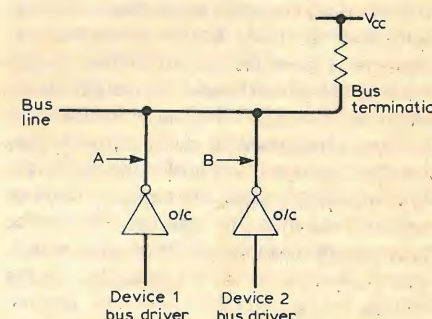


Fig. 8. Open-collector output stage.



A	B	bus line
low	low	low
low	high	low
high	low	low
high	high	high

Fig. 9. Open-collector gates driving bus.

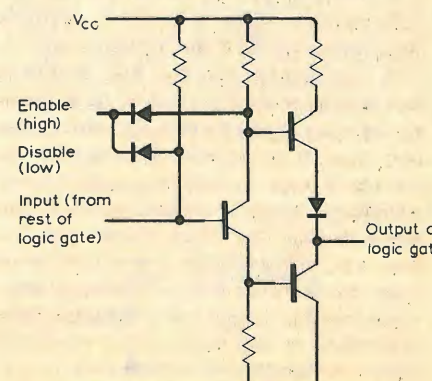


Fig. 10. Tri-state output stage.



proved so useful that a t.t.l. implementation was subsequently invented, which retains the totem-pole output stage of ordinary t.t.l., but adds an extra "disable" input to the gate which turns both the pull-up and the pull-down transistors off, over-riding the logic input. The result is effectively to disconnect the gate from the bus if "disable" is asserted. Fig. 10 shows the circuit. If the enable/disable input is high, the diodes prevent any current leakage through the enable/disable line and so the gate acts as normal. A low level on enable/disable, however, holds the base of the two transistors down and keeps them off. This automatically ensures that the third transistor is also off.

The symbol for several tri-state gates is shown in Fig. 11. Buses are almost always driven by this type of gate nowadays, wherever possible.

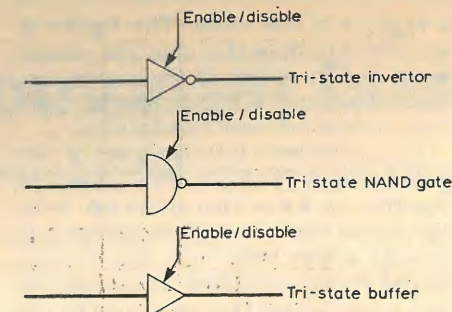


Fig. 11. Logic symbols for tri-state gates.

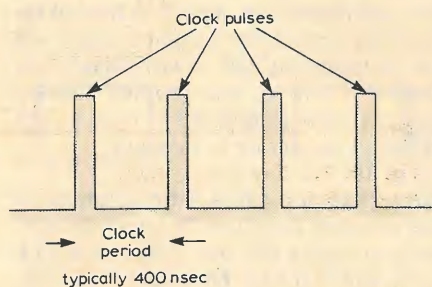


Fig. 12. Clock signal for synchronous buses.

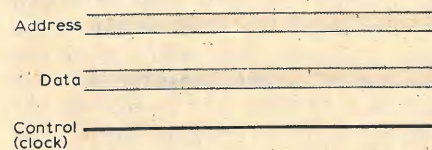


Fig. 13. Synchronous bus.

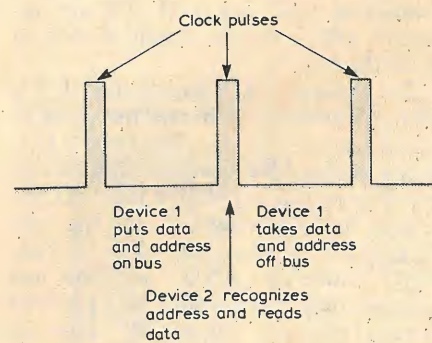


Fig. 14. Transmission on a synchronous bus.

pose that it will be able to read the data from the bus at the same pulse, leading to

2. — a device takes one clock pulse to read from the bus. The sequence of events when device 1 sends data to device 2 is shown in Fig. 14.

A circuit to read from the bus will need an "address decoder" to detect when the address lines are set to select the device, and a register to receive the value that is read from the data lines at the next clock pulse. Such a circuit is

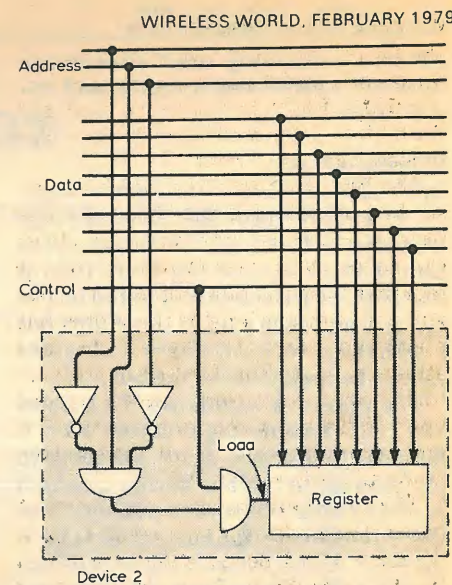


Fig. 15. Reading from a synchronous bus.

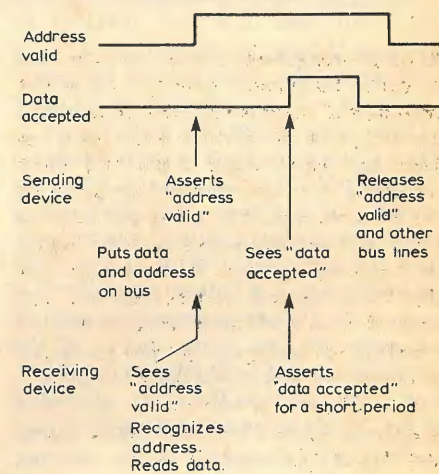


Fig. 16. Transmission on an asynchronous bus.

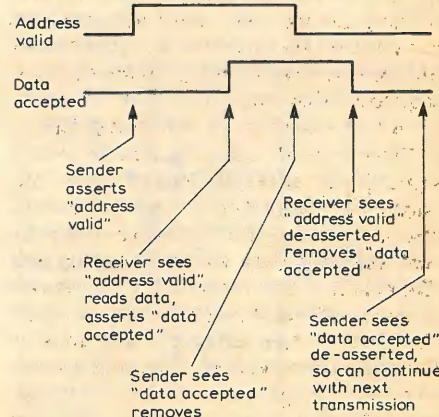


Fig. 17. Fully interlocked transmission.

shown in Fig. 15. There are three address lines, and the address of the device is <010>. Eight data lines are used, and the 8-bit register within the device holds the last data word read.

The bus structure that has been described is called a synchronous bus because each device on it is driven from the same clock: the two conventions together form the bus protocol. The bus will only work as long as the protocol is observed correctly by all devices attached to it. The idea of a protocol (dictionary definition: "terms agreed upon as the basis of a formal treaty") is an important one in all interactive communications systems.

The disadvantage of a synchronous bus is that all devices on it must work at the same speed, because they are driven by the same clock. Thus, the bus must go at the speed of the slowest device on it. For example, if a store module with an access time of 500ns (fast store) is attached to the bus, and another with an access time of 2μs (slow store), the bus will need 2μs clock and will not gain any advantage from having the fast store.

**Asynchronous buses.** The two problems posed by bus synchronisation were: the presence of spurious addresses during a transition of the address lines, and the need to know when data had been accepted by a device. The first problem can be overcome by having an "address valid" control line which is set to logic 1 by the sending device when it is sure that the address lines have settled. The second needs a "data accepted" line which is set to logic 1 by the receiving device as soon as it has read the data.

This allows for devices of different speeds to operate sensibly on the same bus. When the sending device has put data and address on to the bus, and after it has waited for a short time (typically 200ns) to ensure that the lines have settled, it asserts "address valid". This is the signal for all devices on the bus to examine the address lines to see if the data is for them. If it is, they read the data immediately, and then assert "data accepted", the time taken for this operation depending on the speed of the device. In either case, as soon as the sending device sees "data accepted" it can release the bus lines and proceed to its next operation. Fig. 16 shows the sequence of events.

This protocol is called "handshake" or "interlocked" transmission. It uses a signal from the sender to assert that the address and data lines have been set up, and a response from the receiver to indicate that data has been accepted. The result — transmission at the maximum speed possible for the two devices. The price — two extra control lines in the bus, and the risk of "hanging up" for ever if things go wrong — if, for example, the receiving device did not work properly.

In practice, the possibility of hanging up for ever can be rather embarrassing.

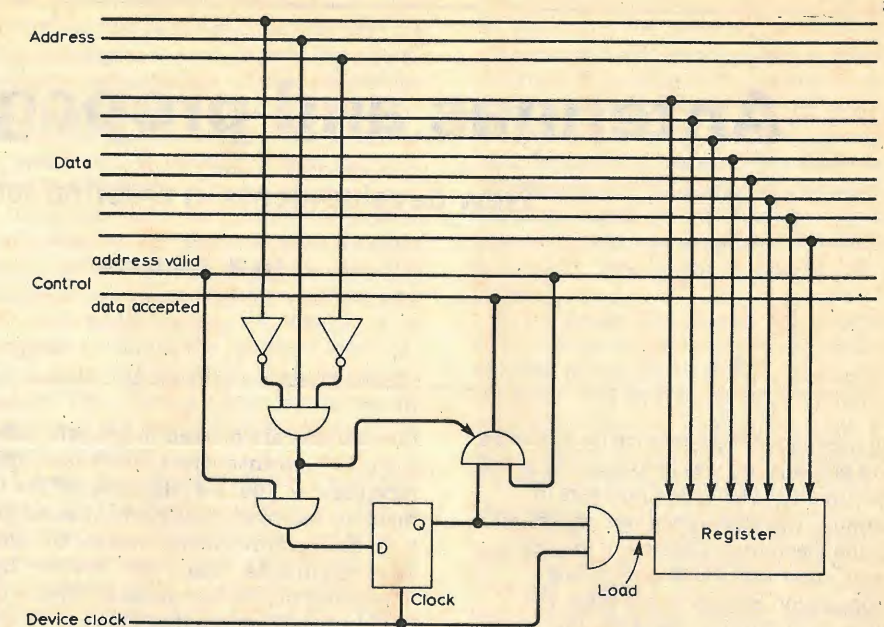


Fig. 18. Reading from a fully interlocked bus.

It only needs someone to run a program which attempts to communicate with a non-existent device, or to unplug a module from the bus, to bring the whole system to a halt. Sometimes this is considered so dangerous that the sender implements a time-out, where if the receiver has not responded after a certain maximum time (say 10μs) the sender takes control again, probably to report a fault condition. The time-out idea is important in all systems which "fail soft", that is, do not collapse completely when something goes wrong. People, of course, implement time-outs automatically when, say, no one answers one's knock on the door.

The only difficulty with the protocol described is that the time for which the receiver asserts "data accepted" has not been specified. If it asserts it for a very short time, it may be missed by the sender. If it is asserted for too long, the sender may be doing its next transmission and falsely conclude that this too has been accepted. A solution is for the receiver to assert "data accepted" until it sees that "address valid" is no longer asserted. The sender should then wait for the receiver to de-assert "data accepted" before using the bus again. This protocol is called "fully interlocked" transmission, and is shown in Fig. 17.

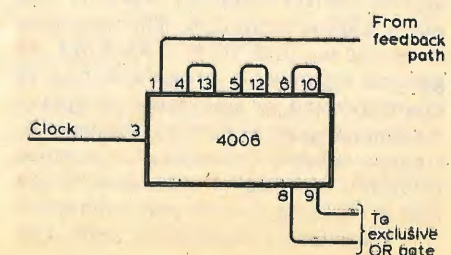
Fig. 18 shows a circuit to read from mission. Note that the use of an asynchronous bus does not necessarily imply that the devices attached to it employ asynchronous logic: normally, as in the figure, the devices have their own internal clocks. When the receiver sees its address on the bus with "address valid" asserted, it awaits the next clock pulse, and uses it to set a flip-flop and load the bus data into the register. Since this will happen virtually instantaneously, it simultaneously gates "address valid" back on to the "data

accepted" during transients.) Since this will happen virtually instantaneously, it simultaneously gates "address valid" back on to the "data accepted" bus line. If the load operation were to take longer, the device could be designed to wait until its completion before asserting "data accepted" to prevent the sender from altering the state of the bus. Note the use of a tri-state gate to drive the bus; the driver is enabled whenever the address lines are recognised. (This may result in driving the bus line during address transients, but no harm will come because the sender will not be looking at "data accepted" during transients. □

To be continued

## The Chatterbox — 4006

Owing to an oversight, we did not publish the pin connections to the 4006 shift register used in the noise generator section of the Chatterbox speech synthesizer (December 1978 issue, p. 41). Apologies to readers for this error. The required pin connections are shown below.





# Antennas and propagation — 1

New developments in antenna technology

by R. Ashmore

An international conference on antennas and propagation was held recently at the Institution of Electrical Engineers in London. The conference was organised by the Electronics Division of the IEE in association with the Antennas and Propagation Society of the IEEE, the Institute of Physics, the IERE, the Institute of Mathematics and its Applications, and the International Union of Radio Science. A large number of papers were presented; at least 97 on antennas and 39 on propagation. The following text is based on extracts from some of the papers on antennas.

THREE authors from the University of Oulu in Finland say that the characteristics of h.f. systems are analogous to industrial processes and in their paper<sup>1</sup> they describe a systems engineering approach to finding solutions for increasing the performance of radio communication systems.

In process control, unwanted effects have been combated by using steady-state optimization and real-time control methods, this being made feasible in the 1970s because of the dramatic advances made in microcomputers and the theoretical tools available. Some twenty scientists and engineers at the University of Oulu have been carrying out experimental work towards applying this approach to radio communication (mainly h.f.) systems since 1970.

The disadvantages of h.f. due to ionospheric propagation, large variations in path attenuation, channel imperfections, noise and interference, limit the usefulness of this frequency band as an economical low-capacity communication medium. Modern solutions for increasing the performance of radio communication have included automatic frequency and gain control, propagation channel and station identification, the introduction of new modulation and signal processing methods, the use of feedback channels and/or central control stations and multichannel reception. The solutions have, according to the authors, in general followed a pragmatic course characterized by unrelated improvements emerging as new technology becomes available. This kind of unplanned evolution, although common, does not lead to optimum overall performance.

In the systems engineering approach,

the systems are broken down into sub-systems — transmitters, antennas, and receivers — and a systematic effort is made to improve their performance. In h.f. communications most of the improvements can be made by improving the performance of receiving and transmitting antennas.

The requirements of h.f. antennas cannot be met by any type of classical h.f. antenna which is designed for average conditions. The antennas required must possess intelligence, so that they may follow the changes of propagation path and the signal and noise environment. For this reason, a new h.f. array configuration, suitable for controllable and adaptive operation, was developed during the university's project. The configuration comprises a planar array of horizontally polarized directive yagi- or log-periodic-type arrays supported by a single rotatable guyed tower. This configuration gives a low-angle main-lobe suitable for medium and long distances (over 1000km) together with high gain and simple beam steering.

Based on the results of tests on the experimental beam, the authors conclude that intelligent equipment needed for large-scale use could be manufactured very economically in the 1980's using v.l.s.i. circuits. Low power analogue and digital circuits would even make mobile and portable operation possible. The main difficulties, they say, lay in the selection and implementation of adaptive methods and algorithms suitable for the h.f. environment and therefore this formed a major part of their project.

In the future the adaptive solutions for the lowest levels could be combined with overall real-time control of entire h.f. links and the operation of these h.f. links could in turn be actively coordinated in real-time by regional and

world-wide centres. H.f. communications, say the authors, could be developed into a complex, integrated, adaptive system where individual links would be able to adapt to changes in their environment.

## Circular array with 360° null steering

Radio communications to and from moving platforms such as aircraft, ships and land vehicles, generally require omnidirectional coverage. However, this makes the system particularly vulnerable to directional interference. One way round this problem is to design an omnidirectional antenna having one or more directional nulls which can be electronically steered to minimize any such interference. A paper<sup>2</sup> by D. E. N. Davies and M. S. A. S. Rizk from University College London, describes such an antenna based on a four-element circular array, having a very small radius of  $0.04\lambda$ . This small radius allows the complete array to be housed in a thin cylinder or whip structure. The antenna pattern permits omnidirectional coverage with a null which may be electronically steered through 360° of azimuth by means of a single phase shifter. Fig. 1 shows the experimental pattern of the antenna.

To confirm some theoretical predictions, described in the paper, a four element null-steering array was constructed for the frequency 153MHz. Since the antenna has a very small radius there is a reduction of antenna gain and some degradation of bandwidth performance, compared with larger arrays. Also, the form of the directional pattern changes with null pointing angle in azimuth. There is also a deterioration of null depth when the antenna is operated away from its design frequency of 153MHz. The results

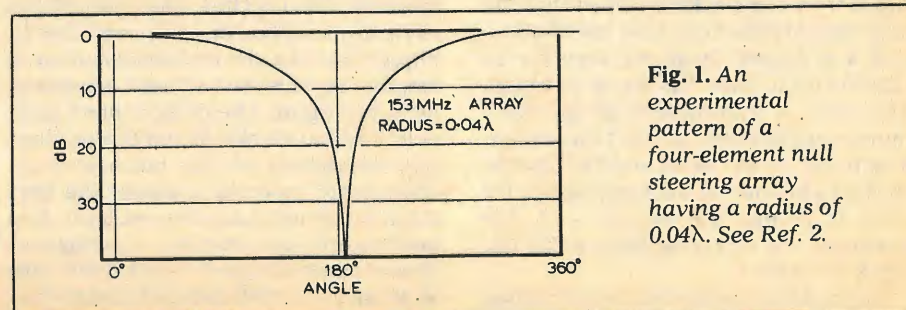


Fig. 1. An experimental pattern of a four-element null steering array having a radius of  $0.04\lambda$ . See Ref. 2.

indicated that the effective bandwidth limitation of the array for a fixed 50Ω termination of the mode was in the region of  $\pm 10\%$ .

It was found that one reason for the bandwidth limitation of the experimental results was a resonant effect due to the centre mast of the structure which supported the array. This was because the mast was designed to work as a single dipole for other purposes.

Although the null steering antenna system may operate in either a transmit or receive mode, the principal interest of rejecting interference is obviously confined to receiving only. This type of array can also operate as a directionally sensitive receiving antenna by connecting the output of the two modes — zero or null — to a phase sensitive detector. The output of the p.s.d. can then be calibrated in angle of arrival of any received wave. In addition, this means that this signal may be used directly as the information source needed to steer a null on to any required signal where automatic or adaptive null steering performance is wanted.

The paper shows that a simple four-element null-steering antenna intended for wideband operation may be modified to be used as a highly compact, small radius, null steering antenna. Null depths on the experimental antenna were in the region of  $-40\text{dB}$ , at the centre frequency of 153MHz, and over the  $\pm 10\%$  bandwidth the depth was not worse than  $-25\text{dB}$ . The configuration gives an antenna gain 5.3dB worse than an ordinary dipole.

This work was supported by the Ministry of Defence Procurement Executive.

## Satellite communications

The capacity of a communications link can be doubled by transmitting independent signals within the same frequency band in orthogonal polarizations. This technique, which is being used more and more in satellite communications (e.g. *Wireless World*, December 1978 issue, p.63), is the basis of work described in a paper<sup>3</sup> by D. H. Brandwood from the Plessey Company at Roke Manor. This work was carried out for the International Telecommunication Satellite Organisation.

When using orthogonal polarizations there is always a certain amount of cross-coupling between the channels caused by imperfect antenna alignment or when the signals pass through heavy rain or ice crystal layers. One method of reducing the cross-polarization is to modify the polarization within a waveguide at the receiver, before the orthomode transducer, a device which separates the signals in the two polarizations. Alternatively, the depolarized signals can be received unmodified at the transducer and a cancellation technique removes interference from a receiver channel by adding in an equal amplitude, antiphase version of the interference waveform\*.

In the orthogonal polarization case there are two receiver channels and the copolar signal in each channel is used to achieve cancellation of the crosspolar signal in the other channel. Since the crosspolarization varies with weather conditions it is necessary that the system be adaptive.

One technique for obtaining control information for the adaptive double cancellation system requires that the satellite transmits two pilot tones, one in each polarization. However, it is possible to obtain the required information from the satellite signals themselves. The Plessey paper discusses an outline of the analysis of the pilot independent system.

For the experimental work a receiver was constructed to operate with a microwave cancellation network. Laboratory tests were carried out on this receiver using a network to produce suitable cross-coupling levels. Tests were then carried out at realistic signal-to-noise ratios using travelling-wave tubes to produce simulated receiver noise.

The associated theoretical work showed that in principle crosspolarization discrimination could be improved to the required level by a cross-coupled cancellation technique which does not require transmission of pilot tones by the satellite. In conclusion, Mr Brandwood says that this cancellation technique is directly applicable to any two-channel system with cross-coupling, however this cross-coupling may be caused. The receiver design could also be modified for use with t.d.m.a. signals and cancellation could be carried out in other frequency bands (the experiments used frequencies between 2.5 and 75MHz) or at i.f. if cancellation were not required over the whole r.f. band.

## H.f. and the amateur service

Les Moxon, G6XN of the Radio Society of Great Britain, says in his paper<sup>4</sup> that there is a considerable number of differences between amateur and other communication services operating in the h.f. band of the frequency spectrum. A number of engineering problems arise because of these differences and the solutions to the problems are unique to the amateur field.

Firstly, in amateur communications, continuous service is not required. The amateur also has, in general, no freedom in the choice of location, but has to make the best use he can of a particular, usually small, area of ground subject to various technical and planning constraints. In most cases he is severely

\*This may be a clue to the technique used in the Plessey Groundsat repeater, described on page 71 of the November '78 issue. It is interesting to note that the new technique used on this repeater was discovered while designers were working on other projects. Both of these works were by engineers at Roke Manor.

restricted in regard to antenna size and height, and he is often affected by screening and local noise sources. Amateur operation is also restricted to a small number of harmonically-related frequency bands. In addition, the amateur is usually forced, by financial limitations, to use low power and simple antennas, and by the terms of his licence, to use output powers restricted to a mean level in the region of 100 to 150W.

In his paper, Les Moxon draws attention to some engineering developments inspired by the above differences, and to evidence that in solving the problem of very-long-distance communication, despite severe restrictions, amateurs are making use of the chordal hop mode of propagation (more about this later).

The effect of height on performance over long distance paths depends on the required angle of radiation, on which there is little data. Ideally the height must be sufficient to allow in-phase combination of the direct and reflected waves at the required angle, but in most cases it must suffice to erect the antennas as high as possible, typically 10 to 20 metres, compared with a desirable figure of at least 40 metres.

Despite this, daily antipodal communication is possible with typical amateur installations over most of the sunspot cycle. Much better results are of course obtainable when the ground slopes in the desired direction, and it has been shown that when ground slopes down steeply to the sea, or a flat plain, up to 12dB gain may be obtained from the two ground reflections. Using portable equipment to seek out slopes of this kind, regular communication with antipodes has been achieved with dipole antennas at heights of only 3 to 8 metres, s.s.b. transmission, and p.r.p.s of between 0.5 and 1.5W. This, says Mr Moxon, suggests the possibility of relatively low-cost commercial installations for ranges upwards of 3000km, by using cliff or mountainside locations.

To be continued

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# NEWS OF THE MONTH

## Proposal to make Carfax and ARI compatible

A proposal by Robert Bosch Ltd, concerning the possibility of making Carfax and ARI compatible, has been presented to the BBC's research division.

Readers who have been following the progress of Carfax and ARI in the UK will probably be left wondering whether Bosch has finally thrown in the towel, or whether, in an attempt to clutch at the last straw, they are taking the attitude that if you can't beat 'em, join 'em. For the benefit of those who have not been following the subject, radio equipment for the ARI (automotive road information) system (see p73, October '78 issue) is manufactured by Blaupunkt Radio, who are owned by Robert Bosch, for the Continental road traffic information service, and Carfax, which is the BBC's answer to road traffic information, is presently being developed and tested by the BBC for the UK. The Carfax system is described in the Jan '78 issue, p28).

Bosch suggests that the UK should adopt traffic information services in three stages. Firstly, the company believes that there is compelling evidence to justify the introduction and installation of ARI immediately in the UK. This introduction, which would be stage one, would cost only £25,000 to install "overnight" in Britain. The next stage is best described as it was put to *Wireless World* by Andrew Rodger of Bosch, as follows: "As stage two, and if we can afford the cost of installing the necessary 80 to 120 transmitters, we consider Carfax could be installed on f.m. between 100 and 104MHz. It could then be internationally compatible with ARI." It is worth noting here that the additional proviso in the first sentence is one of the main reasons why the pro-ARI are so anti-Carfax.

Stage three of Bosch's plan would then involve the installation of a system like ALI (the Driver Guidance and Information Service which is now being installed and tested in the Ruhr) which would provide drivers with individual guidance and information on a dashboard display via induction loops built into the road.

The proposal, which if implemented will enable the car radio equipment of several million motorists from the Continent to work satisfactorily in the UK, and the Carfax receivers from the UK to work satisfactorily on the Continent, was first put to the International Broadcasting Convention in September by G. Bolle, director of research and development at Blaupunkt Werke GmbH, West Germany. He agreed that there was no doubt that the principle of the Carfax system, to install a separate network of stations, was an ideal way to build up a network for the transmission of traffic information. In fact, he said, such a plan to create a network of independent stations had existed in West Germany about ten years ago but this had to be delayed because the constitution of the FRG does not permit anyone other than the ten German radio companies to transmit broadcast-type material and because there were no available frequencies for a new

transmitter network. The ARI system was the result of engineers making use of the existing transmitter network.

Mr Bolle continued by saying that if it is possible in the UK to run such an information network without the aid of the radio station, and if there was sufficient money available to run a completely independent network, then Carfax would be the ideal system and ARI, in fact, only a "mini-Carfax". In order to avoid the problems of possible incompatibility between the two systems, the Carfax system would have to be run in the v.h.f. band at a frequency between 100 and 104MHz, and Mr Bolle believed that a special frequency in this range could certainly be found in the UK.

As far as Bosch knew, there was not much chance that the WARC '79 Conference would allocate a frequency for Carfax in the medium frequency range because these frequencies were reserved exclusively for stations for large area service. Using a frequency in the v.h.f. band would considerably reduce co-channel interference, and one of the excellent features of the Carfax system, the capture effect, would at least be maintained if not improved.

One possible problem is that the ARI system depends on the fact that the carrier of the traffic information station is permanent and to overcome this the Carfax transmitters would also have to run constantly. According to Bosch, the power consumption for the 90 Carfax stations in the UK, about 5000kW/day, which is equivalent to approximately 300 medium sized cars, is reasonable. □

### Computers and privacy

The subject of the effect of computer systems on the privacy of individuals is continuing to attract a large amount of public interest. When a special committee was set up recently by the National Electronics Council to investigate the social and economic impact of using computers, one of the first themes to be taken up concerned the possible effect of computers and computer systems on the privacy of individuals. At an IEE colloquium, held in November and entitled "The interaction between computer technology and society" some of the eight papers presented discussed the social implications of computers, and one paper, by Sir Norman Lindop, director of Hatfield Polytechnic (see further), dealt specifically with the privacy aspect.

In December, the Government released "The Data Bank report", a report of the Committee on Data Protection, headed by Lindop. This disclosed that Sweden has refused to send data to the UK because of the lack of data-protection control in the latter country. Now, another White Paper has been promised for the spring of 1979. A bibliography has also been compiled by a member of the IEE's library staff to inform researchers of a wide range of papers dealing with privacy and computers. This brings

together about 250 citations covering a period from January 1969 to July 1978.

An article, "Computers and Individual Privacy" by F. J. M. Laver, in the NEC's journal, *National Electronics Review* (Nov./Dec. 1978) recalls some of the events leading up to the present situation. The first time the subject attracted a vast amount of public interest was in 1966 when it was proposed that a central installation should be established to bring together information held separately on the computer tapes of various US government agencies. Then, in Britain, after two abortive attempts at legislation by private members, a government committee under Sir Kenneth Younger, produced a White Paper (5012) in July 1972. After a long pause and following some public pressure, the Home Office responded by publishing two White Papers (6353 and 6354) in December 1975. They also appointed Sir Norman Lindop as chairman of the then newly formed Data Protection Committee (DPC) to examine the need for, and the nature of, legislative or administrative measures to protect the privacy of the individual or individuals. It was in August 1976 that the DPC invited the views of representative bodies, including the NEC.

Mr Laver's article is intended to highlight a few of the points that occupied the NEC's Social and Economic Impact Committee during 1976 and 1977. The Council's interest in privacy derives from its belief that there is risk that, over-reacting in its anxiety to allay the fears expressed, the Government might be persuaded to construct a legal apparatus of control which would hamper the future development of computers in the UK. If this was to happen, says the article, heavy costs and complex procedures would be imposed on many computer users in order to guard against carelessness or malpractice by an irresponsible few, and it could restrict the growth of social service applications, in particular, and cut off large benefits which the intelligent use of computers alone could bring. Economic and commercial advantages could also be delayed or diminished and as a result electronic technology would be prevented from contributing as fully as it could to the UK's economic and social health. It would particularly and adversely affect the UK electronics industry's competitiveness in international markets because this industry would be denied opportunities of developing and demonstrating its competence in important fields.

The NEC article suggests that, in the interests of fair competition, as well as to protect the individual, it is highly desirable that a common, or at least a compatible, control regime should be established within the EEC, and preferably also in Japan and the USA. As a minimum it suggests the use of some form of international convention to discourage the exploitation of differences between national regulations, and to prevent easy-going countries from seeking to attract data banking business.

## First low cost speech synthesizer in UK

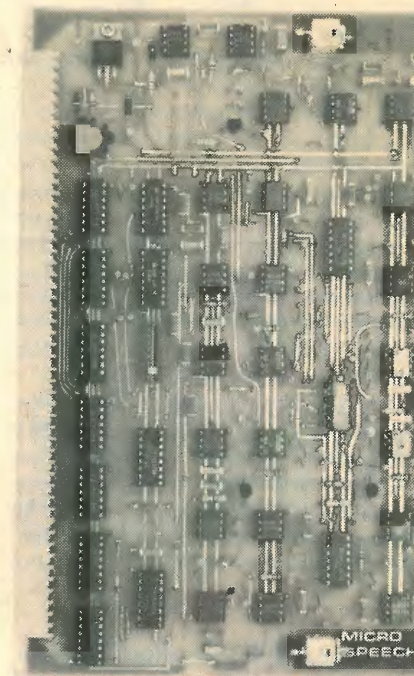
Until recently, electronic speech synthesizers have been very limited or very expensive and this has restricted their use. The recent speak and spell device from Texas, although not a true synthesizer, does provide high quality speech very cheaply, and has opened the door to more versatile pre-programmed units.

Microspeech from Costronics, however, is a true speech synthesizer, priced at around £300, which has been designed to operate directly with the standard SS50 bus as used on the South West Technical Products 6800 microcomputer. The microspeech package contains a speech synthesizer with separate nasal and fricative branches, a digital noise source, and a voltage controlled oscillator. To complement this board, software is supplied on a floppy disc. This translator program occupies about 4K of memory and converts phonetic code into sets of data which control the synthesizer. This data produces nine control parameters which in turn determine pitch, amplitudes, and resonant frequencies. In operation, the characters from the phonetic text are compared with a phoenem look-up table. If a match is not possible, a one character match is attempted while at the same time checks are made for the end of text character and for the

pitch control characters. If a pitch control character is encountered, internal updating takes place as necessary. When a phoenem is recognised, the parameter store for that sound is released. This store defines the starting and stopping values of eight vocal tract parameters together with the phoenem and type of interpolation to be used for the resonances and amplitudes. Each time a new frame of data is produced during the interpolation, it updates the speech synthesizer. When the phoenem has finished, a short interpolation is made between the end values of the current phoenem and the starting values of the next one. The whole process then repeats until the text has finished.

An important feature of Microspeech is its real-time operation. This means that there are no gaps in the speech and the device can be arranged to operate directly from a keyboard. Alternatively, speech can be stored, around 90s for each 1K of buffer space, without the need for a large memory to accommodate the control parameters.

Although the speech quality is by no means perfect, the system is theoretically capable of synthesizing any voice provided that the circuitry is controlled correctly. For this reason, most of the current development is taking place on the software.



## Video disc system now marketed in USA

The Philips and MCA optical videodisc system is now being marketed in the USA, according to an announcement made recently by the Magnavox Consumer Electronics Company, a subsidiary of North American Philips. This move, which actually began in December last year, is presently restricted to the Atlanta market, but distribution is to be extended in 1979 to other markets in the United States.

While this action is unlikely to immediately affect consumers in the UK, Mr Gerrit Jeelof, the Chairman and Managing Director of Philips Industries UK, said that Britain might be the second country to have this development and that they would be trying to get it on the market in the UK as early as possible in 1980.

This particular system has a videodisc player manufactured and marketed by North American Philips' Magnavox subsidiary under the trade name Magnavision, and pre-recorded video discs manufactured and marketed by MCA Incorporated under the trademark and tradename MCA Discovision.

It has been suggested by officials from Magnavox that the videodisc player will cost about £350 (\$695), and according to MCA a typical half-hour videodisc programme will be sold at a retail price of about £3 (\$5.95). Complete two-hour recent feature films are expected to retail at about £8 (\$15.95).

The Philips and MCA system attaches to a home television receiver and the videodiscs are grooveless and resemble a long-playing record. The optical system on the Magnavision player uses a tiny, low-power laser light beam which relays images and sound from the videodisc to the viewer's television screen. No needle or stylus ever touches the disc, so repeated handling or use will not

wear out or diminish the videodisc quality. The system's discrete stereo sound tracks can also be played through a home stereo amplifier to offer a sound which is much better than that obtainable from the television sound system.

One of the advantages of the Magnavision player is that the transmissions of picture and sound are equal to the best available from broadcast tv reception and superior to home videotape playback. In the standard play mode the system also allows the user to stop, slow or reverse the programme at any point on the disc. Related to this are the advantages of rapid access, fast motion, instant replay (no need to wind back tapes) and frame by frame readout. Another advantage, that of the disc's durability, has already been described.

### Largest order ever for UK electronics firm

Racal Electronics Group has received an export order worth £20 million, the largest ever in its 28 year history. The order has been placed by a Middle Eastern country for the supply of radio communications systems.

The Group is to supply a number of fully-equipped and custom-built transportable communications shelters and mobile workshops. The shelters house all the radio and terminal equipment needed for a complete, fixed communication station and can be re-deployed to any site within hours. This feature satisfies a particular demand for flexibility in a nationwide network.

Racal Communications Ltd will undertake all the systems engineering and the equipment itself will be supplied by several com-

## Audio writer award presented

The 1978 Audio Writer Award has been awarded to Barry Fox who writes under the pen name Adrian Hope. The award, a silver tuning fork mounted on a mahogany base and a cheque for £300, was presented by the international conductor, Charles Mackerras, in London during November.

Barry Fox, who started writing for publications about twenty years ago when he made a contribution to *Punch*, was an Oxford graduate in biology who became a patents consultant. It was not until a few years ago that writing became a major source of income for him and prompted him to become a consultant and writer on audio, video and patents. Now, he writes for many audio and video publications and he is a regular columnist in *New Scientist* and *She*. He is also the author of two books on inventions and audio.

panies in the Group, including Racal-Tacticom Ltd, Racal-Datcom Ltd, Racal-Dana Instruments Ltd, Racal Thermionic Ltd and Racal Communications Ltd. The order includes the largest ever logistic support contract so far achieved by Racal.

According to a Racal statement, early delivery is essential and the whole order is expected to be fulfilled by the end of 1979.

● An interim report from Racal Electronics Ltd shows that their unaudited pre-tax net profit for the half-year ended 30th September, 1978, amounted to £24,323,000. This compares with a 1977 figure of £19,398,000, an increase of 25.4%. Turnover during the half year was £99,894,000 and the previous half-year was £89,886,000, up 11.1%.

News continued on page 74



# High-performance preamplifier

Low-cost design with active gain control

by D. Self, M.A., M.Sc.

In his previous article Doug Self described a no-compromise preamplifier which was designed using high voltage transistors to give exceptional performance. This new design sacrifices very little of that performance and uses a small number of low-cost transistors to significantly reduce the cost. A novel active gain control makes best use of the dynamic range and removes the problem of volume control placement.

THIS PREAMPLIFIER offers a similar performance to that of the advanced preamplifier published previously<sup>1</sup>, but with a simpler design that reduces the parts count and hence cost. In normal use, the signal levels are kept around 50mV by exchanging the normal potentiometer volume control, which acts as an attenuation control, for an active gain control. Therefore, the signal receives only the amplification required for a given output and so makes best use of the amplifier's dynamic range.

The distortion performance is also improved because unwanted gain will be used to give higher negative feedback and thus greater linearity. The active gain control uses a shunt feedback circuit where the volume control varies the resistance of a feedback arm as shown in Fig. 1. The disc input stage has a relatively low gain of +20dB at 1kHz which allows a very high input overload margin. This is followed by a third-order high-pass filter which removes subsonic signals while they are still at a low level. Both bass boost and treble cut portions of the RIAA equalisation take place in the first stage. The gain control stage is positioned after the input switching and has a maximum voltage gain of +20dB. This is followed by a Baxandall tone control which has unity gain at 1kHz.

The use of an active gain control eliminates the problems associated with a normal volume control. If all of the gain is placed before the control, the supply voltage limits the overload margin. If some gain occurs after the volume control, then the signal-to-noise ratio is degraded because noise generated in the later stages does not undergo attenuation. The use of two controls, one early and one late in the signal chain, is one method of avoiding this compromise<sup>1</sup> but a true gain control

## Preamplifier specification

**Input sensitivity**  
for 500mV output  
Disc 5mV at 47k $\Omega$   
Line 1 100mV at 20k $\Omega$   
Line 2 100mV at 20k $\Omega$   
Line 3 500mV at 100k $\Omega$

**Disc input overload level**  
1.1V r.m.s. at 1kHz  
3.8V r.m.s. at 10kHz

**Outputs**  
Main output  
500mV r.m.s. at 100 $\Omega$  source impedance  
Tape output  
50mV r.m.s. at 1k $\Omega$  source impedance  
Maximum possible level from main output  
8.5V r.m.s.

**Frequency response**  
Disc input (RIAA equalisation)  
 $\pm 1.0$ dB 20Hz to 20kHz  
 $\pm 0.5$ dB 100Hz to 20kHz  
Line inputs (flat)  
+0, to -0.5dB 20Hz to 20kHz

**Total harmonic distortion**  
From disc input to main output, at 1kHz with the gain control set to  $\times 6$   
less than .008% at 8V r.m.s.  
less than .005% at 5V r.m.s.  
Because the output signal level will normally be around 500mV the t.h.d. level will be much lower.

**Noise**  
Disc input better than 68dB below 5mV r.m.s.  
Line inputs below -75dBm at full gain  
Residual below -90dBm at zero gain

## Tone controls

Bass  $\pm 14$ dB at 50Hz  
Treble  $\pm 10$ dB at 10kHz

**Power consumption**  
Approx 78mA each channel from a +38V supply.

is considered to be a more elegant solution.

Because a low-cost, single unregulated power supply is used with first-order RC smoothing to reduce ripple, all sections of the preamplifier are designed with high ripple-rejection performance.

## Disc input stage

The most difficult stage to design in a preamplifier is the disc input, and the problems are compounded if, as in this case, the gain of the stage is low to allow a high overload margin. A low voltage gain at 1kHz means that the feedback network which defines the gain and RIAA equalisation will have a relatively low impedance, and thus appear as a heavy load to be driven by the disc amplifier. This situation becomes worse at higher frequencies when the reactance of the equalisation components falls. Therefore, as a large voltage swing at the output is desirable, a large amount of current must be able to flow into and out of the feedback network at high frequencies. A second, and related problem, is that if the gain at 1kHz is low, the gain at 20kHz must be 19.3dB lower due to the RIAA equalisation, which makes it close to unity. Therefore, it becomes more difficult to set the top end of the RIAA curve accurately. For this reason an extra low-pass section, with a -3dB frequency of about 22kHz, is added after the disc amplifier to ensure that the high-frequency gain continues to fall at a steady rate. It should be noted that if the correct turn-over frequency is chosen for the final low-pass network, the RIAA amplitude and phase curves are obtained exactly.

Another consequence of the fall in closed-loop gain at high frequencies is that the compensation for Nyquist stability is more difficult, and in this design it was necessary to add a conventional RC step-network to the normal dominant-pole compensation. The dominant-pole capacitor is kept as small as possible to preserve the slew-rate capability of the stage.

The basic disc input stage is shown in Fig. 2. In this series-feedback configuration almost all of the voltage gain is provided by the second transistor, which has a bootstrapped collector load for high open-loop gain and linearity. The final transistor is an emitter-follower for unity-gain voltage buffering. This configuration allows the use of a p-n-p input transistor for optimum noise performance, but it also means that the collector current must flow through the feedback resistance  $R_F$ . This places another constraint on the design of the feedback network because an excessive

voltage drop must be avoided.

As the disc input amplifier must be capable of sourcing or sinking large peak values of current to drive the capacitive feedback loop at high frequencies, the conventional emitter-following output circuit in Fig. 2 is not suitable because the sink current causes a voltage drop in  $R_E$ . Lowering the value of  $R_E$  reduces the effect, but this is a poor solution as it leads to a high quiescent power dissipation. Replacing  $R_E$  with a constant-current source is more effective because the maximum sink current becomes equal to the standing current of the stage. However, this would still limit the output of the disc stage at high audio frequencies due to an inability to sink sufficient current. For this reason, the push-pull class A configuration in Fig. 3 was chosen. The bottom transistor is a current-source which is modulated in anti-phase to the top emitter-follower, via the current-sensing resistor  $R_A$  and a capacitor. This can also be considered as a negative-feedback loop that attempts to keep the

current in  $R_A$  constant. However, the open-loop gain is only unity and so with 100% negative feedback the current variations in the top transistor are reduced to one half by the capacitor. Due to the anti-phase drive of the lower transistor, this stage can sink a peak current of twice the standing current, and therefore give twice the output swing at high frequencies.

A practical circuit of the disc input amplifier and its associated subsonic filter is shown in Fig. 4. All of the d.c. bias voltages are provided by the potential divider  $R_2$ ,  $R_3$ ,  $R_4$ ,  $D_1$  and  $D_2$ . This chain is heavily decoupled by  $C_2$  to prevent supply-rail ripple entering this sensitive part of the circuit. Note that  $Tr_3$  and  $Tr_5$  are isolated from the bias voltage by  $R_{14}$  and  $R_{22}$  to simplify any fault-finding.

The RIAA equalisation is provided by  $R_{10}$ ,  $R_{11}$ ,  $C_6$  and  $C_7$  in the feedback loop, and  $R_7$ ,  $C_4$  forms a step network that aids h.f. stability. Resistor  $R_{17}$  and  $C_{11}$  make up the low-pass section that corrects the top octave of the RIAA curve. The

subsonic filter is a 3-pole Butterworth type with an ultimate slope of 18dB/octave. Although the frequency response shows a loss of only 1.5dB at 20Hz, the attenuation is increased to more than 14dB at 10Hz. The unity-voltage gain element of the filter is formed by  $Tr_5$  and  $Tr_6$  arranged as an emitter-follower with a current-source load. This configuration was chosen for its excellent linearity. An output of about 50mV is available for tape recording although the exact voltage will depend on the cartridge sensitivity. Resistor  $R_{24}$  prevents damage to  $Tr_6$  if the tape output is shorted to earth, and resistor  $R_{25}$  maintains the output of the disc stage at 0V d.c., and also prevents switching clicks.

The total harmonic distortion from input to tape output at 6V r.m.s. is below 0.004% from 1kHz to 10kHz but because the anticipated signal level here from most cartridges is about 50mV r.m.s. the distortion during use will be even lower. The disc input will accept more than 1V r.m.s. at 1kHz, and about 3.8V r.m.s. at

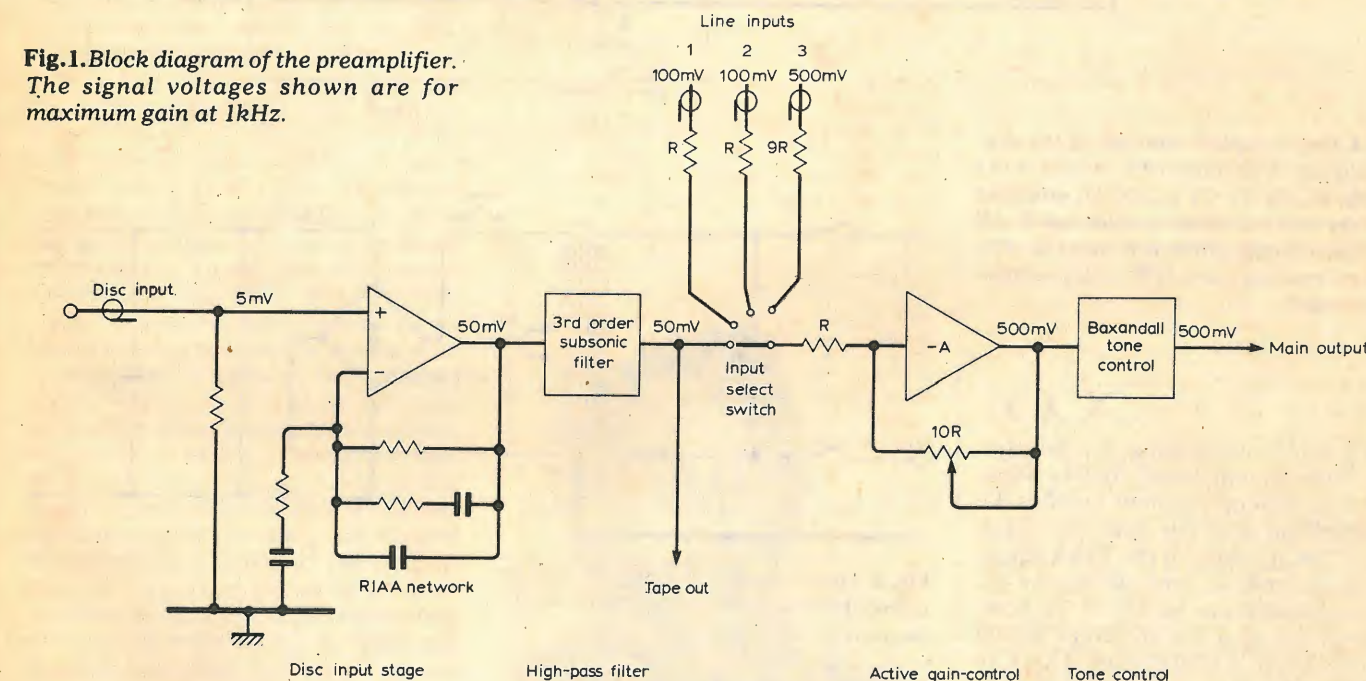


Fig. 2. Series-feedback disc input amplifier with an emitter-follower output.

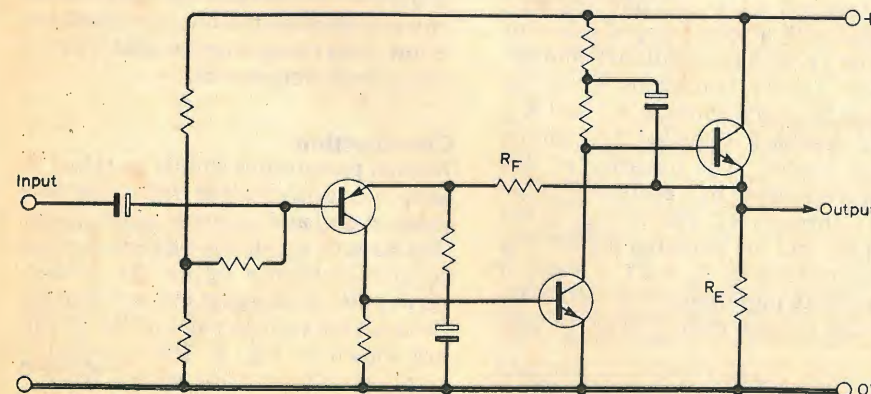
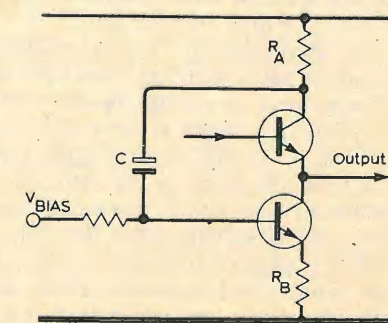
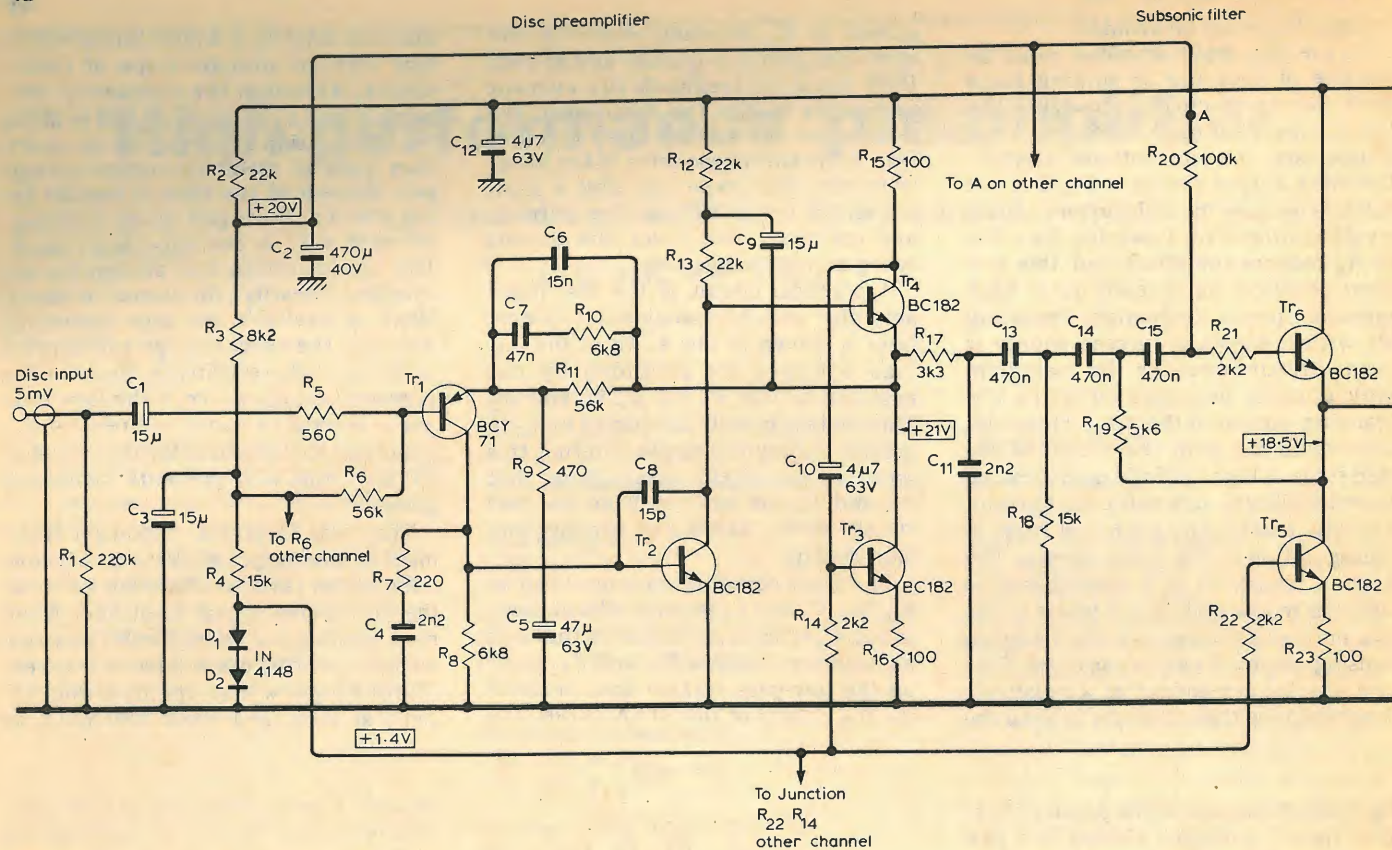


Fig. 3. Improved push-pull class A output stage.





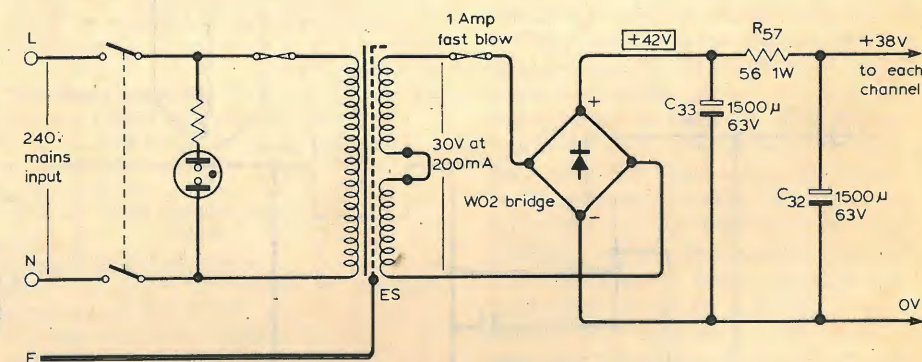


**Fig. 4.** One complete channel of the preamplifier. Components in the bias supply,  $R_2$ ,  $R_3$ ,  $R_4$ ,  $C_2$ ,  $C_3$ ,  $D_1$ ,  $D_2$  and also  $C_{12}$  are not repeated in channel 2. All electrolytic capacitors are rated at 40V and all resistors are  $\frac{1}{4}W$  unless otherwise stated.

10kHz before overloading. It is felt that the improvement these figures show over conventional methods justifies the complication of a low-gain disc input stage. The accuracy of the RIAA equalisation depends on how closely the RC time-constants can be set. If 5% components are used the deviation should be less than  $\pm 0.5dB$  from 1kHz to 15kHz, and within  $\pm 1dB$  from 20Hz to 20kHz. The signal-to-noise ratio for a 5mV r.m.s. input at 1kHz is better than 68dB.

The remaining part of the preamplifier comprises an active gain-control and the tone-control stage. The input switching is simple and requires only one switch section per channel. Also, any line input of suitable sensitivity can be used as a tape monitor return.

The shunt-feedback configuration of the active gain control enables each line input to have its sensitivity defined by the value of a single series input resistor. The maximum voltage gain available from the stage is the ratio of the feedback resistance to the input resistance, and is +20dB when the volume control is at maximum resistance. This gain is only used in the disc mode. The most sensitive line input is rated at 100mV for a 500mV output and the least



**Fig. 5.** Power supply. If a higher voltage transformer is used,  $R_{57}$  should be increased accordingly.

sensitive input has unity gain. Any sensitivity between these two limits may be provided by using the appropriate series resistor value.

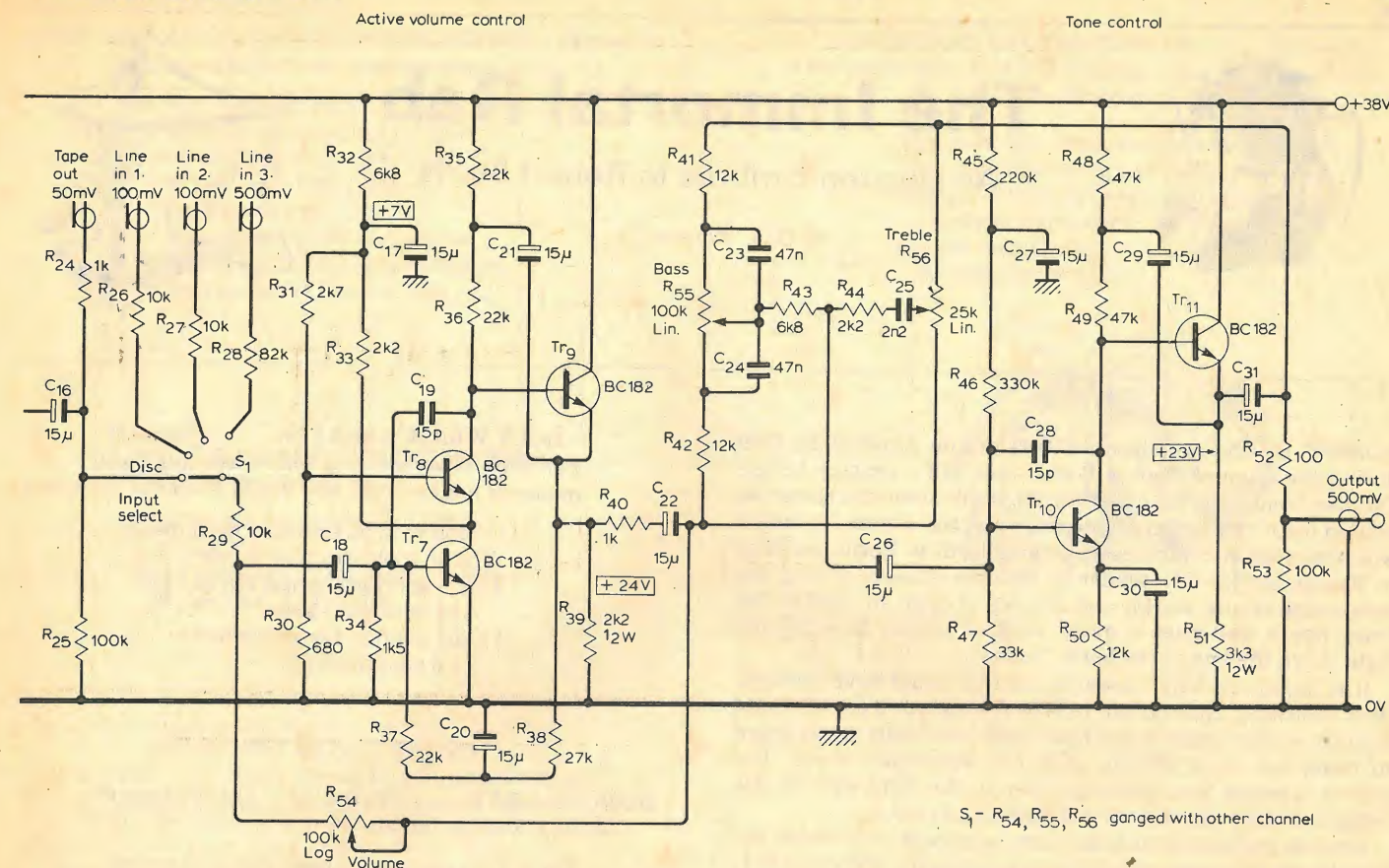
The gain control comprises  $Tr_7$  and  $Tr_8$  arranged as a cascode voltage amplifier with a bootstrapped collector load and  $Tr_9$  as a conventional emitter-follower. The d.c. conditions are set by negative feedback through  $R_{37}$  and  $R_{38}$ , and a.c. feedback is applied through the volume control. The linearity of this circuit is increased by a current injected into  $Tr_7$  through  $R_{33}$ . The voltage at the top of  $R_{33}$  and the potential divider  $R_{30}$ ,  $R_{31}$ , is smoothed by  $R_{32}$  and  $C_{17}$ . Resistor  $R_{40}$  prevents high-frequency instability when the volume control is set to zero gain.

The tone-control is a conventional Baxandall circuit, with  $Tr_{10}$  providing a

high voltage-gain by its bootstrapped collector load. Transistor  $Tr_{11}$  is another emitter-follower which buffers the high impedance at the collector of  $Tr_{10}$ . The output is taken through  $R_{52}$ , which protects the output against short-circuits. Because the output impedance is low, long cables may be used without loss of high frequencies.

#### Construction

Normal precautions should be taken to keep a.c. power away from the disc input stage, and to avoid earth loops. The leads to  $R_{54}$  should be kept short to prevent hum pick-up on the virtual-earth point of the gain control. Typical voltages for various parts of the circuit are shown in Fig. 4. These measurements should be made with a 20k $\Omega$ /V meter.



$S_1$  -  $R_{54}$ ,  $R_{55}$ ,  $R_{56}$  ganged with other channel

Several modifications can be made to the preamplifier to suit individual requirements. Firstly, the treble turnover frequency of the tone-control section can be increased from 2kHz as shown in Fig. 4, to 5kHz, for example, by reducing  $C_{25}$  to 1000pF. For variable turnover frequencies  $C_{25}$  can be made switchable. Some purists may feel that the provision of a tone-control is unnecessary, and even undesirable. In this case, the output should be taken from the junction of  $C_{22}$  and  $R_{54}$ , but  $R_{52}$  and  $R_{53}$  should be retained at the output. Because the current drawn by the preamplifier will now be less, it is advisable to raise the value of  $R_{57}$  to keep the supply rail at +38V.

In the circuit of Fig. 4, no balance control is included. This function was performed in the prototype by a dual-concentric volume control. If, however, a conventional balance network is required this can be added at the output of the preamplifier although the low output impedance will be sacrificed.

#### Printed circuit board

A glass fibre p.c.b. which accommodates two audio channels is available for £5.00 inclusive from M. R. Sagin at 23 Keyes Road, London N.W.2.

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#### The author

Douglas Self studied engineering at Cambridge and also obtained an M.Sc in experimental psychology at Sussex University, where he examined the perception of speech, and some related problems of psychoacoustics. In 1974 he entered the audio industry as a design engineer and has been involved in the development of a wide range of audio equipment. He has spent the last few years working for Electrosonic Ltd, where he designed several types of audio mixing desk. More recently, Doug has been engaged in the design of high-quality audio modules for the home construction market.

## BOOKS RECEIVED

**A First Course in Computer Technology**, by Michael Hartley and Martin Healey, looks at the subject of computing in broad outline, introducing the complete range of computing systems and techniques; each being treated in sufficient depth to afford a good grounding for more advanced work. The first chapter is a particularly interesting and useful survey of the development of computers from Babbage to integrated-circuit microcomputers and is immediately followed by a general section which provides a rationale of computers and computing; with a look at the organization of component parts in a typical system. Further chapters continue with more detailed exposition of computer systems and the development and handling of software, the final section being devoted to calculators and microprocessors. Many of the chapters contain problems for the interested reader to solve, and each chapter has a useful set of references, particularly that concerned with computing history. The book is published at £4.95 in paperback by McGraw-Hill Book Company (UK) Ltd, Shoppenhangers Road, Maidenhead, Berks.

**Video Yearbook 1979** is a mass of information, compiled in a way that allows relatively easy access to a large variety of equipment references. All types of equipment which could possibly be collected under the "video" umbrella are covered, together with a glossary of jargon and several lists of programme production companies, organizations and services. The second part of this fairly expensive, 545-page yearbook is devoted to indexes of addresses, abbreviations and commercial information. Angus Robertson edited it. Blandford Press, Link House, West Street, Poole, Dorset BH 15 1LL. Price £12.50, only in hard back.





# The Immortal Rab

An electronic tribute to Robert Burns

by D. T. Walker



ROBERT BURNS was born in Ayrshire on January 25, 1759 during the reign of George II and over half a century before Faraday conducted his experiments. In his colourful career he poured forth all manner of ode and lyric, the content of which was adequate in comradeship to give birth to Burns societies in Russia, sufficiently tender to become classics among the love songs of the world, and diverse enough to display the fears, hopes, and grief in every walk of society from royalty right down the line to the lowly "louse".

It is highly unlikely, however, that he could have foreseen their following application, in which excerpts from his more popular works (and others) have been minimally re-arranged to bring out their affinity with our electronic world. The author, a fellow Scot, presents them in this light with all due respect to the poet's immortal memory.

English readers with computer access should resist the temptation to process the following into it for decoding. ICL, IBM and the like would never forgive you. Any self-respecting library will provide a glossary of terms.

## ADDRESS TO THE DE'IL (Verse 3)

To those with transmitters and morse keys,

Great is thy power and e.r.p.  
Far kenn'd and noted for f.c.  
Makes spots and splurges on tv  
Stops housey housey  
But faith thou are baith lag and lame  
Your morse is lousy



## UP IN THE MORNING EARLY

and again

Up in the mornin's no for me  
This field day stunt's too early  
Who plugged my transformers into d.c.  
And tried my patience sairly

## THE EXCISEMAN (To the tune . . . .)

To the hams among us

CQ among the QRM  
Captured by the DX man  
His XYL cried "its only a 'G' "  
Why don't you sell your TX, man

You M.O's off tune, your aerial's doon  
Your feeder's mis-matched, DX man  
Why don't you sign off, you muckle great loon  
And try out a new CX, man

## HOLY WILLIE'S PRAYER (Verse 1)

For those who replace aviation warning lights  
on masts (and oil rigs) and a plea from our storemen

O thou who up the mast doth dwell  
Wha as it pleases best thyself  
Tak a' oor mast lamps, ten or twel'  
And cock your snoot  
I must confess I wish tae hell  
Ye'd sign them oot

## DUNCAN GRAY (Verse 1) (To the tune of . . . .)

To the television broadcasting engineers among us

The S.S.E's (or S.M.E's) cam here tae tweek  
Ha, ha, the tweekin' o't  
With S.B.A. he looked real chic,  
Ha ha, the tweekin' o't.  
With pulse and bar, and staircase too,  
We'll soon be on full power the noo,  
Ye've never sung a song more true,  
Ha, ha, the tweekin' o't.

## TO A MOUSE

Valves (No 1)

Wee sleekit 4CX250  
Oh what a panic's in thy breastie  
Don't go into self-osc. sae hasty  
Wi' parasitics  
I wad be laith tae hae tae 'neut' ye  
Wi' electrolytics

## ON A SUICIDE

In memory of valves (No 2)

Earthed up here lies an imp o' hell  
A parasitic 807  
Poor silly tetrode's damn'd himself  
And missed the gates o' heaven



## TO A HAGGIS (Verse 1)

In a moment of huff towards chief engineers  
with parallel operated transmitters

Great puddin' o' the chieftain race  
O' what a frown is on thy face  
Man, dinna glower  
Tae save ye frae complete disgrace  
There's aye half power

## FIRST BALLAD ON MR HERON'S ELECTION (Verse 1)

Dedicated to the P.O. engineers among us

Whom will you send to London town  
To Gresham Street, and a' that  
Wha in a' the country round  
Will get 'exec' and a' that  
For a' that and a' that  
We upward grind and a' that  
Where is the laird or belted knight  
Who'll make H.Q. and a' that

## A BOTTLE AND AN HONEST FRIEND

Valves (No 3) and hams

Here's a bottle, an honest friend  
A 4CX250  
Wha' ken's before its life may end  
My share o' CQ's nifty

## THE HENPECKED HUSBAND

Some sympathy for those among us whose DIY efforts  
are denied free rein on the living room table

Cursed be the ham, the poorest wretch in life  
The crouching vassal to the tyrant wife  
Who without her say has no transmission  
Who has no 807 but in her possession.  
And if a junk sales dare buy or sell  
Suffers domestic ragchew from his XYL  
Were such the wife had fallen to my part  
I'd break her spirit or I'd break her heart  
I'd charm her with the magic of a switch  
And some kV's, and electrocute the bitch!



## TO A LOUSE (2nd and last verse)

Speaking generally

Ye ugly creepin' blastit sneaker  
Detested, shunned by true fault seeker  
Ye electronic duff gen reeker  
O' simple brain  
Awa and plop that itchin' tweeker  
And pu' the chain

Oh wad some power the giftie gie us  
Tae see oorsel's as ithers see us  
It wad frae mony a blunder free us  
And foolish notion  
Technical incompetence will lea'e us  
Nae promotion

## ADDRESS TO THE TOOTHACHE (Verse 3)

For union members and productivity deal enthusiasts,  
and those who like their annual 'flu jag

When temper burns and goodwill freezes  
Well injected 'gin the sneezes  
Instead o' workin in their threeses  
Staff outwards drift  
Comes then the hell o' a' diseases  
The two man shift



## TAM O'SHANTER (Verses near the end)

For those of us with unattended stations to  
maintain by mobile maintenance teams

The station had barely made its mark,  
When in an instant all was dark,  
And scarce the teledac had rallied,  
When out the hellish legion sallied.

Like bees wi' low intent they thundered,  
In their Austin 1800,  
Along the roads, ower field and dyke,  
Assail'd the gremlin in their byke.

Once more translators wail and banter,  
Like Tam, playin' dirges on his 'shanter'.

## AND FINALLY . . .

Needing no introduction

Should auld acquaintance be forgot  
When gaun to ither station  
Should auld acquaintance be forgot  
When spread throughout the nation  
With you I've climbed promotion's ladder  
An' pu'd the switches fine  
But we've wandered many a weary shift  
And separated syne.





## ESSENTIALS OF RELATIVITY

It is about 15 years since Dr Essen and I last aired our different views on the merits of Special Relativity (*Nature*, 203, 395-6, 1964), and though it is clear from Dr Essen's article that there is still a difference in our fundamental beliefs, I must concur with some of his comments in the October issue of *Wireless World*. In particular I agree that students have had to swallow a lot of dogma and that relativity has become too much the province of mathematicians.

I further agree with Dr Essen that there are many cases when the observed rate of timekeeping is given by the relativistic Doppler shift and not simply by the second order dilatation factor. As an example of this I can quote an example erroneously interpreted by the science correspondent of the *Daily Telegraph* on 16th April, 1978. Adrian Berry described "What a high speed astronaut would see when travelling towards the earth, if by means of a fabulous telescope he could watch the hands of Big Ben. He would see the hands move round very slowly, just as he would see them move on his outward trip". Now this is pure nonsense, a high speed astronaut travelling towards the earth would see the hands of the clock of Big Ben spinning round very rapidly indeed and not very slowly as claimed by Adrian Berry. Mr Berry forgot the relativistic Doppler shift which includes the time dilatation that he was trying to illustrate but which also contains a predominant term proportional to the velocity of approach. Adrian Berry's example would have been perfectly correct if he had stated that the astronaut was passing the earth but not if he was approaching or receding. When I wrote to Adrian Berry pointing out his slip he replied "I must tell you that I absolutely stand by what I wrote". So much for spreading the gospel of true science!

There is much of Dr Essen's article with which I cannot agree but I will take up just two points:

- "more ticks are transmitted than received"; this is not correct, the length of time for which the whole series of ticks is received varies proportionately and the total number of events is conserved.
- "Einstein's prediction contains no mention of the ordinary Doppler Effect"; this is a remarkable statement; Einstein derived the electromagnetic Doppler effect in his 1905 paper and showed that it differed from the classical Doppler effect by the second order term. I quote directly from his paper:

"... the frequency  $\nu'$  of the light perceived by the observer is given by the equation

$$\nu' = \nu \frac{1 - \cos \phi \cdot v/c}{\sqrt{1 - v^2/c^2}}$$

This is Doppler's principle for any velocities whatever. When  $\phi = 0$  the equation assumes the perspicuous form

$$\nu' = \nu \sqrt{\frac{1 - v/c}{1 + v/c}}$$

We see that, in contrast with the customary view, when  $v = -c$ ,  $\nu' = \infty$ .

It is possible to take this equation as a starting point for the whole of special relativity and it then appears much more homely than the usual formal approach of progression from the rather abstract Lorentz



transformations. There is an unfortunate tendency to regard special relativity as a formal discipline and general relativity as a separate discipline, whereas both are really only a small part of the science of measurement. Common sense relativity recognises no barriers. It is surprising how much can be derived by using simple physics, the invariance of the local (two way) speed of light and Euclidean geometry to account for and predict local observations. This has had much success in the treatment of kinematically accelerated systems where formal special relativity is not strictly valid and where general relativity is entirely the wrong tool, for it can be applied only by invoking the equivalence principle from which general relativity itself was derived.

R. C. Jennison (Professor)  
Electronics Laboratories  
University of Kent at Canterbury

## SPECTRA OF TONE BURSTS

I note that Mr C. F. Coleman has replied, in your December 1978 letters, to my explanation of the tone burst signal published in June 1978. Mr Coleman still pins his faith on some quoted formula, when it is his own interpretation that should concern him. He does not accept, even intuitively, that when the carrier sine wave is shifted an amount  $h$  with respect to the start of the burst, this must produce a change in the phase spectrum of the carrier which is essentially equal to  $h$ , with proportionate, linear, phase changes in all other harmonic terms. He cannot deduce, even with his presently restricted knowledge and a little common sense, that the value of the function at the start of the burst,  $\sin(h)$ , could not possibly be arrived at from the tiny amplitude and phase changes to which he refers.

I now see that Mr Coleman does not begin to understand that an audible phase effect results from an instantaneous spectral amplitude change, which may also be viewed as a change in the time description of a signal. He points, instead, to the spectral amplitude changes due to a small order term in the expression for the tone burst spectrum as being significant, and insists that the dominant term in this expression is to be ignored. Far worse than his mistaken belief is the totally false argument, accusation and theory that he provides to support it.

He dismisses the linear phase shift mentioned above because he says it results from, and is equivalent to, a translation of the entire tone burst function by a time corres-

ponding to  $h$ . He then suggests that my undergraduate has confused the two and made an error in his computation.

It was quite clear from my analysis, and made as clear in an earlier letter (July 1977) that it was the argument of the carrier that was varied, not the position in time of the total function. This is confirmed by my undergraduate, who says that "no mistake was made; the computation involved sampling the signal over its fixed period, using the starting point as reference zero, which does not change as  $h$  is changed".

Mr Coleman's false accusation does not conceal his totally false argument that a shift of the carrier with respect to the square wave is equivalent to a time translation of the total function. In the former case, the one I have demonstrated, the reference zero is preserved, the waveform changes and with it the instantaneous amplitude spectrum of the signal. In the latter case, any reference time is lost (Coleman certainly ignores it), the waveform does not change and so its instantaneous spectrum is preserved. The phase shift I have demonstrated does not greatly affect the amplitude spectrum of the tone burst, but it importantly and audibly affects the amplitude spectrum with time, as is evident in the 'running' transform of the signal, a concept of which Mr Coleman is totally ignorant.

We have long since formed the impression that Coleman's views are unaffected by experience (for they are not even theoretically sound), but we can and do challenge him, or any more informed reader, to provide evidence that the minute effects he favours could be reliably reproduced, let alone detected, in preference to the changes we have demonstrated both in the theory and experiment.

If any further indication is needed of Coleman's shaky hold of this subject, it is surely to be found in his penultimate paragraph (still December 1978 letter). In this, his latest example of hand-to-mouth learning, he asserts that my analysis 'collapses' because I do not consider every  $\omega$  (as the transform requires!) in my approximation for the physical spectrum. In fact for all real functions  $f(t)$ , the transform has the property that  $F(\omega)$  and  $F(-\omega)$  are conjugates, with identical moduli, so that in general  $f(t)$  may be precisely recovered from its transform by considering only positive (physical) values of  $\omega$ , in the inverse relation

$$f(t) = \frac{1}{\pi} \text{Real} \int_0^\infty F(\omega) \cdot e^{i\omega t} d\omega.$$

We do not suppose that Mr Coleman will be any the wiser, but perhaps a little better informed.

Roger C. Driscoll  
The Polytechnic of North London  
London N7

## DISPLACEMENT CURRENT

The explanation given by Messrs Catt, Davidson and Walton (December 1978, p.51) of the flow of current 'through' a capacitor without resorting to Maxwell's concept of displacement current is attractive to me, because notwithstanding my immense respect for Maxwell I have always felt that displacement current was a kind of subterfuge to get over a logical difficulty\*. But

before wholeheartedly accepting this alternative I would like to be given certain reassurances.

At the foot of column 1 the authors point out that the parallel elements of the disk capacitor depicted can be regarded as transmission lines whose characteristic impedance ( $Z_0$ ) is continuously decreasing towards the far end. So there would be gradual reflection all the way. But in the mathematical proof  $Z_0$  is treated as constant and there is reflection only at the far end. This made me feel I was being conned.

According to Ampère's Law, the connecting leads carrying the charging current must be everywhere encircled by a magnetomotive force numerically equal to the current. In the authors' Fig. 1 the leads are horizontal and the plates are in vertical planes, parallel to one another and also to the n.m.f. around the leads. But what about the m.m.f. in the space between the plates, due to what we have become accustomed to calling displacement current? This current, being a continuation across the capacitor gap of the external circuit current, one naturally sees its m.m.f. also as in a vertical plane. Can the authors show clearly how this follows from the geometry of their transmission line currents, which flow everywhere at right angles to the current in the leads? This aspect is of some importance, since the propagation of radio waves depends on it. Can the authors convincingly get rid of displacement currents in space?

M. G. Scroggie,  
Bexhill,  
Sussex.

\*But I never had, or heard of, a difficulty created by imagining current having to flow across the capacitor plates faster than light. Where did the authors get that idea? And why wouldn't it apply also to the current in the leads?

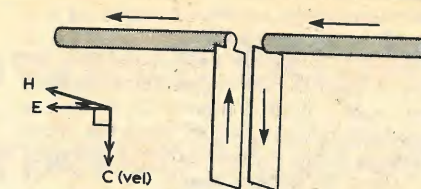
### The authors reply:

The article discusses a circular capacitor. The appendix discusses a rectangular capacitor in order to minimize mathematical complexity. The appendix proves that if a voltage source is switched across a resistor and a rectangular capacitor in series, a waveform results which approximates to an exponential. As Mr Scroggie points out, it does not prove the same for a non-rectangular capacitor.

If you ask us to resolve paradoxes in classical theory, you are asking us to say that we are saying nothing that is fundamentally new; you are asking us not to publish anything. Do you believe that "new" information is only acceptable if it indicates no flaws in the conventional wisdom, i.e. if it is not really new?

As to the m.m.f. in the space between the plates, this have never been measured. If it had been measured it would have been found to be non-uniform, and the reversed B. I. Bleaney and B. Bleaney ("Electricity and Magnetism," Clarendon, 1957, p.238) and others would not have written "... the field in between the plates is uniform...", which of course it is not; a TEM waveform advancing between the plates of a capacitor (= transmission line) creates a field behind itself but not ahead of itself.

The last paragraph of Mr Scroggie's letter is crucial. If the capacitor were rectangular and oriented much as shown in our Fig. 1(c) then no m.m.f. in the vertical plane would result from current in the capacitor plates. Vertical m.m.f. would mean that the waveform was not TEM, but we know that it is TEM and travelling vertically downwards between the capacitor plates. That is,  $E$  and  $H$



fields are at right angles to the (downwards) direction of propagation, and therefore are horizontal. This is no more paradoxical than trying to apply Ampère's Law to a TEM step travelling along any transmission line.

Ampère did not know that a TEM wave ( $E \times H$ ) travels forward between two wires at the speed of light. He did not know that a capacitor is a transmission line; he did not know about transmission lines.

These matters will be discussed further in a forthcoming article in *Wireless World*. A paper "The Heaviside Signal" will further clarify the situation (see "Electromagnetic Theory Vol 1," published by C.A.M. Publishing, 17 King Harry Lane, St Albans).

I. Catt, M. F. Davidson, D. S. Walton

Further letters on this subject will be published later. Ed.

## FERRITE ROD AERIALS

I have been reading with interest the article by Professor H. Sutcliffe in your December 1978 issue entitled "The effective length of ferrite rod aerials." The lack of treatment in the literature which Professor Sutcliffe finds is perhaps due to the fact that he appears to have renamed the parameter which is usually known as effective height. His own result seems to incorporate the Q factor for the resonant circuit in the form of a loss resistance so that the figure he arrives at is about 100 times the usually quoted results. The reader concerned with the design of ferrite rod aerials is recommended to refer to the article by Mr W. A. Everden in *Wireless World* September 1954, pp.440 to 444, which contains a number of useful graphs and information which enables the effects of different rods and coils to be calculated.

N. C. Helsby  
Department of Electrical Engineering  
Science  
University of Essex

### Professor Sutcliffe replies:

I agree with Mr Helsby that 'effective height' is in more common use than 'effective length' in expressing the relation between an electric field strength and the voltage in a circuit associated with an aerial. Nevertheless the word 'height' with its suggestion of elevation above ground or sea level appears to me to be misleading when applied to radio receivers with ferrite rod aerials and would be incorrect for horizontal polarization. With regard to the second comment it is of course important to distinguish between the two definitions of circuit voltage, the coil e.m.f. or the capacitor voltage at resonance, and as Mr Helsby points out these are in the ratio 1:Q. The article by Everden (WW Sept 54) certainly provides detailed information for designers but it uses the wholly artificial concept of 'effective permeability' which has to be presented by graphs and charts as a function of several variables. I believe that the concept of 'equivalent magnetic dipole length,' with an associated current generator  $HI/n$  across the  $n$  turns of the coil, is simpler

and more useful. As I pointed out in my article,  $I$  is in practice about one-third of the full length of the rod. If the equivalent circuit in series form is preferred then the Thevenin-Norton theorems show that for coil inductance  $L$  the corresponding magnitude of series e.m.f. is  $HI/n$  times  $\omega L$ . Since  $L$  is proportional to  $n^2$  the e.m.f. is proportional to  $n$  as we expect from the concept of linked flux.

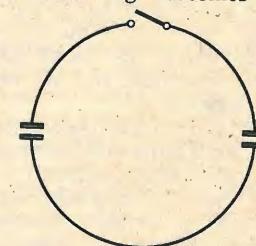
H. Sutcliffe

## "DID YOU KNOW?"

Despite having found Epsilon's article in the December 1978 issue most interesting, I feel that the explanation of his problem concerning two parallel capacitors (Fig. 1) is rather obvious.

Two different perfect voltage generators connected in parallel produce an indeterminate resultant voltage. The same applies to two similarly connected capacitors, because a charged capacitor is initially a voltage generator. Therefore Fig. 1 is a mathematical impossibility (!)

Furthermore, while it may be possible to achieve zero circuit resistance and have a perfect switch, inductive effects will always be present. Hence Fig. 1 becomes



which is an LC loop and this will allow the energy difference to radiate away into the surrounding space, without violating any of the two laws mentioned in the article.

P. Holy  
Worthing  
Sussex

Editor's note: We have received other letters commenting on this article but are waiting till the author (who lives abroad) can reply before we publish them.

## "SOFTWARE DABLERS"

I would like to comment on the article "Software dabblers may take over electronic system design" in your (December 1978 issue, p. 49) concerning Professor H. Barker's comments on the lack of circuit design technique required in today's digital systems.

I would agree that the number of discrete components within a system has diminished drastically; however, I cannot admit that circuit design technique suffers to the extent expressed in the article. Indeed modern Schottky t.t.l. requires that good r.f. practice be introduced into circuit design and layout. Many new techniques have evolved simply due to the fast edge speeds and data rates of logic signals.

Professor Barker than moves on to 'software' and expresses concern over its general availability; is this an application of the old adage 'A little bit of knowledge' or is he worried over some form of non-professional



encroachment into the subject? Unfortunately the second seems to hold sway, as he goes on to say that dabblers contribute nothing with their amateur approach. This I cannot hold with; very often it is the person 'dabbling' in a subject, other than his own, who will make important advances in knowledge.

As to the professor's call for unification, I would go further, and unify the 'hardware' and 'software' design functions as much as possible; very often a lack of communication between these two areas causes many unnecessary problems. A typical example of this co-ordination would be in the design stage of some form of computer using micro-programming techniques. The software requirements, in terms of instruction set, would be the guidelines followed in the hardware design and vice versa.

The professor also states that many engineers would be quite happy for 'software' to start and end at assembler level; why not? High level languages tend to separate the engineer from the hardware he is trying to control.

Finally, it must be asked, is the professor falling into the 'technology trap', a pseudo-religion based on the 'worship' of high technology; expressed in buzz words, total devotion and the exclusion of the untrained? This trap will finally spring when we are unable to see that our rate of pure advance is almost non-existent, we console ourselves with merely re-decorating the things that have already been invented, and use new terminology to hide the fact.

"The architecture of the modern computer is not significantly different from those of the 1940s" (*Computer Worship*; I. Catt.).

We must remember that man has, up to now, survived by being non-specialist; specialism could be the modern ice age, bringing with it stagnation.

M. A. I. Wilson,  
Swindon,  
Wilts.

## ELECTRONIC ORGAN TONE SYSTEM

Dr Ryder is to be congratulated on his interesting and elegant design for an electronic organ tone generating system (October issue) using additive synthesis techniques. However, I feel he dismisses rather too readily the alternative technique, that of subtracting unwanted harmonics from a complex waveform such as a sawtooth. He justifies this by commenting on the variations in loudness and tone quality which occur if a single tone-forming filter is used for each stop. These effects do, of course, occur but there are various well known ways to ameliorate the situation. Also (and contrary to what he says) many pipe organ stops exhibit similar effects!

Whilst I cannot enter here into lengthy debate over the pros and cons of additive versus subtractive synthesis, the constructor who is contemplating building an electronic organ would be well advised to pursue the matter further before he begins to invest heavily in hardware. In particular I would emphasise the following points:

1. A chorus of fully preformed tones using subtractive (e.g. diapasons of pitches 16, 8, 4, 2% and 2 feet plus mixture) is very much more satisfying than the equivalent number

of sinewave ranks, which sound cloying and "electronic" by comparison.

2. The "bite" of the reeds and mixtures of the classical organ is quite impossible to simulate using only a limited number of sinewave pitches per key, particularly when the upper frequency limit of the instrument is only 10kHz as in Dr Ryder's design.

3. The subtractive technique enables some delicate and beautiful effects to be obtained using resonant circuits which imitate the characteristic formant bands of real organ pipes or orchestral instruments. This is not possible with a sinewave instrument, where realistic synthesis is limited to flute and possibly diapason sounds.

4. In general, stops can be combined more successfully with a subtractive tone system than with an additive one. This facilitates a better imitation of the tonal building possibilities of the pipe organ.

These points can best be illustrated by comparing a sine-wave organ tonally similar to Dr Ryder's (e.g. a sine-wave Hammond) with one of the many subtractive instruments in existence. Local dealers are usually most accommodating!

Finally, having designed and built an organ using subtractive tone forming methods, I am convinced that the extra complexity required to generate sawtooth waves is amply justified, and that my preference for this method stems from practical experience and not mere academic considerations.

C. E. Pykett  
Malvern  
Worcs

Dr Ryder replies:

In reply to Dr Pykett's letter, I can only echo his advice to hear as many organs as possible before investing. One lesson I have learnt is that it is not possible to say generally that one method is good and another bad. Almost everything depends on the finer points of design, and good results are not achieved without attention to detail. (Perhaps this is where the home constructor has a real advantage.) In my own case, I concluded that a modernised synthesis approach offered the most direct – and economical – route to a "musical" instrument, especially one using full-size keyboards, and the prototype has not failed to please those who have played it, by now a considerable number. The limitations of this particular design are quite specifically mentioned in the articles, but this does not mean that the synthesis method is itself limited, as witness the achievements of Compton. (In my view, Hammonds are not comparable.)

Amongst the options, there is a brief discussion of the use of pulse waveforms, subtractively filtered, primarily to help with reed tones. Mixtures do not present the same problem, and sparkling upperwork can be obtained even with pure sines limited to 10kHz. This is one object of the suggested coupling circuits.

The combining of stops is a subtle effect, depending on many aural cues, and again, I do not think one can safely generalise. Too-perfect blending is undesirable, and small frequency/phase differences, such as result from using more than one set of generators, are probably more important than precise waveforms. I have not seen a satisfactory discussion of this question, which would have to start by defining some rather elusive concepts.

I confess that I am much concerned with

regulation and balance which, like evenness of touch, are important ingredients of a satisfying instrument, perhaps more important than is generally realised. I did not say that all pipe organs are satisfactory in this respect, but the better they are, the less obstacle there is between the player – or listener – and the music. Pipe organs are regulated in detail to suit their final siting; electronic organs also have the loudspeaker problem, and it was a particular object of my design to provide for note-by-note regulation as well as overall balance. In this respect, a sinewave Hammond, as suggested by Dr Pykett, may not be heard to the best advantage. I am not familiar with all their models, but I believe that some at least had only a limited number of frequencies, and so "borrowed" the harmonics, a practice which does appear to contribute significantly to an "electronic" sound. It also weakens the effect of deriving mutations from another department, if indeed this was done by Hammond. I really must protest against Dr Pykett's suggestion that my design may be judged by listening to a veteran electro-mechanical instrument in a music shop! For obvious reasons I have made available a recording, but I am arranging personal trial where it is practicable.

The judgement of value for money/effort must be an individual one. I personally believe that the additive method is the most effective, but there is no reason why subtractive voices should not be used as well, perhaps based on pulse waveforms, which are easily generated and keyed. My design is essentially a foundation to build on, and I hope that in due course constructors will be able to communicate improvements they have made.

David Ryder

## RADIO AMATEURS' EXAMINATION

I feel I must reluctantly disagree with Pat Hawker (October issue), and try to ease the minds of would-be candidates. The new examination is not more difficult, but it does cover the whole syllabus.

Valves have been left out for two reasons: there is already plenty of material in the syllabus; and valves (with few exceptions) are out of date.

Your readers should not be alarmed by the use of microcoulombs. On page 61 of your October issue I read "The capacitor is of course a component for storing charge,  $C = Q/V$ ...". Knowing this equation and the meaning of  $\mu$ , a candidate will find the question to be quite easy. (Perhaps some of your advertisers should take a course of instruction. Whilst PF and pf are absurdities, the use of mF as a trade description could lead to an error of 1000 times!) Incidentally the instrument on page 61 is marked "Capacity". When I was a lad we measured capacity in jars!

Let us get up to date! According to the Advisory Council for Applied Research and Development, "all the available evidence suggests that only 5% of British firms are aware of and are exploiting the new semiconductor technology".

I'll now go and listen-in on my Japanese solid-state wireless set.

G. C. Oxley  
Chesterfield, Derbyshire

# Microwave duplexing for radar systems

And a discussion of mixer noise in radar receivers

by Alan G. Hood, M.Sc., M.I.E.E. Kingsway Technical College, Dundee

WITH FEW EXCEPTIONS pulsed radars have always used some form of TR-cell duplexer<sup>1</sup>. The TR cell is a resonant chamber filled with an appropriate mixture of gases at low pressure. A coupling aperture or window is provided at either end sealed with a dielectric which is usually glass but occasionally ceramic.

The two main types of cells are narrow-band but tunable, and broad-band fixed tuned. The tunable cells consist of a re-entrant resonant cavity with non-resonant windows, while the broad-band type have resonant windows with one or more resonant irises suitably spaced along the waveguide body. Breakdown gaps are provided in both types of cell, and to ensure rapid ignition at high power, a supply of ions is maintained in the gap by a continuous weak auxiliary discharge. This discharge requires an extra electrode, known as the "keep-alive" or primer which draws about 100  $\mu$ A from a high impedance d.c. source. A small quantity of radioactive gas, usually tritium, included in the gas filling also acts as a primer.

In the gaps the electric field builds up to a much greater intensity than in the normal waveguide. If the field in the gap is insufficient to cause breakdown, as in the case of the echo signals, the cell transmits the incident wave from the antenna in its frequency band to the receiver with little attenuation. Under the influence of the transmitter pulse, however, the cell becomes intensely ionized. A plasma usually forms across the input window and the de-tuning effect reflects almost all of the incident r.f. power effecting a switching action which connects the antenna to the transmitter.

On de-ionization, the antenna is again switched back to the receiver. Ionization time is very short, while de-ionization or recovery time may be several microseconds or longer. The recovery characteristic of a duplexer determines the short-range performance of a radar. Too long a recovery period will prevent detection of small targets at short range, while too short a recovery time may result in receiver overload with short-range targets.

The branched duplexer and the balanced duplexer are the most widely used gas tube duplexers. The branched duplexer shown in Fig. 1 uses a TR cell

together with a second gas-filled cavity resonator called an ATR cell placed on a stub line, or directly in the wall of the main line, between the transmitter and the branch leading to the TR cell.

During the transmission period both cells ionize and cause little reflection of the transmitter wave. The ATR cell is resonant at the transmitter frequency and it is tightly coupled to the input line.

As a result a high standing-wave ratio for low-level signals is produced in the main line between the TR cell and the ATR cell. The spacing between the two branches is chosen so that the received signal power is coupled into the receiver with little reflection or absorption due to the line terminating in the transmitter.

The branched duplexer has certain

## The need for duplexing

The applications of modern radar are vitally important as the eyes of today's complex weapons and navigational systems. In operation, most radar transmitters send out a short pulse of microwave energy, which may strike a reflecting target that scatters it. The scattered wave, still in the form of a short pulse, although very much reduced in amplitude, is picked up by the radar receiver which detects it. The majority of transmitters generate trains of these pulses at peak powers in the range 1k to 1MW. The range of the target is obtained from the length of time between transmission of the pulse, and reception of the echo, while the direction is found from the orientation of the antenna when the echo is received.

The relation between the power  $P_t$  of the transmitted pulse and the power  $P_r$  of the received echo is well-known as the radar equation, which is

$$P_r = \frac{P_t G A \sigma}{16\pi R^4}$$

where  $G$  is the gain,  $A$  the effective area of the radar antenna, and the target has a scattering cross section  $\sigma$ . The maximum range  $R$  at which a target can be detected is obtained from this equation if the value of  $P_r$  corresponding to the minimum detectable signal and the other system parameters are inserted. It is common practice in radar systems to use a single antenna for both transmitting and receiving whenever possible. This is done not only to ensure collimation of the transmitting and receiving patterns of the small but highly directive scanning antennae used in microwave systems, but to minimize space and weight requirements of the antenna system, a critical factor particularly in airborne system design. To use a common antenna for both transmitting and receiving poses a severe receiver isolation problem because of the extreme difference in signal level,

between the strong transmitted signal and the feeble echo return from a distant target. Some form of duplexer is required.

The prime function of a duplexer is to protect adequately the vulnerable radar receiver – which generally contains semiconductor devices such as point-contact diodes, Schottky-barrier diodes, tunnel diodes, or varactor diodes – from the paralyzing power of the transmitter with a minimum insertion loss to both transmitted and received signals. Because of the extremely fragile nature of the best receiver elements, great effort must be made to achieve adequate isolation. A radar duplexer may be required to isolate a 100kW pulse from a mixer crystal which would deteriorate if exposed to 100mW pulses. This is a power ratio of a million-to-one, indicating the need for more than 60dB isolation between transmitter and receiver for the duration of the transmitted pulse.

Where transmission and reception take place sequentially, as in pulsed systems, this high isolation is achieved using transmitter-actuated gas tube switches of one form or another, usually TR (transmit-receive) cells supplemented in some cases by ATR (anti-transmit-receive) cells, pre-TR tubes, or by solid-state devices such as ferrite circulators or varactor limiters. Duplexing in f.m. and c.w. systems, on the other hand, cannot be accomplished by switching as transmission and reception are simultaneous and continuous; so other means must be sought. Pulsed systems are by far the most common however, and require a duplexing system which is amplitude selective.

## Introductory reading

Airborne Radar by D. J. Povejsil, R. S. Raven & P. Waterman. Van Nostrand, 1961.  
Principles of Radar by J. F. Reintjes & G. T. Coates. McGraw Hill, 1952.



disadvantages, notably rather limited bandwidth. In its de-ionized state the ATR cell presents a serious mismatch to the magnetron, and under certain phasing conditions troublesome "moding" can occur<sup>2</sup>. For this reason the magnetron-to-ATR spacing becomes important in a branched duplexer.

The balanced duplexer now being used extensively offers several advantages. The ATR cell together with its mismatch problem are eliminated, better receiver protection is achieved, and the insertion loss can be lower. The balanced duplexer employs two TR cells suitably mounted between two 3dB hybrid junctions in a bridge arrangement. Several types of hybrid junctions can be used, but the slot hybrid duplexer<sup>3</sup> has found widest acceptance because of its compactness and unexcelled broadband performance. A dual TR cell having a common gas filling is used to obtain best bandwidth through closely balanced characteristics.

The signal path in a slot hybrid duplexer for the transmit condition is shown in Fig.2(a). Transmitter power is split by the 3dB hybrid and on ionizing the TR cells is reflected, but because of the quadrature phase characteristic of the slot hybrid, the power leaves by the adjacent antenna port. After recovery, the received echo signals follow the paths shown in Fig.2(b). The load on the fourth arm of the duplexer serves only to absorb the leakage power during transmission.

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THE IDEAL DUPLEXER is a matched, lossless three-port device in which the first port couples only to the second, and the second couples only to the third. The microwave ferrite circulator<sup>4</sup>, if perfect, would be the ideal duplexer for either c.w. or pulsed radars. In practice, circulators can provide 30dB isolation and insertion losses of a fraction of a decibel up to moderate powers. A duplexing arrangement using a four-port circulator is shown in Fig.3; a differential phase-shift type<sup>5</sup> is commonly used because of its superior power handling capability.

To achieve high isolation the antenna must be exceptionally well matched. The leakage through the circulator will depend more on the quality of antenna match than on circulator performance and is likely to be between 15 and 20dB down on the transmitter power, the shape being essentially the same as the transmitter output except that the leading and trailing edges of the pulse may be slightly modified due to mismatches in the antenna run. A TR cell is used for receiver protection.

The advantage of this duplexing system is that the power incident on the TR cell is greatly reduced, which prolongs cell life. Also, the reduced ionization intensity in the cell improves the recovery time and gives better close-

range reception of weak signals. The circulator provides load isolation for the transmitter, as the antenna reflections are absorbed in the fourth port load during transmission.

At higher powers balanced gas duplexers are preferable as the insertion loss of a circulator increases with power level due to ferrite non-linearity. TR cells may be used alone at peak power levels of up to about 100kW, depending on the frequency band and the life required from the cell before replacement. At higher levels it is advisable to use pre-TR tubes which reduce the power incident on the cell and provide the

## Noise figures of radar receivers

Microwave radar receivers<sup>7</sup> must be able to detect the weakest possible echo signals from distant targets which may be comparable with, or even weaker than, the noise in the system. The level of atmospheric noise is very low, so it is essential to do everything possible to reduce the receiver internal noise, and to reduce losses in the input circuits. Almost every common type of receiver has found application at some time, but the superheterodyne receiver embodying a point-contact diode mixer<sup>8</sup> is now used almost universally.

Crystal mixer development has been the subject of a great deal of effort, and until recently no microwave r.f. amplifier to precede the mixer could be made to give a better overall receiver noise figure than a crystal mixer used directly. Because of the conversion loss associated with the passive mixer, the noise figure is fairly high, and an intermediate frequency in the range 30 to 60MHz is usually chosen to achieve the lowest-noise i.f. amplifier.

The noisiness of a crystal mixer is characterized by the amount of noise produced by the mixer compared with the noise from a resistance at the same temperature, expressed as a noise temperature ratio  $t_m$ . The noise factor of a superheterodyne radar receiver is

$$F_{rec} = 1 + \frac{T}{T_a}(L_t - 1) + \frac{T}{T_a}L_t[(t_m - 1) + (F_{IF} - 1)]$$

assuming a negligible noise contribution from the local oscillator which is realistic if a filter or a balanced mixer is used. In this expression  $t_m - 1$  is the excess noise of the mixer,  $F_{IF} - 1$  is the excess i.f. amplifier noise, and  $L_t$  is the product of the conversion loss of the mixer and loss in the microwave transmission circuitry between the antenna and mixer expressed as a power ratio,  $T$  is the temperature of the receiver,  $T_a$  is the temperature of the antenna, and noise figure is  $10 \log_{10}$  of the noise factor. A radar receiver is however connected to a directional antenna which can be represented as an equivalent generator at less than room temperature when the antenna is directed toward space. An antenna temperature under this condition may be about 4°K, and from the above equation the noise figure of the best radar sets is about 30dB.

Noise figure is usually defined with respect to room temperature, about 290°K, and under such a definition  $T/T_a$  is unity. The equation then reduces to

$$F_{rec} = L_t(t_m + F_{IF} - 1)$$

which represents the condition under which the noise figure of a crystal mixer is measured in the laboratory. The input termination, usually a noise tube in the unfired state, is at room temperature and  $L_t$  is the product of the conversion loss of the mixer and the insertion loss between the noise tube and the mixer. A good mixer diode has a noise figure of about 6dB at X-band (8.2 to 12.4GHz) which includes a noise contribution from the following i.f. amplifier.

A radar receiver commonly employs a balanced mixer which has the advantage of local-oscillator noise cancellation, and in addition each mixer diode is exposed to only half the duplexer leakage.

extra isolation needed with the higher power. The pre-TR tube is very often a sealed coaxial quartz tube with a low-pressure gas filling and is used mounted in an iris across the waveguide.

Recently a solid-state device has been proposed which may ultimately be competitive with the TR cell at the lower power levels. Lax & Button describe an X-band ferrite-varactor limiter<sup>6</sup> which operates up to 10kW and has a leakage comparable with a TR cell.

To obtain a deeper insight into duplexing and receiver diode burn-out a discussion of radar receivers is needed.

Because of the conversion loss of the crystal mixer, the receiver noise figure is highly dependent on a low i.f. noise figure. While in some applications improvement in receiver noise figure may be relatively unimportant due to a high level of external noise, the operating frequencies of most radars are in the spectral region where external noise is low, and a low noise figure can be fully used.

The range of an ideal pulsed radar in terms of its system parameters is

$$R = \sqrt[4]{\frac{P_t \sigma A^2 \tau}{\lambda^2 k T_0 F_{op}}}$$

where  $P_t$  is peak transmitted power,  $\sigma$  effective target scattering cross section,  $A$  effective antenna area,  $\tau$  pulse duration,  $\lambda$  wavelength,  $k$  Boltzmann's constant,  $T_0$  290°K, and  $F_{op}$  operating noise factor. The range is a slowly varying function of noise figure; for example, a 6dB improvement in  $F_{op}$  yields only a 40% increase in range, but a reduction in noise figure can be as effective as an increase in transmitter power or antenna aperture. Often a 3dB improvement in noise figure may be more economical to achieve than doubling the transmitter power or increasing the antenna area by 40% especially for high power radars.

From a consideration of the noise figure of cascaded networks, a large available power gain in the first stage minimizes the noise contributions of the later networks. A low-noise r.f. amplifier preceding the crystal mixer would provide the power gain required to offset the loss of the mixer. RF amplifiers that could be used are the travelling-wave-tube amplifier, the simpler, low-noise parametric amplifier<sup>9</sup>, or more recently the field effect transistor amplifier.

The noise factor of two cascaded networks with noise factors  $F_1$  and  $F_2$  and available power gains  $G_1$  and  $G_2$  is

$$F = F_1 + \frac{F_2 - 1}{G_1}$$

If the first network is a parametric amplifier, noise factor  $F_p$  and gain  $G_p$ , and the second a crystal mixer and i.f. amplifier, the expression for receiver noise factor  $F_{rec}$  can be substituted into the above to give a new receiver noise factor of

$$F_{rec} = F_p + \frac{1}{G_p}[L_t(t_m + F_{IF} - 1) - 1]$$

This shows that if a parametric amplifier with a noise figure of 4dB and a gain of 20dB is followed by a mixer and i.f. amplifier with a 12dB noise figure, the overall receiver noise figure is still less than 4dB.

In general, parametric devices are frequency converters; they are amplifiers when the output is at the same frequency as the input, and they use non-linear varactor diodes which act as voltage-dependent capacitors drawing energy from a pump source which is usually much higher in frequency than the signal for a low noise figure. The reflection type of amplifier is most common and requires a ferrite circulator to separate the input and output signals.

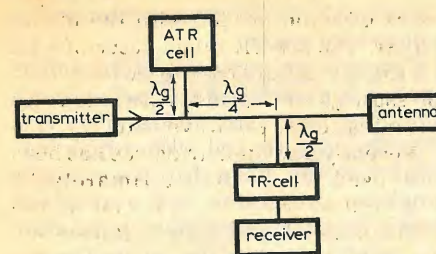


Fig. 1. One of the most widely used gas duplexers is the branched duplexer using a TR cell and a second gas cavity resonator on a stub (ATR cell).

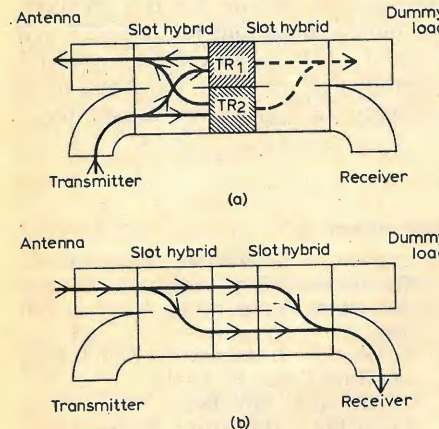


Fig. 2. Balanced duplexer avoids the use of an ATR cell using two TR cells between two hybrid junctions. Transmitter power is reflected by the TR cells and leaves by the adjacent port (a). Lower figure shows echo return route after cell recovery (b).

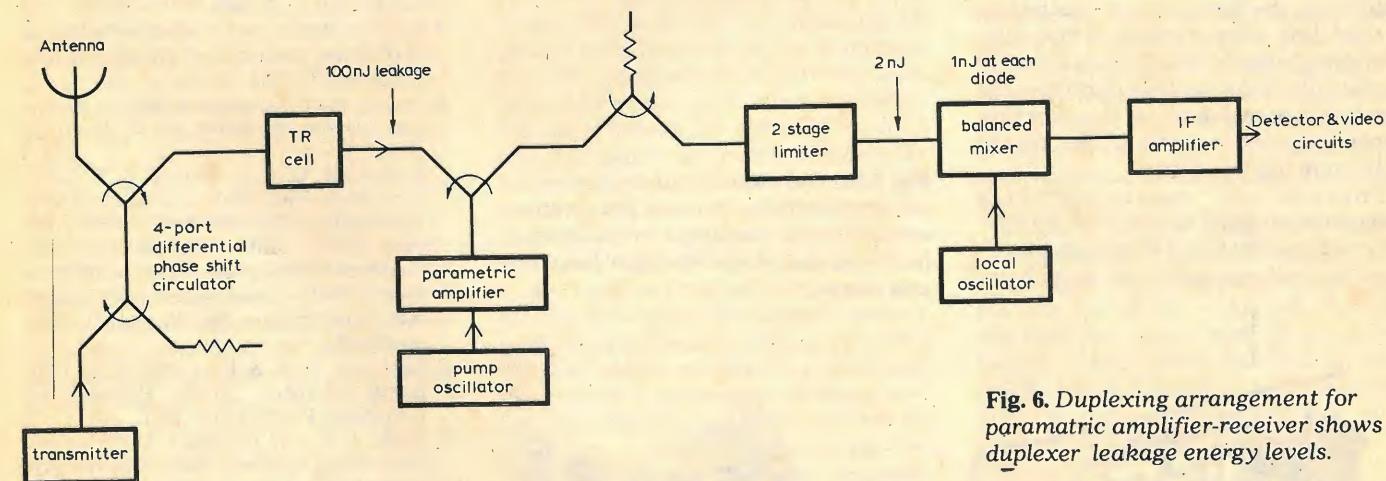


Fig. 6. Duplexing arrangement for parametric amplifier-receiver shows duplexer leakage energy levels.

## Duplexer leakage and receiver burn-out

The problem of "burn-out" is of prime concern to microwave diode users, and may be defined as a change in the rectifying and converting properties of crystal rectifiers as the result of excessive electrical overload. Overload may be due either to accidental static discharge through the diode, or to microwave or pulse power delivered to the diode. Because the rectifying element is highly sensitive, it is subject to damage when excessively high current densities pass

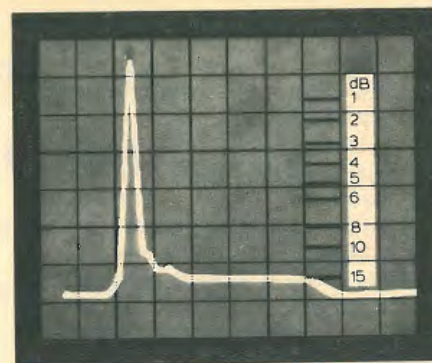


Fig. 4. TR cell leakage, shown above from a sampling oscilloscope trace-one with expanded time scale showing random variation of spike leakage energies, is the most common cause of microwave diode burn-out.

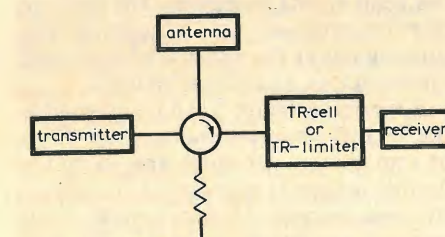


Fig. 3. Up to 30dB of isolation can be provided by a four-port ferrite circulator with only a fraction of a dB insertion loss.

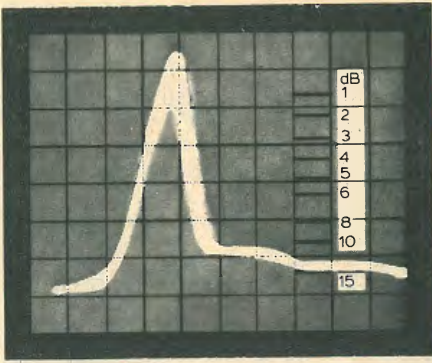


Fig. 5. Cumulative distribution of cell spike language energies.

through the rectifying barrier. Lower frequency diodes with a larger junction area are therefore more resistant to burn-out.

The most common cause of burn-out is duplexer leakage power. The TR cell in a duplexer requires a finite time to ionize and attain full attenuation under the influence of the transmitter pulse. Thus the leakage through the cell consists of a short but high pulse called the spike immediately followed by a long low-amplitude pulse called the flat. At X-band the spike duration is typically 2

to 10 ns, at its half-power point, and depends on the amplitude and rate of rise of the magnetron pulse. A sampling oscilloscope is required to view such short pulses, and a photograph of a TR cell spike and flat leakage is shown in Fig. 4, and the spike on a reduced time scale next to it.

The leakage from a TR cell is not constant from pulse to pulse. Due to the random nature of ionization of the gas<sup>10</sup>, breakdown occurs at different levels, and the spike leakage energies for a train of pulses have a statistical



distribution, the width of which will depend on the operating conditions, on the cell type, and on the shape of the leading edge of the transmitter pulse. As a result, the mean spike leakage energy for a cell may be a 20nJ per pulse which occasional spike pulses can have energies as high as 500nJ per pulse depending on the particular operating conditions.

Some measurements of spike leakage energy distribution for cells in S-band are shown, with cumulative distribution shown in Fig. 5. Cell leakage distributions are skew with a sharp cut-off at the lower energies. At high energies the frequency of occurrence normally decreases rapidly with increasing energy and approximates to a log-normal distribution. In addition to the steady-state distribution of spike energies, high energy pulses are also generated in the transient conditions of switching the magnetron modulator on and off.

McMillan & Wiesner<sup>11</sup> conclusively demonstrated that the spike part of the leakage was the primary cause of mixer burn-out. The thermal relaxation time of the diode junction is of importance when burn-out is due to short pulses. An effective value can be estimated or experimentally determined, and for mixer diodes at X-band is about 10 ns. For pulses shorter than this, the heat developed at the diode junction does not have time to appreciably diffuse away and the temperature reached is determined solely by the energy dissipated at the contact and by the thermal capacities involved. Therefore, if the pulses are shorter than 10 ns their energy content, and not their shape determine the temperature reached at the junction and the extent of burn-out. For leakage pulses longer than about 10 ns the pulse shape is of importance; the temperature reached at the junction depending on the peak power rather than on the pulse energy.

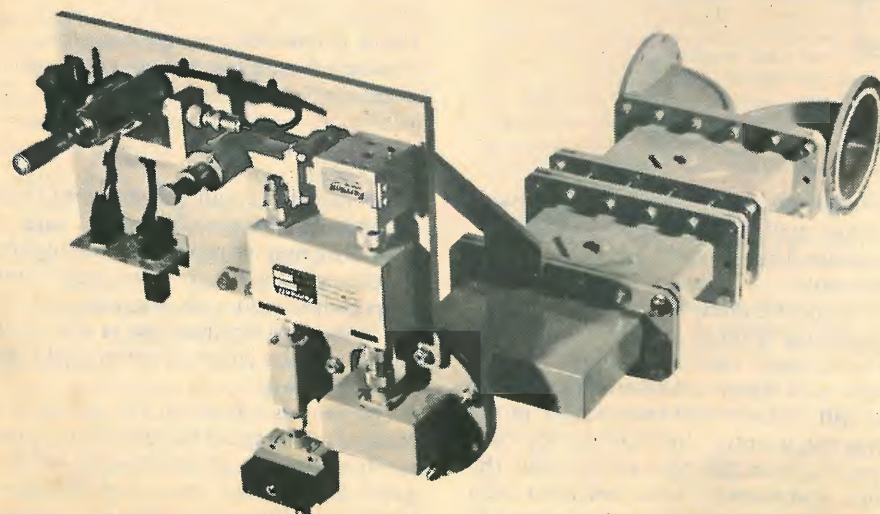
Burn-out in radar receivers sometimes consists of a sudden deterioration caused by a rare very high energy duplexer leakage pulse, but often how-

ever it consists of a gradual deterioration which takes place over several hundred hours of operation. Damage of this type is not a direct result of heating, but may be associated with the diffusion of impurities in the semiconductor or a chemical reaction at the junction. Such processes consuming a great deal of time at room temperature would be accelerated by the high temperatures attained during a leakage pulse.

The leakage from a TR cell will depend on the particular application for which the cell has been designed. A cell intended to protect mixer diodes may have a spike leakage energy of 10nJ or less, while a cell intended for parametric amplifier protection can be made to have a lower insertion loss and be run without priming, because leakage of a few hundred nanojoules per pulse can be tolerated. Cell design is always a compromise between leakage energy during transmission, and attenuation of the echo during reception. The leakage can be reduced by increasing the priming but at the expense of increased insertion loss, and primer noise<sup>12</sup>.

A varactor limiter<sup>13</sup> can be used after the TR cell where extended mixer life is of importance. Because the varactor limiter action is non-linear, its effectiveness increases proportionally with leakage amplitude. Isolation for flat leakage may be only 3dB, but for high amplitude spikes isolations of about 15dB can be attained, and the statistical fluctuation of spike energies is reduced. Although this leakage reduction is achieved at the expense of increased insertion loss adding to the receiver noise figure, a TR limiter is often the

Fig. 7. S-band duplexer and receiver assembly showing in co-ax parametric amplifier with circulator and X-band pump oscillator and varactor limiter and mixer.



optimum device for mixer protection in a duplexing system.

The optimum duplexing arrangement for a receiver using a parametric amplifier is shown in Fig. 6. A low insertion loss TR cell with a high leakage can be used for parametric amplifier protection and a varactor limiter for mixer protection placed after the parametric amplifier where the extra insertion loss has little effect on the overall receiver noise figure.

In certain special systems a TR cell may not be necessary; for example, if the maximum peak power is only a few watts, a circulator and limiter may be adequate protection for the receiver. But most applications require a TR cell even if only to guard against the possibility of receiver burn-out by a neighbouring high power transmitter.

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## Some milestones in electronics

An interview with Professor Bernard Tellegen

by Arthur Garratt

In this article Professor Bernard Tellegen, inventor of the pentode, discoverer of the Luxembourg Effect, pioneer of the gyrator and great figure in the history of radio talks to Arthur Garratt, a British scientific and industrial consultant now living in France.

BEFORE the invention of the transistor, the three most significant inventions in radio were the diode (Ambrose Fleming, 1904), the triode (Lee de Forest, 1906) and the pentode by Bernard Tellegen in 1926.

The pentode originated in the Philips Research Laboratories, the *Natuurkundig Laboratorium* or *Nat Lab* as it is generally called, at Eindhoven in the Netherlands.

The Nat Lab was set up in January 1914 under the directorship of Dr Gilles Holst, who had previously worked in Leyden with Kamerlingh Onnes — one of the great figures in low temperature research. Holst soon had an assistant, Dr Ekko Oosterhuis, who became Holst's second-in-command. (Oosterhuis, incidentally, was the grandfather of the British golfer, Peter Oosterhuis.) After the first world war the Nat Lab grew apace and among other well-known members of its staff were Bal- thasar van der Pol who published some two hundred papers on theoretical aspects of radio, with particular emphasis on relaxation oscillations, and Klaas Posthumus who, as we shall see, was a co-discoverer of negative feedback and also carried out pioneer work on split-anode magnetrons. Apart from Bernard Tellegen, a prolific generator of ideas who still comes to the Nat Lab regularly to "work on things which interest him" and Klaas Posthumus, the other men we have mentioned are now dead.

Arthur Garratt visited the Nat Lab at Eindhoven and recorded an interview with Bernard Tellegen who explained how the pentode and gyrator came into being and also traced the history of wave interaction in the ionosphere, the so-called Luxembourg Effect.

Bernardus Dominicus Hubertus Tellegen, to give him his full name, was born in 1900 and trained as an electrical engineer at Delft Technical University, graduating in 1923. After completing his military service he joined the Nat Lab in May 1924 and spent his entire working

career with Philips. Arthur Garratt asked him if he immediately joined Van der Pol's radio group:

TELLEGEN Not immediately, this happened a few years later. When I first joined Philips I worked with Oosterhuis. One of my first assignments was a tungsten arc lamp which had recently been developed in the Nat Lab. This operated well on d.c. and we tried to make it work on a.c. by using some kind of transformer — unfortunately this was not successful. After that I worked on the development of a battery eliminator; this was taken over by someone else who carried it through to production.

It was after this that I joined Van der Pol. From the beginning I was more theoretically than practically minded and this naturally caused me to gravitate towards Van der Pol, and I started in the field of radio about which I didn't know a great deal at that time. However, I studied a paper of Van der Pol's — a general paper on electron paths; this was my introduction to radio. I then began to study amplification. W. Schottky had written some papers on screen grid tubes (tetrodes) and these interested me very much. I also read papers on the use of triodes as output tubes and I observed that the triode should have a low internal resistance in order to get the maximum output. Then I put the two tubes together in my mind — I realised that with the tetrode you move the anode-current/grid voltage characteristics over to the left and this was clearly a desirable thing to do. I did some calculations — nothing about secondary emission at this stage — and I came to the conclusion that a screen grid tube, notwithstanding its high internal resistance, was very well fitted to the role of an output tube. You must remember that these were the days when the anode supply was from dry batteries and we wanted the maximum output from a given battery voltage. From this starting point I saw that one should not only get a higher output but also greater stage gain and less frequency distortion because the current in the loudspeaker should then be proportional to the control grid voltage. Putting all these things together led me to the conclusion that a tetrode should make an excellent output tube.



Professor Bernard Tellegen, now aged 79, at the Natuurkundig Laboratorium in Eindhoven, Netherlands

Of course, when you try this out you immediately come up against the problem of secondary emission. Secondary electrons are always emitted when primary electrons strike an electrode with an energy above about 10eV. In a triode they have no effect because they are drawn back to the electrode from which they are emitted, but in a screen grid tube secondary electrons emitted from the anode are attracted to the screen grid when the anode potential falls below that of the screen. This produces impossible distortion if the tube is driven hard — as an output tube must be. So I introduced a suppressor grid between the screen grid and the anode — this prevented the exchange of secondary electrons between the anode and the screen grid.

I talked with Holst about other means of suppressing secondary emission and he proposed some methods which were put into the patent — this was the reason that the patent itself is under the names of both Holst and me. In fact Holst's suggestions were never put into practice, the suppressor grid was successful and that was that.

GARRATT When you first constructed a tube with a suppressor grid, did it work right away or did you have to do more experiments?

TELLEGEN We had to do some experiments to measure anode current as a function of anode voltage at various values of suppressor voltage. You can find these in the original paper together with some of the results. The optimum dimensions for the suppressor grid were later studied by Jonker in the Nat Lab, but at the beginning we did not have much difficulty in finding a reasonable construction for the suppressor grid.



**GARRATT** When you found this was a satisfactory tube, had you any realisation at that time of the implications of the pentode to the radio industry?

**TELLEGEN** Well the effect it would have on the industry was too big a world for me! As far as I was concerned it was a tube which could be used practically in a radio receiver. It was just at this time that the first receivers that Philips had decided to build were under construction and new ideas were coming in from every side. Not only was there the pentode, but also the indirectly heated triode as a detector and what was called the 'Hull tube' (named after A. W. Hull who had studied a screen grid tube designed to have a very low anode/grid capacitance). Hull had stated in the paper describing this tube that it was purely experimental and was not intended as a production tube. It was Posthumus who developed the idea into a practical tube. So our first commercial mains set had a screen grid r.f. amplifier tube with a very low anode/grid capacitance, an indirectly heated triode and a pentode output tube.

**GARRATT** Was the pentode indirectly heated as well?

**TELLEGEN** No, the pentode was directly heated off a.c. There were actually two sets, a battery version which was, of course, d.c. throughout and the a.c. mains version.

**GARRATT** This was the a.f. pentode. How did the r.f. pentode come into being?

**TELLEGEN** I think the first r.f. pentodes were constructed by Bell Telephones who needed rather high r.f. voltages. They realised the importance of the pentode for r.f. amplification — this was also covered by the original patent.

**GARRATT** The patent is interesting. Philips patented the pentode in eighteen countries, but it was never patented in the Netherlands — why was this?

**TELLEGEN** The reason was that some date was overlooked. When you apply for a patent in the Netherlands you get a letter back saying, "We have such and such objections and we want certain information" before a certain date. Because of some administrative error, the expiry date was overlooked. Despite efforts to get permission from the Dutch Patent Office to extend the time, they refused and this is why the patent was not granted in the Netherlands.

**GARRATT** This was a patent of immense importance to Philips. It raised a great deal of money...

**TELLEGEN** Yes, and it opened doors — the doors of other important electronics firms.

**GARRATT** It was about this time that your colleague, Posthumus, was working on negative feedback. Negative feedback is generally attributed to H. S. Black of Bell Telephones, but Posthumus was working on it almost simultaneously.

**TELLEGEN** You can say "simultaneously" and quite unknown to each other.

**GARRATT** In fact there was an interesting difference. Black used a bridge circuit while the circuit Posthumus developed was much more like the feedback circuitry used in radio today.

**TELLEGEN** Yes, Black concentrated largely on telephone applications where a bridge was necessary, while Posthumus was interested in radio sets — so the two men had rather different points of view.

**GARRATT** Negative feedback took some time to get into commercial practice.

**TELLEGEN** It was used in radio sets because there were objections that pentode output tubes gave third harmonic distortion due to the impedance of the loudspeaker rising with frequency. This could be overcome by applying negative feedback but it took a number of years before it was used generally in radio receivers.

**GARRATT** What about its use in telephony?

**TELLEGEN** At that time Philips were not in the telephone business while Bell, on the other hand, were primarily interested in telephones and, I believe, they used negative feedback commercially before Philips.

**GARRATT** At that time, at the end of the twenties and beginning of the thirties, the Philips Nat Lab was very much in the front row. In electronics, you had produced the pentode, Posthumus had produced negative feedback simultaneously with Black and he had also done pioneer work on the magnetron, Van der Pol was doing a great deal of important theoretical work — this must have been a remarkable time and place to be doing research?

**TELLEGEN** That's right. It was very stimulating to work on these problems and it was not very difficult to make progress in different directions and to find new applications for one's ideas. It was very exciting for all of us.

**GARRATT** A little later on came the beam tetrode output tube, another way of avoiding the effects of secondary emission...

**TELLEGEN** And a way of avoiding the patents!

### The Luxembourg Effect

**GARRATT** You were also a pioneer in the discovery of what was originally called the 'Luxembourg Effect'.

**TELLEGEN** Yes, it's now called "wave interaction". After some time with Van der Pol I had moved to work in the group headed by Oosterhuis. Van der Pol and I looked at things from a different point of view. He was a physicist, I am an engineer. Van der Pol was almost exclusively interested in understanding things, I thought understanding was no use if you didn't apply the results in the right way — that's the point of view of an engineer. This is why after some years I joined Oosterhuis's group which was busy developing radio receivers. One problem we faced was cross-modulation, so we built straight sets with two tuned circuits in front of the first tube — by increasing the selectivity before the first tube we reduced the cross-modulation. We had built such an experimental receiver in which we were quite certain we had eliminated all cross-modulation and I tested this out at home. I found that when I was listening to the Swiss station Beromünster I could hear another station faintly in the background. This puzzled me a lot. I thought to myself this can't be cross-modulation as we had taken every possible measure to eliminate it. I could only hear the second station when there was silence in the Beromünster programme. I tuned around and found that the interfering programme was being broadcast by Radio Luxembourg.

**GARRATT** Which was a long-wave station.

**TELLEGEN** Yes, a long-wave station. I called my friend Rinia and told him to listen in. He did so and got the same effect. At first we thought it might perhaps be something complex happening in the electric light mains. So we went outside the town and repeated the experiment... and got the same results. I then realised that Luxembourg was just about half way between Eindhoven and Beromünster and was a very powerful station transmitting a lot of energy into the ionosphere and something must be happening in the ionosphere that affected the signal from Beromünster on its way to Eindhoven, and this was the reason for the interference.

The ionosphere was outside my field but when people reported similar experiences with other receivers it became obvious that the effect was not caused by something in the receiver but must occur in the ionosphere. So I wrote a letter to *Nature* which was duly published. G. W. O. Howe wrote an editorial about it in the *Wireless Engineer* and he called it the "Tellegen Effect", but that didn't stick and it became known as wave interaction. It was V. A. Bailey in Australia, a man working on the ionosphere, who even-

tually explained exactly what was happening — the mean-free path of the electrons in the ionosphere was being influenced by an external field, the r.f. field, and so one transmission was modulated by the other.

**GARRATT** What did you do during the war, when there were Germans in the Nat Lab?

**TELLEGEN** Having worked on circuit problems in connection with radio receivers, I had become interested in circuit theory. We used to run courses for each other in the laboratory — I gave a course on circuit theory, one of my colleagues took notes and it was later reproduced and I eventually extended it into a book. The notes came into the hands of one of the professors of electrical engineering at Delft Technical University. He suggested to me that I should come to Delft and lecture in this field. As a result I became an Extraordinary Professor there in 1947 — Extraordinary in this sense means part-time. It was the policy of Delft University to have some specialists from industry to lecture on their particular specialities to the students in order to establish closer ties between the university and industry. This was how I first came to Delft and I continued there for twenty years.

**GARRATT** And you received a doctorate from Delft?

**TELLEGEN** I retired from Delft in 1968 and the following year I was awarded a doctorate in electrical engineering. Needless to say this pleased me very much.

### The Gyrator

**GARRATT** As though the pentode and the Luxembourg Effect were not enough for one man, you worked in several other areas?

**TELLEGEN** Yes, first of all on receiving tubes, then on complete receivers and later, on circuit theory which led to the gyrator.

**GARRATT** How did the concept of the gyrator arise?

**TELLEGEN** Around 1930 I was studying the properties of loudspeakers. The first equations concerning loudspeakers and telephones were developed by Poincaré. They are electro-mechanical systems, that can be described as systems with two ports — an electrical one and a mechanical one, which are related to each other by the telephone system. When you write down the equations something very curious happens depending on how you relate the electrical and mechanical quantities. If we say a current corresponds to a velocity and a voltage corresponds to a force and you write the two-port equations in these

quantities for an electromagnetic telephone, you get a set of equations which don't obey reciprocity, but if you do this for an electrostatic telephone the equations do obey reciprocity. Reciprocity, which is a property of a two-port, can be expressed as a mathematical relationship between the equations which link the current through one port to the voltage across the other port when the two-ports consist of standard circuit elements, namely resistors, capacitors and inductors.

Reciprocity is one of the important general properties in electrical networks — of course I was aware of these properties — but the telephone introduced equations which were of a different type. I realised that you could combine an electromagnetic and an electrostatic telephone in such a way that you have an electrical input and an electrical output; in principle, at least, you could imagine an electrical two-port which did not obey reciprocity. This led to the concept of two sets of equations; one pair defined an ideal transformer, which is a separate network element, the other pair defined another separate network element which I called a gyrator. In the transformer there is a proportionality between primary and secondary voltages and also between primary and secondary currents. In the gyrator there is a proportionality between primary voltage and secondary current and also between primary current and secondary voltage. The gyrator is thus defined but I had no means of realising it in the field in which I was working, i.e. radio receiver design. But I kept it in mind and during the war I pondered over the possibility of trying to find some means of actually realising it without mechanical means — I had been carrying the equations around for some years on a small piece of paper. Well eventually I found a means by using the gyromagnetic effects in ferrites and I carried out some experiments. When the war was over I wrote a paper about it — the time was then ripe for publication. During the war I had written a number of laboratory notes about the gyrator and I remember that when the Germans came around I collected all these notes and put them away so that they wouldn't get into Nazi hands.

The gyrator is connected in a way with a very famous book, *A Treatise on Natural Philosophy*, by Thomson and Tait, which was very well known among physicists at the end of the nineteenth century. It was Thomson and Tait's book which had everything about mechanics in it. It discussed vibrations of small amplitude which are described by linear equations. In these equations special terms may occur, called by the authors "gyroscopic" or "gyrostatic" terms because they occur when flywheels in a state of rapid rotation form part of the system by being mounted on frictionless bearings connected through a framework with

other parts of the system, and because they occur when the motion considered is motion of the given system relative to a rigid body revolving with a constrained constant angular velocity round a fixed axis. What I did was related to this and, as a result, I found a good name for the element — gyrator, the *gyra* from gyroscope and the final *tor* from many other electrical words like capacitor, inductor and transformer (Dutch). Gyrator is a practical word and can be used equally well in English, German and French — it's short and the *tor* implies that it is electrical. If you want to popularise a concept it is very important to have a good name for it. This did a lot to arouse interest in the concept of the gyrator.

**GARRATT** So you were the inventor of the gyrator as well as the pentode?

**TELLEGEN** It depends what you mean by inventor. The gyrator has been the subject of some patent applications. As I said before, I realised in 1930 that by combining electrostatic and electromagnetic loudspeakers you can, at least on paper, find pairs of equations that are non-reciprocal. I didn't publish it at the time; it was published by E. M. McMillan in the United States some years before my paper on the gyrator. So whether you call me the inventor is up to you, but certainly I invented the name.

**GARRATT** When was the gyrator first used practically?

**TELLEGEN** It was first put into practice by Bell Telephones. They realised before I did that you could use gyromagnetic effects in microwave technology — Casimir (one of the Directors of the Nat Lab at the time) was not very happy about this! I had missed the point, but then I was always thinking more about the frequencies used in ordinary radio receivers than about microwaves — these were not my special field. Perhaps it was stupid of me not to have looked into this. I had realised that the gyromagnetic effect could be very much greater at high frequencies than at lower ones, but they took this very much farther at Bell and introduced ferrites and waveguides and various other things related to gyrators. There are several things you can do with gyrators in waveguides and so introduce non-reciprocal effects.

Gyrators are now very popular and can be realised in an extremely efficient manner using solid-state components. This did not interest me at first because the gyrator itself is a passive element. With active networks the field is very much broader and, at first, there is very little reason to limit yourself to producing a passive element by active means. But it turns out that it is valuable even in active networks; for filters, for example, it has proved very useful.



# Sunspots, the ionosphere and h.f. propagation

Records of F2 layer critical frequencies, hour-by-hour over three years

by Kurt Feldmesser, Appleton Laboratory

IT must have been almost an international sigh of relief that greeted the improved ionospheric conditions last autumn, following the lean years of sunspot minimum. The charts reproduced here are a record of the F2 layer critical frequencies during those years. They illustrate the well-known behaviour of the ionosphere rather strikingly and allow a rapid mental picture to be formed of probable propagation conditions which could otherwise only be acquired by years of experience.

Before discussing details it is relevant to mention the source of the measurements. Appleton's experiments in this country and Breit and Tuve's in the USA, which established the presence of the ionized regions in the upper atmosphere, thanks to whose presence Marconi's Poldhu transmission had proved successful, are classics of radio history. Perhaps not so familiar is the fact that these measurements, begun routinely in 1931 at Slough by the Radio Division of the National Physical

Laboratory, have been carried on continuously under different administrations, supported by other observatories in other countries until well in excess of one hundred stations as well as various 'topside' sounding satellites monitoring the ionosphere.\*

For the routine measurements Breit and Tuve's pulse echo sounding method was adopted almost universally as providing more readily interpreted records. It is these measurements, mostly made hourly, that form the basis of the several ionospheric prediction systems in use.

In Fig. 1 the monthly median values of critical frequency, i.e., the highest frequency returned from a vertically incident signal, are plotted for the three regular reflecting layers E, F<sub>1</sub> and F<sub>2</sub>, at noon. The well-known dependence of

\*The data thus acquired are currently deposited at (and available from) World Data Centre C1 at the Appleton Laboratory in this country and from W.D.Cs. at Boulder in the U.S.A., Moscow in the U.S.S.R. and Tokyo, Japan.

the F2 layer ionization on sunspot activity is readily recognized. The seasonal variation of critical frequency seen in Fig. 1 and the "180° phase shift" between the F1 and F2 layers prompted the construction of another form of diagram which would allow the plotting of every single hourly observation.

Since measurements are normally made to 0.1 MHz accuracy for the F<sub>2</sub> layer critical frequency, a compromise had to be adopted, namely that the numerical values had to be quantized to the nearest MHz. In practice this has proved quite satisfactory since hour-to-hour changes are frequently in excess of this.

A graph of sunrise and sunset with time of day as ordinate and day number as abscissa looks like Fig. 2, where in addition to ground level sunrise and sunset, the times for these events at 100 km and at 300 km are also plotted. If we assume that a measurement of critical frequency is valid for plus and minus half an hour and assign a colour to each integral value of MHz, plotting to the same co-ordinates produces the diagrams in Fig. 3. The colour code used is — naturally — the resistance colour code, modified to allow values in excess of 10 MHz to be represented. (The modifications were, in fact, hardly needed until 1978.) Thus one may select any hour of any day between 1975 and 1977 and instantly locate the critical frequency that obtained at the time — over Slough! For communication within the British Isles this would also be the maximum usable frequency and radio operators might care to compare their log books.

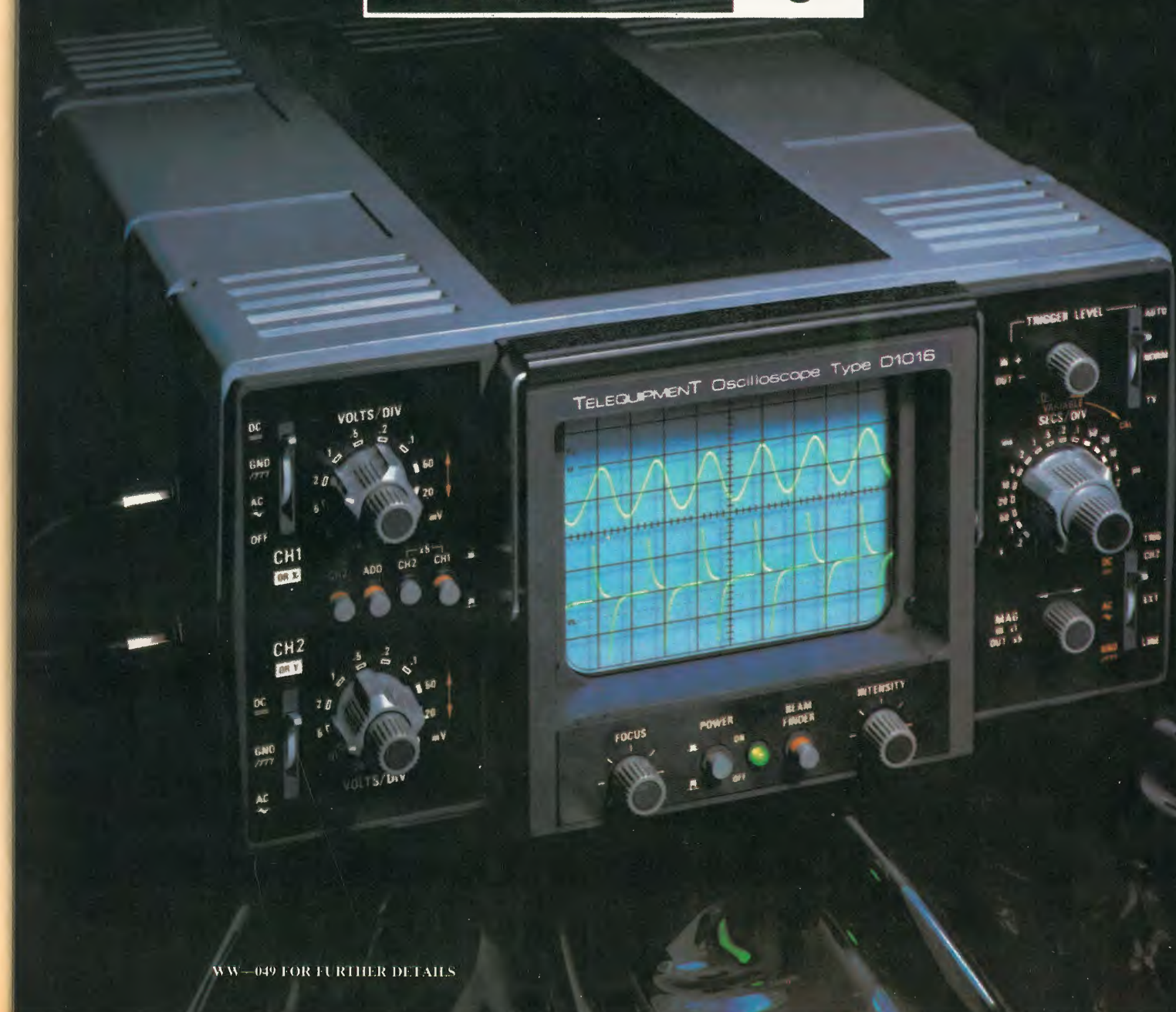
The aerial system at South Uist in the Outer Hebrides. Two rhombics are used, firing vertically, one covering the frequency range 250kHz to 4.5MHz and the other 4.5 to 20MHz. In the small hut, the equipment is a Swedish Magnetic AB ionosonde, Type 1005W, which consists of frequency-swept transmitter and receiver, marker generators and recording equipment, producing a 35-mm film record. Film is processed and analysed in South Uist by a local housewife, who also happens to be a geographer, and sent to Slough for correlation. One of the functions of the Hebridean station is to act as backup for ionospheric sounding rockets.



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# NOON VALUES OF CRITICAL FREQUENCIES MEASURED AT SLOUGH FROM 1932

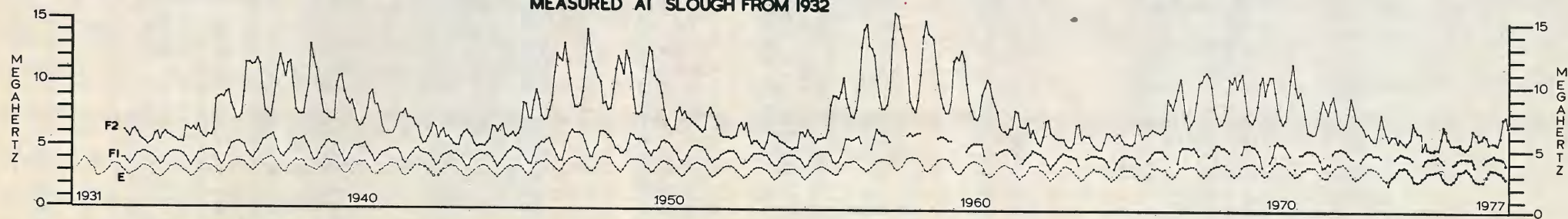


Fig. 1. Highest value of frequency returned (critical frequency). Large peaks in  $F_2$  layer frequencies correspond with the periods of sunspot maxima. Note inversion between  $F_1$  and  $F_2$  layers.

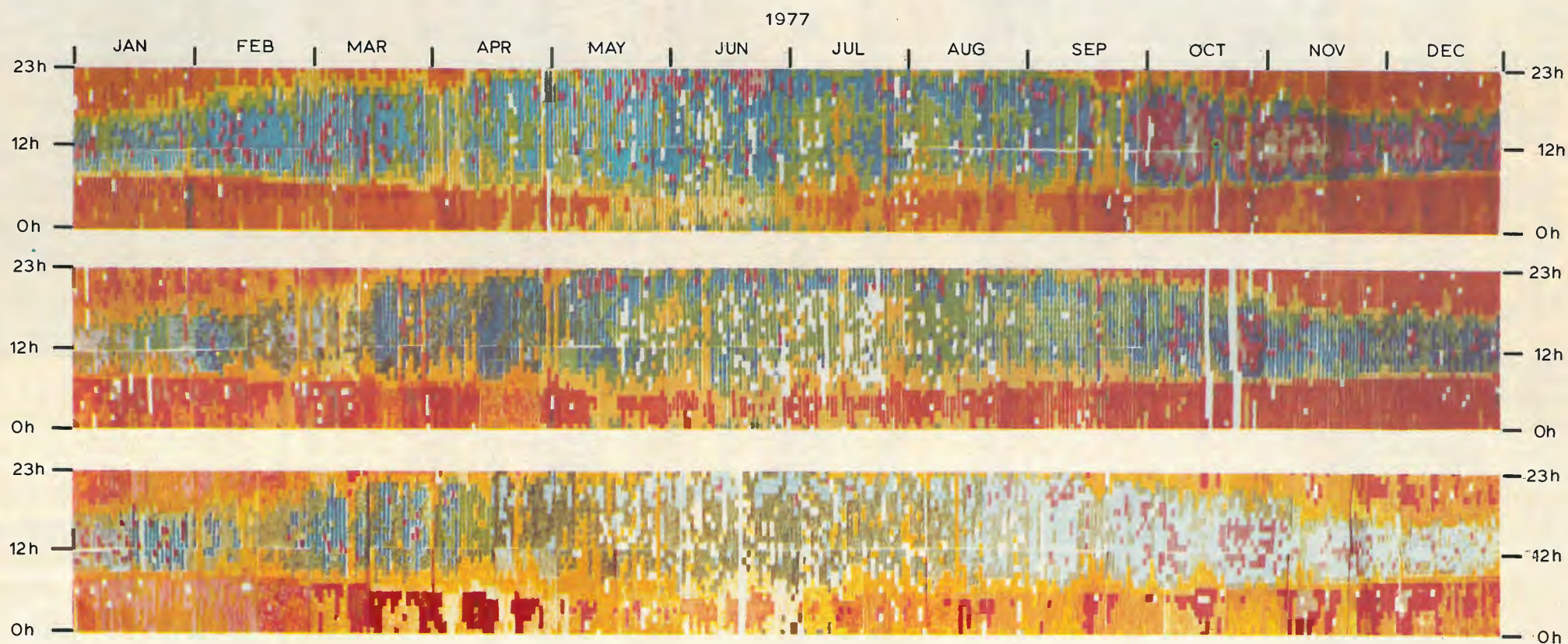


Fig. 3. Critical frequencies for each hour of 1975, 1976 and 1977. Colour code is the resistor code; e.g., brown is  $1\text{MHz} \pm 0.5\text{MHz}$ , orange  $3\text{MHz} \pm 0.5\text{MHz}$ . Note the similarity in shape between Figs. 2 and 3.

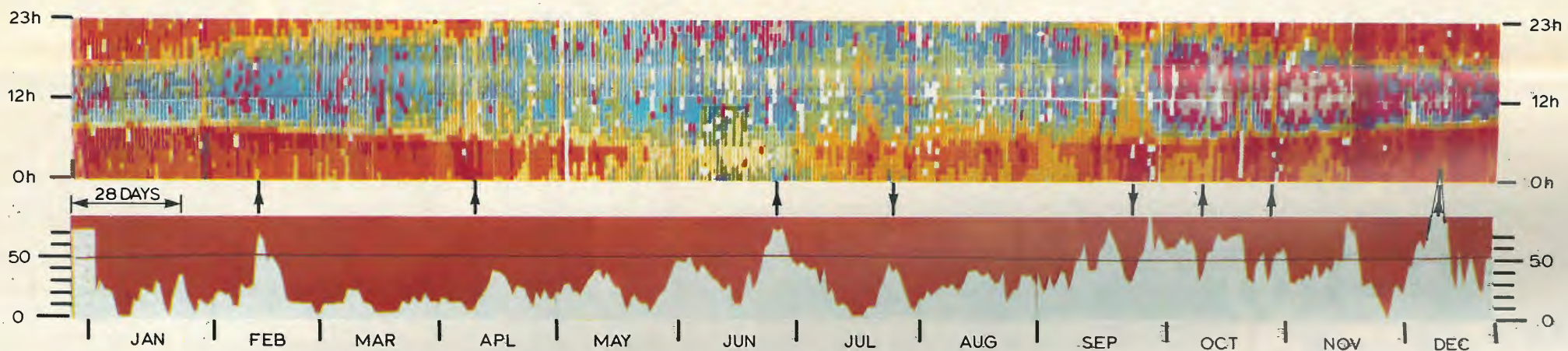


Fig. 4. This illustrates the correlation between sunspot number and critical frequency. Downward pointing arrows show apparent contradictions.

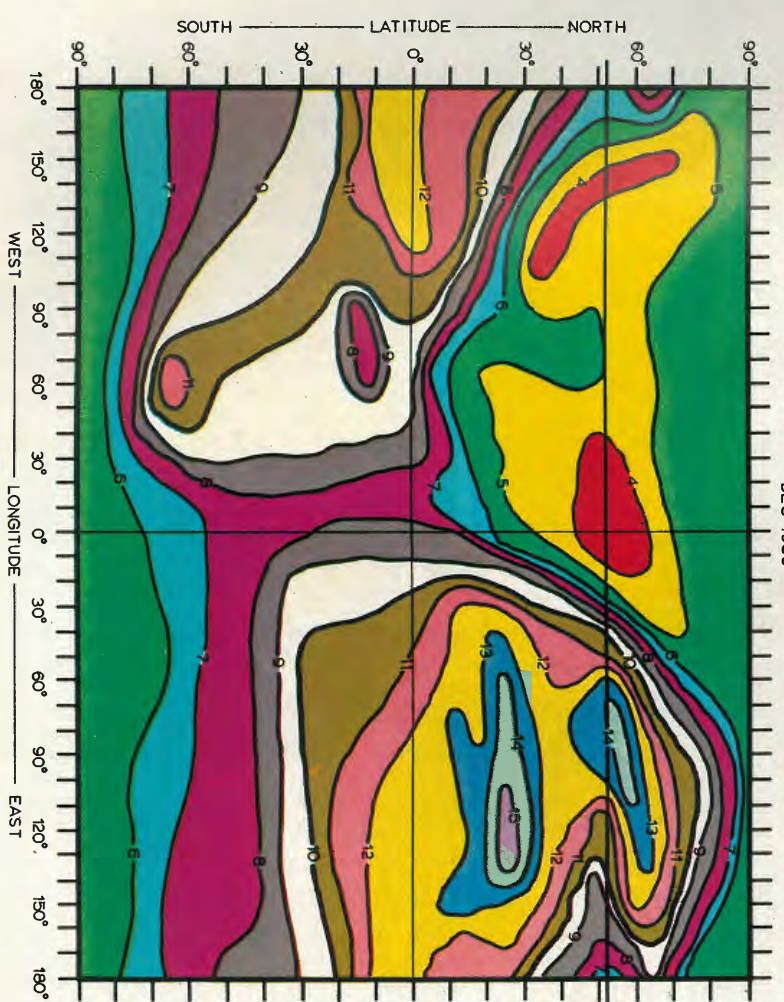


Fig. 5. Distribution of critical frequencies over the earth's surface, measured during the International Geophysical Year.

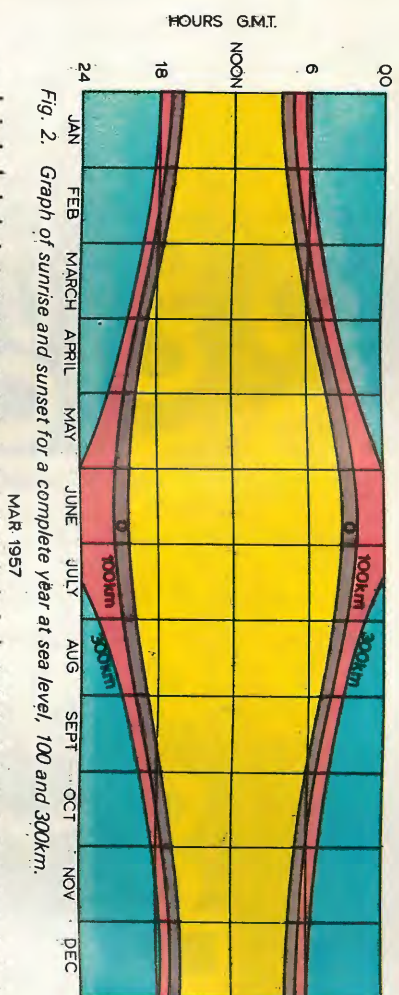
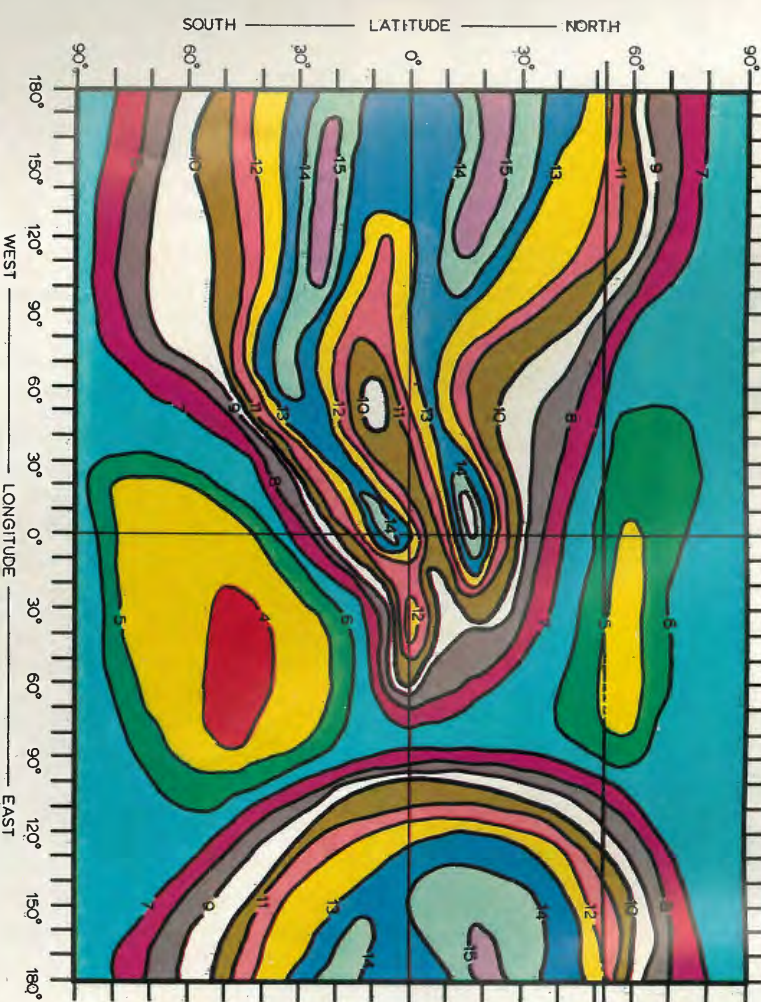


Fig. 2. Graph of sunrise and sunset for a complete year at sea level, 100 and 300 km.





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Sample plots have shown that very similar diagrams are obtained at 10° East when allowing for local time, although variations are already apparent in comparison with measurements made in Scotland, especially at night. Thus for control points due East and West one would expect to be able to form estimates of maximum usable frequency by making allowance for local time and for the shallower angle of incidence. The latter will give an increase up to three times the measured frequency depending on the launch angle and F2 layer height at the time.

In the diagram at Fig. 3, the hour of sunrise is readily recognized. One can assume as a good rule of thumb 4MHz as a typical value of critical frequency at that time, although, because of tilting in the layer anomalous propagation is expected. The same applies about one hour after sunset in winter. As the seasons advance the deionization time increases so that the 'centre of gravity' of the diagram is shifted towards the evening hours — a well known phenomenon!

As Fig. 2 shows, at 300 km the sun does not set for about six weeks in midsummer (over Slough) and partly because of this, high frequencies are maintained well past midnight. Throughout the year, the lowest critical frequencies are found just before dawn. Apart, that is, from days that don't fit the pattern. These are readily apparent from the diagram and for supporters of the 28-day recurrence theory a 28-day scale is shown in Fig. 3(c). A pair of compasses will yield ample evidence. The most common type of "abnormal" day is one in which the critical frequency never exceeds 4 or at most 5 MHz. Such days seem to be particularly common in the vicinity of the

## The author

Kurt Feldmesser came to this country in June 1939 after beginning his secondary education in Vienna, which he eventually completed in Cornwall. He interrupted a degree course in science to volunteer for the Royal Air Force as meteorologist. In 1948 he resumed studies in engineering at the then Plymouth and Devonport Technical College.

He began radio engineering commissioning Standard Telephones and Cables' Precision Approach Radar, and following two years with the General Electric Company's Stanmore Laboratories joined Marconi Instruments where he engineered the B.B.C.'s v.h.f. admittance bridge into production. He also patented

one form of digital display device whilst with the company.

A keen amateur cinematographer, he was chairman of the M.I. film society for several years. This interest led him to join an audio-visual company in London where he was responsible for the engineering of the first commercially produced language laboratories. Returning to radio engineering with Rank-Bush-Murphy he jointly patented a high-resolution pico-ampere generator at the Welwyn Garden City laboratory. For the past eleven years he has been with the Science Research Council and for the last seven he has been concerned with the ionospheric observations at Slough and South Uist in the Outer Hebrides.

equinoxes, though they can also be seen in midsummer. On those days sounder echoes are obtained from the regularly formed F1 layer, but the F2 layer electron density never exceeds that of the F1 layer and thus cannot be 'seen' by the sounder, nor is it available to carry radio traffic.

To put these diagrams in perspective, it is helpful to see the variation of the F2 layer critical frequency on a global scale (Fig. 5) — note the concentric subsolar contours and the 'tail' extending into the night-time zone. The charts themselves are section of the global maps for a single latitude (in the present case 51°30' N approximately). The fact that a substantial measure of 28-day recurrence could be found among days of lower electron density prompted the plotting of Fig. 4 where for 1977 the plot of the F2 layer critical frequency is aligned with a graph of daily sunspot number. Considerable correlation even on a day-to-day basis is readily

apparent. Of particular interest is the peak in critical frequencies in the second half of July (sunspot number near 40) and the trough in critical frequencies in the second half of September (sunspot number near 40). Critics of the correlation between sunspot number and critical frequency would find justification for their view in this, until we note that in the first instance we are looking at a short period maximum sunspot number and in the second instance a short period minimum. Thus a significant parameter would seem to be not only the absolute sunspot number but also the rate of change of sunspot number.

This article is published by permission of the Director of the Appleton Laboratory. The author would also like to thank his colleagues for much advice, freely given, and the Photographic Department for their annual painstaking work.

## Literature Received

**Test leads**, probes of various kinds including d.i.p. types, and the range of E-Z-Hook components in a catalogue from Special Products Division, British Central Electrical Co. Ltd, New Street, Ringwood, Hants .... WW401

**Colour television still picture recorder**, the Arvin Echo, described in a leaflet from Crow. Hundred frames per side of magnetic disc. Crow of Reading Ltd, P.O. Box 36, Reading, RG1 2NB ..... WW 402

**General electronic components**, materials, tools and equipment in Greenweld catalogue, which includes prices and quantity discounts for schools or private customers. Greenweld, 443 Millbrook Road, Southampton, SO1 0HX. Price 45p by post.

**Connectors** by Amphenol, in Circular Environmental range, to EL2112 Patt. 602. Def. Stan. 59-56 and Panavia Spec. 6432-1 in new

catalogue from Amphenol Ltd, Thanet Way, Whitstable, Kent CT5 3JF ..... WW 403

**Quarterly-updated stock list** from Crellon includes stock level and price of components from G.I.M., International Rectifier, Key-switch, Motorola, Plessey, RCA, Sprague and Waycom (WIMA capacitors). Crellon Electronics Ltd, 380 Bath Road, Slough, Berks ..... WW 404

**Fibre-optic systems** and kits from Belling-Lee, designed for computer peripherals, power control, etc., described in catalogue P.832C, obtainable from Publicity Department, Belling and Lee Ltd, Great Cambridge Road, Enfield, Middlesex ..... WW 405

**Programmable semiconductor devices** described and techniques of programming explained in booklet from Data I/O. Somewhat unrestrained title shortened to "Jungle book". Obtainable from Microsystem Services, Duke Street, High Wycombe, Bucks ..... WW 406

**Catalogue sections** from Bulgin now avail-

able on fuses and fuseholders, lampholders, code switches and digital displays. Sections available from A. F. Bulgin and Co. Ltd, Bypass Road, Barking, Essex IG11 0AZ ..... WW 407

**Add-on memory** for 158, 168, 3031 and 3032 c.p.us described in brochure from Storage Technology Corp. Maintenance panel which displays and diagnoses errors for up to 36 locations also described. STC, 2270 South 88 Street, Louisville, Colorado 80027, USA ..... WW 408

**Miniature magnetic circuit breakers**, intended for close protection, down to individual p.c. boards, described in catalogue from Highland Electronics Ltd, Highland House, 8 Old Steine, Brighton, Essex Sussex BN1 1EJ ..... WW 409

**R.f. and microwave components**, such as couplers, mixers, rotary joints, in 240-page catalogue by Sage Laboratories, with application information. Obtainable from R.E.L. Equipment and Components Ltd, Croft House, Bancroft, Hitchin, Herts SG5 1BU ..... WW 410



# A low-cost digital frequency meter — 2

Power supply and constructional details

by M. Tooley, B.A. and D. Whitfield, B.A.

The first part of this article described the main circuit of the digital frequency meter and included a list of the components required. This concluding part describes the power supply and the meter construction, and illustrates the printed-circuit board track and component layout details.

THE METER is designed to operate from a single unregulated d.c. power supply in the range 6.5 to 14V. In principle, it is possible to power all of the circuits from a single voltage rail but in practice a number of considerations (power dissipation, isolation, decoupling, etc.) lead to the use of two separate regulated voltage rails. Figure 7 shows the circuit diagram for the power supplies.

The e.c.l. circuits are supplied from a 5.6V rail, the regulator for which consists of a single-transistor, series regulator arrangement,  $D_5/Tr_7$ . The remainder of the meter is supplied from a 5V fixed regulator i.c., which incorporates current and thermal overload protection circuits. Overall protection of the meter against reverse polarity and voltage transients is provided by  $D_4$ .

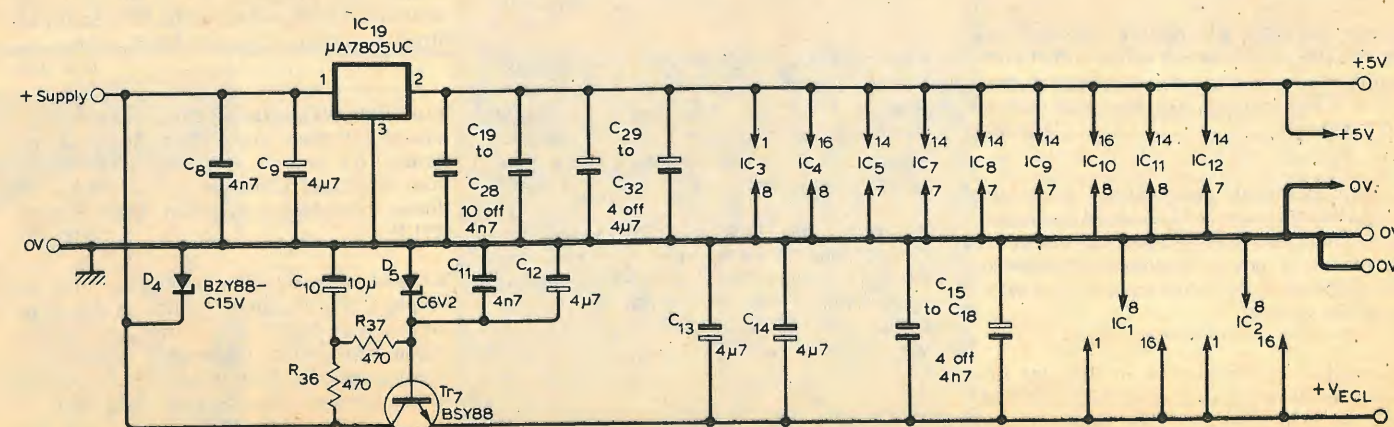
## Practical details

Construction of the digital frequency meter should not be attempted without the use of the recommended double-sided printed circuit board layout. The component side of the printed circuit board is employed as a common earth

plane and the other side of the board is printed. This not only simplifies construction and ensures unconditional stability but also allows the keen constructor to produce his own p.c.b. A true double-sided board would prove to be extremely difficult to make due to problems associated with the correct alignment of the opposite sides of the board.

The simplest method of producing the printed circuit board is with the use of double-sided copper clad board, one side of which is coated with a suitable positive photoresist. A transparent mask should be made, according to the p.c.b. layout given, and this laid over the coated side of the board and then exposed to ultra violet light. The other side of the board is completely masked using self adhesive p.v.c. tape and remains masked during the developing and etching of the track side. After cleaning and drilling the board, holes to non-earthed components and i.c. lead-outs are cleared to approximately 3mm diameter by means of a standard countersink drill. Component and i.c. leads to be earthed are soldered directly to the copper foil. The usual precautions should be observed when handling and soldering the c.m.o.s. devices. After completing the assembly of the board it is recommended that the component side be cleaned using a solvent cleanser and then treated with a light coating of protective lacquer spray.

Fig. 7. Power supply.



Adequate decoupling at all frequencies is essential for both supply rails and this is accomplished by the use of a mixture of miniature ceramic and tantalum capacitors. Test points are provided at the supply rails, buffered clock oscillator output, and at the output of the time standard divider — the waveforms at the latter two points being five volts peak to peak at 1MHz and 250Hz respectively. Further test points are provided in order to monitor the signal waveform after the input amplifier and at the output of the level translator (input frequency divided by four). The MC10231 is rated as having a typical maximum toggle rate of 225MHz, and is pin-compatible with the MC10131 which has a typical maximum toggle rate of 160MHz for approximately half the cost of the faster device. Tests indicate that the MC10131 is capable of satisfactory operation up to approximately 180MHz.

Provision has been made in the p.c.b. layout for the display to be wired for use with right- or left-hand decimal point indicators. In normal use a DL707 is used for  $IC_{16}$  and its right hand decimal point is returned to 0V by means of a 220Ω resistor in order to produce a read-out in MHz. If indicators having left hand decimal points are used, then the decimal point of  $IC_{15}$  is returned to 0V by means of a 220Ω resistor. The decimal point wiring may, of course, be completely ignored if a read-out in kHz is desired. For full six-digit display operation it would be necessary to replace  $IC_{12}$  by a decade counter and a

suitable latch/decoder, such as those listed for  $IC_9$  and  $IC_{10}$ . If this modification is incorporated, it should be noted that  $IC_{13}$  should be altered to the same type of display as  $IC_{13}$ .

Since the d.c. supply voltage input may take any value between 6.5V and 13.5V, adequate provision for heat sinking the regulators must be included. The 5V regulator,  $IC_{19}$ , makes use of the copper earth plane and this is adequate for supply voltages up to 12V. Above this value, the dissipation of the device should be increased by means of a simple vaned heat sink of around 20°C/W.  $Tr_7$  requires a push-fit heat sink of 85°C/W for inputs not exceeding 10V, and 50°C/W above this value.

It has been found that some types of crystal require a small value capacitor, around 22pF to 47pF, connected in series with the crystal in order to allow the oscillator frequency to be adjusted to exactly 1MHz. The total cost of all components required for the digital frequency meter is around £40, inclusive of v.a.t. and a single source for all the semiconductors is Semiconductor Supplies (Croydon) Ltd.

## Performance

The performance of a prototype meter with an MC10231 fast prescaler i.c. was measured using an 8V regulated power supply, a Racal type 365A signal generator, a Racal type 835 digital frequency counter and a Hewlett-Packard oscilloscope type 1722B with two model 10017A 10:1 probes. The oscilloscope was used in its unrestricted bandwidth (275MHz) mode.

Figure 10 shows a graph of the minimum input voltage to the counter against frequency for satisfactory triggering over the range 1MHz to 220MHz. Waveforms for the operation of the input stages are shown for 12MHz in Figs. 11 & 12 and for 100MHz in Figs. 13 & 14. In all cases the upper trace is the input signal waveform and the lower trace is either the output of the input amplifier at test point 1 (Figs. 11 & 13) or the output of the level translator at test point 2 (Figs. 12 & 14).

In operation, the counter draws a maximum supply current of approximately 650mA, with a standby (zero display) drain of approximately 400mA.

## Acknowledgement

The authors would like to thank their colleagues for all the help and encouragement received during the development of the meter described in this article. Particular thanks are due in this respect to Mr T. Y. Chan, Mr I. A. Laurence and Mr C. P. Orwell of EMI Medical Limited, and to Mr A. Stables of Brooklands Technical College.

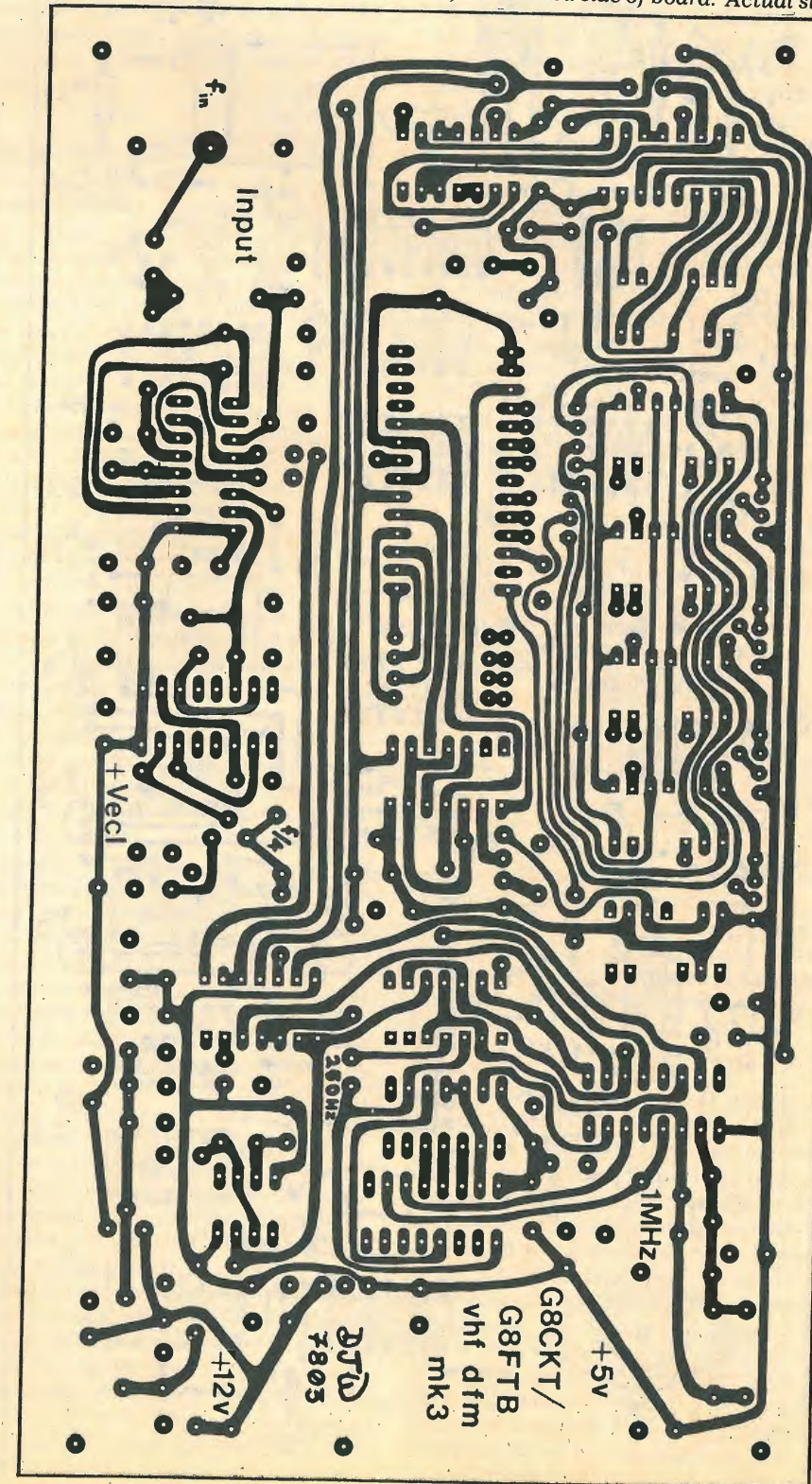
Printed circuit board see Part 1.

## The authors

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UK FM Group (London), for over 3½ years. David Whitfield is a senior development engineer for EMI Medical Ltd at Hayes in Middlesex. He obtained his engineering degree (BA) at Cambridge and has, since this article was commissioned, had a medical electronics and physics degree (M.Sc., London) conferred on him. His interests include amateur radio (G8FTB), constructional projects, photography, and hi-fi. Both authors are also involved in youth work, as scout leaders.

Fig. 8. P.c.b. copper track pattern viewed from etched side of board. Actual size.





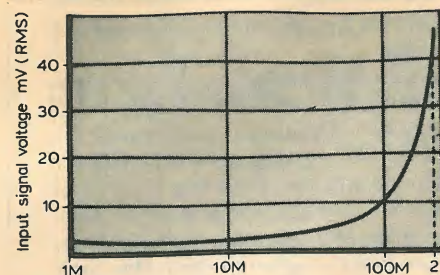


Fig. 10. Input sensitivity of frequency meter shown as a function of frequency.

Fig. 9. P.c.b. component layout viewed from component side (earth-plane side). Crosses indicate connections to earth plane and all others are to have clearance through earth plane. Approximately actual size.

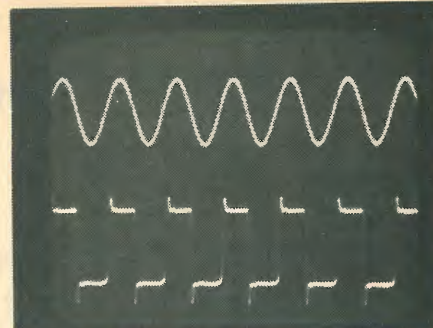
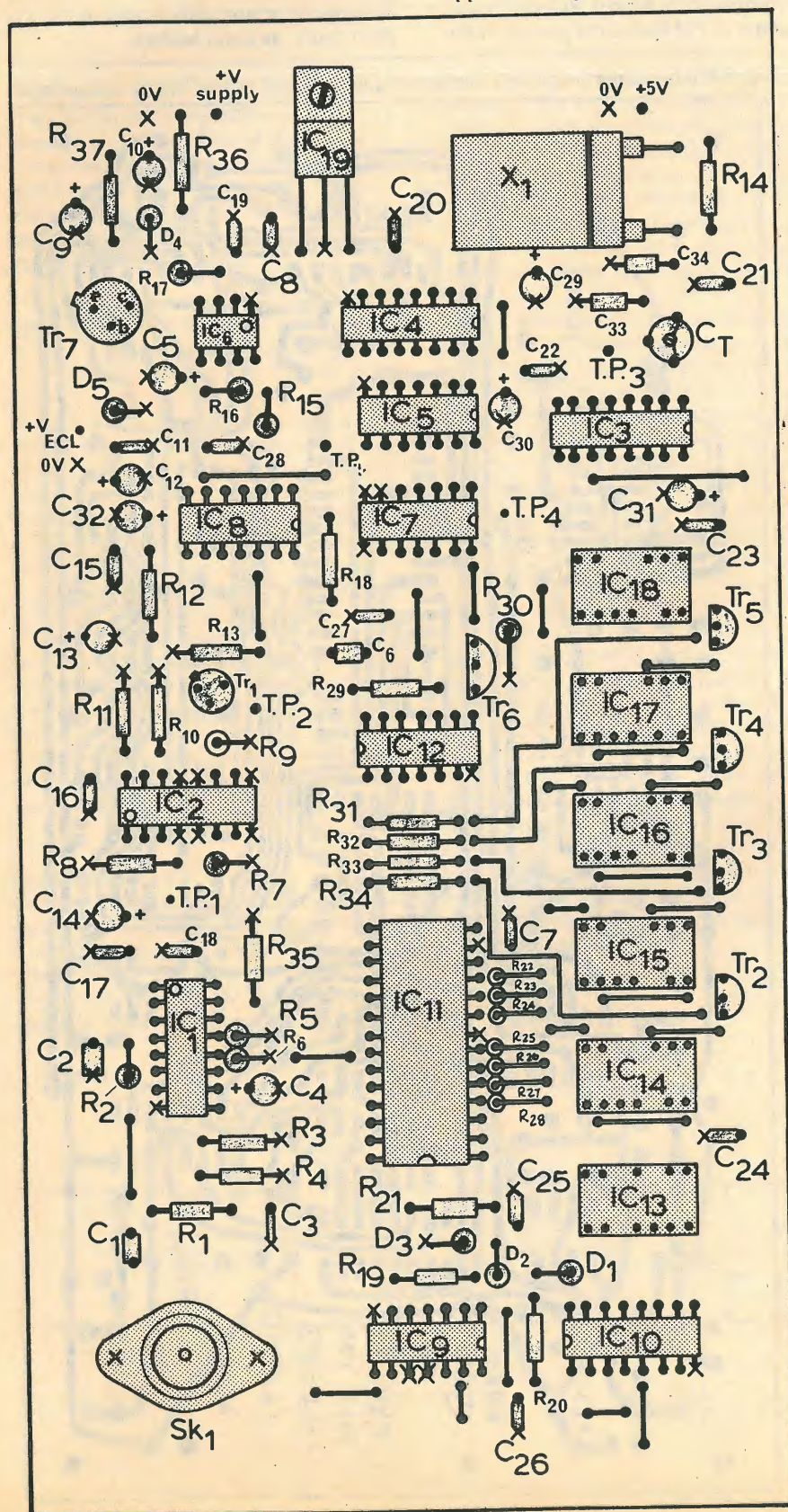


Fig. 11. 12MHz oscilloscope traces: upper, input signal at SK1; lower, output of input amplifier at test point 1. 50ns/div. 100mV/div. (upper) and 500mV/div. (lower).

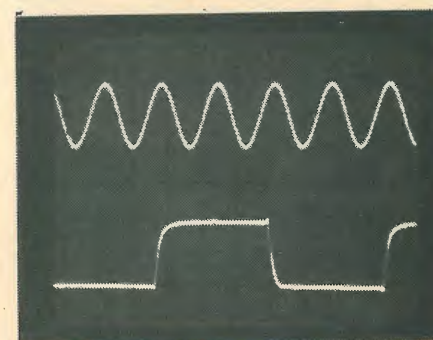


Fig. 12. 12MHz oscilloscope traces: upper, input signal at SK1; lower, output of level translator at test point 2. 50ns/div. 100mV/div. (upper) and 2V/div. (lower).

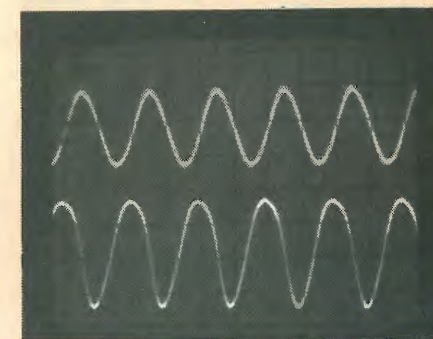


Fig. 13. 100MHz oscilloscope traces: upper, input signal at SK1; lower, output of input amplifier at test point 1. 5ns/div. 100mV/div. (upper) and 500mV/div. (lower).

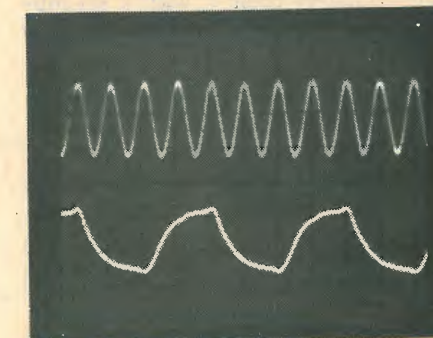


Fig. 14. 100MHz oscilloscope traces: upper, input signal at SK1; lower, output of level translator at test point 2. 10ns/div. 100mV/div. (upper) and 2V/div. (lower).

# The technique of technical writing

Its value as an aid to thought

by S. W. Amos, B.Sc., M.I.E.E.

To some technical workers the task of writing reports is a chore more to be endured than enjoyed. Such workers may claim, for example, that their skill is in electronic design not literary composition and that their time is better devoted to the next design than to writing up the previous one. This attitude to technical writing is understandable but it shows a lack of appreciation of the purpose and value of technical reports and ignores the advantages that can be obtained from writing them. The mental discipline (chiefly the need for logical thought) essential in successful writing can help in design work and usually improves understanding of the subject. In this article the author gives some useful information on the processes involved in technical writing and hopes to show that it can offer challenges as exciting and rewards as satisfying as those of electronic design.

THERE ARE many possible reasons for a lack of enthusiasm for technical writing. For example, technical staff may feel that they have been asked for a literary masterpiece and that writing is not their speciality. The mistake here is to suppose that technical literature is read for the beauty of its language. It serves, in fact, a strictly utilitarian purpose — to provide technical information as clearly and concisely as possible. The facts in technical literature are more important than the way in which they are presented i.e. the content is more important than the style.

In great literature, on the other hand, the most significant quality is the beauty of the language i.e. the style is more important than the factual content. For example the story of Macbeth can be told in a few pages but without Shakespeare's English it would make dull and pointless reading. The distinction between the two types of literature stems from a difference in purpose: technical literature is intended to inform; the Shakespeare play is for entertainment (a view unlikely to be shared by young schoolboys!).

Although content is the most important factor in technical literature, style must not be dismissed as of minor concern. On the contrary style matters considerably because it determines the readability and thus how easy it is to understand the technical content. Much

of this article is concerned with style and it is probably true that more of the shortcomings in modern technical literature stem from errors in style than from poor selection or arrangement of facts.

## The specification

A second reason which may cause engineers difficulty in technical writing is the need for a specification. An engineer asked to design an amplifier cannot begin work until he knows what kind of amplifier is required: he needs information on, for example, the gain, output power, load resistance, frequency range and permissible distortion. Such a detailed specification is the starting point from which design begins. Yet the need for a specification for a piece of technical writing is often overlooked or insufficiently appreciated. In fact an engineer's instructions may simply amount to a request for a report! It is true that some information on the type of technical content and style may be available from reports on previous designs but this is an unsatisfactory and secondhand way of deducing the specification.

A specification for a piece of technical writing is as necessary as one for a piece of technical equipment and should include information on a number of factors. Obviously the subject matter must be made clear and this must be defined with precision. If certain aspects of the subject need not be included, this must be stated. Equally important if not so obvious is information on the readers. For whom is the literature intended? Who needs the information which has been so carefully defined? What are the general technical backgrounds of the readers? Information on this is vital because it determines the technical standard and the form of presentation to adopt in the writing. The author also needs to know the purpose of the literature. Often it is simply the provision of technical information but there are other possibilities: for example a research report may compare two methods of solving a technical problem and may be required to decide and state which is the better. But the most important questions which the specification must answer are usually "who wants to read this literature and why?"

The questions are easy to answer if, for example, one is writing a textbook on a specific subject and aimed at students of a particular level of academic standard. Here the subject matter is laid down in a syllabus and the readers' background is known. Similarly a research report published in the proceedings of a learned institution is usually intended for readers with a background and technical standard similar to those of the author. Technical writing is particularly difficult where the age, experience and technical standard of the readers covers a wide range. A good example of such a piece of writing is this article! The only thing we know for certain about *Wireless World* readers is that they are interested in electronics. It would therefore be a good idea to give an electronic flavour to all the examples in the article.

Vagueness of purpose is another difficulty which sometimes plagues the technical author. An example arises in the accounts of laboratory experiments which students at college or university are required to write. For whom are these accounts intended? Clearly not the instructional staff for they will learn nothing from them nor the students because they are unlikely to open their laboratory notebooks after leaving the college. No-one else is likely to read the reports. Perhaps the best that can be said is that preparing these reports gives the students some practice in technical writing and in keeping records of work done. These reports are in the same category as students' lecture notes which are usually ignored when the relevant examination has been passed.

## Structure

Once the specification for a piece of technical writing is known the author can begin collecting and selecting the facts to be included. It can take time to amass a relevant and comprehensive set of facts because to find them the author may need to consult technical books, articles, reports, specifications or other literature, to talk to experts, to build equipment or to carry out experiments. When the facts are ready they must be classified i.e. those relating to a particular aspect of the subject must be gathered together and the groups so obtained must be arranged to give a smooth flow of information. There are



many ways of doing this. The obvious method is to put the ideas down on paper and to build up the required order by insertions, deletions and transpositions. In doing this the draft usually becomes untidy and a number of rewrites (or retypes) are needed before a satisfactory arrangement is achieved. A second method is to write the facts (in the form of brief notes) on cards which can be shuffled to give a suitable order.

Whilst the author is concentrating on achieving a satisfactory arrangement of technical content (sometimes called structure) he cannot normally pay much attention to details of style but this does not matter because style can be tidied up afterwards. The most satisfactory structure is normally one which has a logical step-by-step progression of ideas and the search for such an order can have a number of useful consequences:

- Any gaps in the sequence of ideas become obvious and to fill these the author may need to renew his fact-finding research to discover the information needed to complete his story.
- The division of the subject into sections, subsections and sub-subsections usually becomes clear and this helps considerably in choosing appropriate headings and sub-headings.
- The search for suitable groupings and arrangements of facts or for suitable headings sometimes produces delightful surprises: This illumination of dark corners of knowledge is one of the rewards of technical writing.

Some authors divide their work into five or six generations of heading. This is useful in clarifying and classifying the author's thoughts but it is unwise to include so many generations in the final draft because they would be confusing to the reader and would make cross-references complex and difficult to find. As a general rule it is better to use no more than three generations and the number can usually be reduced to one by repeating one or more key words. In the example illustrated below two generations are reduced to one by repeating the words 'feedback' and 'derived':

#### Derivation of feedback

By series connection  
By parallel connection

can be replaced by

#### Series-derived feedback

#### Parallel-derived feedback

Once the structure of a piece of technical writing has been settled the author may be tempted to think that the major part of his job is done. The setting down of ideas in words is often regarded as a straightforward process calling for no great skill and thus insufficient attention is paid to this aspect of the task. An author may argue that any reader who finds his work difficult to understand should concentrate more because all the

information is present and is, moreover, arranged in logical order. This is a comforting philosophy for the author because it shifts to the reader the responsibility for extracting the meaning from a difficult text. It is, however, the author's responsibility to see that his work is easy for the intended reader to understand and this usually requires that as much attention is paid to the style as to the selection and arrangement of facts. The author should not forget that his words may remain in print for the rest of time as a monument to his success or failure.

#### Style

Style usually implies such considerations as choice of words and sentence length. Certainly these are two important factors which contribute to the readability of text. But in technical literature text is not the only means of conveying ideas: extensive use is made of other aids such as illustrations (photographs, circuit and block diagrams, graphs), lists, tables, mathematics, calculations and any other means at our disposal for recording ideas on a piece of paper. All these factors are aspects of style and frequently a given series of facts can be presented in many different ways. The author has to decide which form of presentation to use and his choice depends primarily on the type of subject matter and on the reader's knowledge and background.

To quote a simple example, many *Wireless World* articles are concerned with electronic circuits and detailed information on circuitry is always given by circuit diagrams. This is the natural way of conveying such information and is the method the reader expects. It is a most powerful way of presenting circuit information: anyone who has attempted to describe a circuit diagram in words (e.g. over the telephone) will know how difficult and time-consuming a task this is. Because the circuit diagram conveys most of the information required by the reader, there is no need to describe the circuit in words also. It is useful to point out any unusual or new aspects of the circuit but the author's and reader's time is wasted by duplicating in words information which is obvious from the diagram.

As another example, in a contribution to the proceedings of a learned society mathematical presentation is generally acceptable. Here again equations are a shorthand way of expressing ideas which are sometimes inexplicable in words. But in an article on the same subject to technicians, an author would probably present the facts in the form of graphs or abacs: the technician is usually more interested in the applications of ideas rather than their derivation.

"A good illustration is worth 500 words". "We like to have at least one diagram per page." Advice such as this is of little value. To what extent diagrams should be used depends on the

subject matter and on the reader's background. Articles on electronic circuitry naturally abound in circuit diagrams. In fact experiments are being carried out at the moment to see if it is possible in literature on electronic equipment to dispense with conventional text and to give explanations which require words in little boxes within the circuit diagrams: some examples of such diagrams were given in *Wireless World* for October 1973. This is an example of one extreme where technical writing consists almost entirely of circuit diagrams. At the other extreme there are some technical subjects where there is no need for diagrams at all: this article is an example.

Another recommendation which is in practice impossible to follow strictly is that each paragraph should embody one idea. Some ideas, it is true, can be dealt with satisfactorily in a few lines, a convenient length for a paragraph. Other ideas require a page or more of description and it is a little offputting to a reader to see a solid page of text unrelieved by headings or paragraph breaks. So in practice breaks are deliberately introduced at intervals down the page. Where you pause to take breath, start a new paragraph!

There is a popular misconception to the effect that writing, to be authoritative, must be pompous, legal-sounding and liberally sprinkled with long words preferably of Latin extraction. A third reason why engineers may not enjoy technical writing could be that they feel that they cannot write this kind of high-flown English. There is no need: to carry conviction writing should be simple and direct. This need stems directly from the main purpose of technical writing — to give information. It should be possible for the reader to obtain the information he wants quickly and accurately: there should be no need to wrestle with involved syntax, puzzle over vague phrases or to refer to the dictionary. The greatest boon authors can confer on their readers is to present them with a clear and simple story which tells them accurately and unambiguously what they want to know. Long sentences liberally endowed with dependent clauses are better broken up into simpler sentences. Long words of Latin derivation so beloved by politicians and the Civil Service are better replaced by shorter Anglo-Saxon equivalents. Wordy phrases (such as "in the majority of instances") should be reduced to a single word ("usually") or omitted altogether. These common errors of style are enumerated in greater details in the appendix to this article which also gives preferred alternatives.

There is a simple way of avoiding these errors. Examine the words in every sentence one by one and ask yourself whether each contributes anything useful to the meaning. If not, throw them out. It is true that a ruthless application of this policy may lead to a concentrated staccato style which

makes reading slow and indigestible but this difficulty can readily be overcome by inserting at the beginnings of sentences words such as 'but', 'however', 'nevertheless' etc. to link the sentences together. This does not contradict the advice quoted earlier that each word should pull its weight: the promotion of smoothness in reading is a worthy purpose for any word.

The rules for ensuring good style can be summarised thus:

- See that every word is contributing usefully to the meaning
- Choose the simplest word
- Prefer sense to elegance.

#### Spoken and written English

These rules for good style apply only to written English. The effect of applying them to spoken English would be disastrous — making many of us speechless for long periods! In speaking we continually use vague and meaningless phrases to give us time to think out what we are going to say next. Speech is therefore strewn with 'Be that as it may', 'As a matter of fact' etc which perform a useful service in promoting continuity of speech (although the best speakers do not use them). But these phrases have no place in written English where they dilute the facts and make reading unnecessarily long and tedious. Repetitions, too, are commoner in spoken than in written English. A teacher or lecturer punctuates his lectures or lessons with statements of basic principles, summaries of what has been said and information on forthcoming attractions. This is good teaching practice but frequent repetitions are unnecessary in a book or article: the reader can also turn the pages forward or backward if necessary.

#### Degrees of perfection

No example of technical writing is ever perfect. It is always possible to improve the readability or clarity, and the degree of perfection achieved depends on how many times the author is prepared or allowed to rewrite the text to improve the style. Each new version is better than the previous but the improvement becomes less obvious with each successive rewrite. This is an example of an asymptote (Fig. 1): the curve approaches 100% (perfection) as new versions are prepared but perfection is never achieved. This is why it is always possible to suggest improvements to any piece of technical writing. This does not mean that the piece was badly written but that the author did not devote infinite time to its preparation. This is a good point for the author of this article to make because it provides the answer to any critic who feels moved to write suggesting improvements! This point can be summarised by saying that smoothness rarely arises from inspiration: it is more usually the result of perspiration.

At what point then does one give up

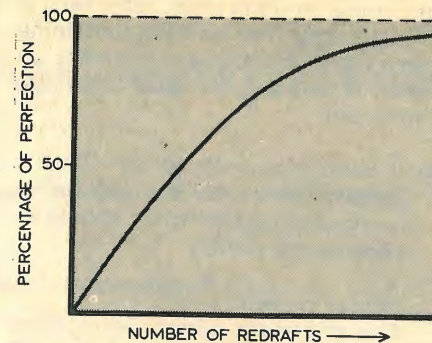


Fig. 1. The style of a draft improves with every redraft but perfection is never reached.

polishing? This is often decided on economic grounds e.g. two articles each 70% perfect are better value than one 95% perfect which might take the same time to prepare.

As mentioned earlier it is easy to criticise style. Indeed when comments are invited on drafts it is usually small stylistic alterations which are offered. These may be useful in improving readability and perhaps clarity but are not so valuable as comments on the technical content. Doubtless this is because there is no need to be an expert on the subject to be able to comment on style but a good knowledge of the technology and a detailed study of the draft are necessary before structural comments can be offered.

#### Accuracy and clarity

Several times in this article we have implied that clarity and readability are

vague phrase	preferred alternative
at this point in time	now
in the vast majority of instances	usually
in the event that	if
owing to the fact that	because
according as to whether or not	if
a considerable amount (or number) of	most
referring to the diagram it can be seen that	the diagram shows
in the first place	firstly
from the practical point of view	in practice
when it comes to	in
it should be noted at this junction	(omit)
be that as it may	
in point of fact	
as a matter of fact	

#### Overuse of impersonal phrases

Continual use of phrases such as "it can be shown that . . ." "it is important to note that . . ." make text dull and

monotonous. Most can easily be avoided (and a saving of words achieved) as the following examples show.

#### impersonal phrase construction

It is essential that the receiver be retuned after each adjustment

#### possible alternative

retune the receiver after each adjustment.

It will be noted that, as far as the carrier frequency is concerned, the circuit is basically a bridge.

At the carrier frequency the circuit has the form of a bridge.

important aspects of technical literature. But factual accuracy is also important and circumstances can arise where the interests of clarity and accuracy clash. Which is then sacrificed to the other? The answer depends on the subject matter and on the technical standard of the writing. In an elementary textbook, for example, it is almost impossible to make any statement without calling to mind a number of exceptions or important conditions which ought to be mentioned. To include all these would probably not help the young reader and would certainly rob the statement of much of its punch. In such writing it is usually better to forget the qualifications and so sacrifice accuracy to achieve clarity. But in describing, say, the sequence of starting-up operations for a high-power transmitter where an error could lead to delay or even to damage of the equipment, technical accuracy is clearly essential and, if necessary, clarity should be sacrificed in its interests.

#### APPENDIX: COMMON ERRORS OF STYLE

##### Use of vague phrases

Perhaps because authors are paid according to the length of their contributions or because they write as they speak, there is a tendency to use unnecessary words and phrases. Such verbal padding dilutes the meaning and is best avoided. Some of the commonly-used phrases, with preferred alternatives, are given below: this list is by no means complete.



### Use of complex syntax

It takes time for readers to appreciate the grammatical structure of a sentence with a number of dependent clauses and this delays understanding of the meaning. Moreover there is a danger that a clause, if not carefully placed in the sentence, may be taken as qualifying the wrong word. Such sentences should preferably be recast as a number of simpler sentences.

### Unnecessary use of long words of Latin extraction

As mentioned earlier the use of long words is sometimes believed to give written work an air of authority. This is quite wrong: the equivalent Anglo-Saxon words are usually shorter and have more impact. The following are few examples.

#### Latin-based word

initiate  
terminate\*  
visualise  
facilitate  
necessitate  
elucidate  
utilise

#### simpler equivalent

begin, start  
end, finish  
think, picture  
aid, help  
need  
explain  
use

### Unnecessary use of abstract nouns

One of the common ways of using more words than is necessary to convey the meaning is to introduce an abstract noun as in the following examples.

#### wordy phrase

incurs a heavy  
cost penalty  
in the low-  
loss category

#### simpler equivalent

is expensive  
is of low loss

### Use of 'former', 'latter' and 'respectively'

These words cause the reader to look at the previous sentence to find out what words are meant. This distraction interrupts the flow of the argument. Often 'respectively' can be omitted without risk of confusion. If, however, there is any danger of misunderstanding it is better to repeat the key words.

### Punctuation

Anything which is likely to distract the reader should be avoided. Complex constructions which rely on punctuation marks to convey the desired meaning are better recast as simpler sentences. But hyphens linking words in adjectival phrases can improve clarity e.g. a "high-value low-loss capacitor" is easier to read than a "high value low loss capacitor."

To conclude this appendix here are some examples taken from textbooks

\* But a transmission line is terminated: 'ended' would be wrong here.

containing errors in style. After each is given an error-free version (containing approximately half the number of words) which is easier and clearer to understand.

- (a) It is very difficult to lay down a precise value for the minimum receiver passband which should be allowed. (19 words)

#### Recast version

The minimum allowable receiver passband is difficult to assess. (9 words)

- (b) Although it is generally accepted practice to lay the earth plates in a regular pattern, cognizance must be taken of the fact that the earth resistivity may vary over a given area... (32 words)

#### Recast version

Earth plates are normally laid in a regular pattern but the earth resistivity may vary over a given area... (19 words)

- (c) Although there does not appear to be much to choose between silicon and germanium transistors as far as electrical characteristics are concerned... (23 words)

#### Recast version

The electrical characteristics of silicon and germanium transistors are similar... (10 words)

### Further reading

Readers seeking further information on the presentation of technical information may find the following publications useful:

P. Wright, "Presenting Technical Information: A Survey of Research Findings", Medical Research Council, Applied Psychology Unit, Cambridge, CB2 2EF.

H. J. Uchy, "Effective Writing for Engineers, Managers, Scientists" John Wiley, New York.

## More letters

### MICROCOMPUTER BUSES

Some months ago, in quest for a suitable universal bus for a computer system, having already rejected S100 as primitive and incompatible (even with itself!), I came across reference to such a bus designed in this country, namely E78, outlined by Mr Aylward in December 1978 letters. Further investigation revealed a very active committee working on the bus specification and they were only too pleased to give detailed information about it.

E78 has been designed with all current and future microprocessors in mind, and appears fully compatible with 8-to-16-bit systems, to the extent that a system running as an 8-bit system can be instantly upgraded to a 16-bit system, just by substituting the 8-bit c.p.u. card for a 16-bit c.p.u. card plus the relevant software. The specification includes descriptions of all bus signals and their timing, and great care has gone into important areas such as power requirements, transmission line effects and future expansion possibilities. The specification supports single or double Eurocards using one or two indirect connectors, depending on system complexity.

I am a little alarmed to see that another bus has been proposed by Mr Borril (October 1978 letters). I only hope that these two groups can get together quickly on amicable terms, because I and several others have already committed designs to paper, using the E78 system, and will be manufacturing equipment in the near future.

It seems the fate of British designers to suffer indecisions on standards, particularly at a time when to delay may mean lost business and the perpetuation of inferior standards such as S100!

M. A. J. Hutchings  
Chatham  
Kent

### NORWEGIAN 2M REPEATER

As mentioned in "World of amateur radio" in the December 1978 issue, we would like to inform you that LA6ER is not a beacon, but a 2-metre repeater transmitting/receiving on the European repeater channel R5. Power to charge the batteries comes from a solar panel and from a small windmill.

We are enclosing full technical information of the equipment, which we believe is the first repeater at least in Scandinavia to be powered by the sun and wind. [Readers interested may obtain photo-copies from this office by sending a stamped, addressed envelope — Ed.]

The repeater is well known to British amateurs and can easily be started with just a little bit of "lift" by stations in Scotland.

O. Lindberg, LA5WT  
President NRRL group.  
Egersund  
Norway

# Audio power amplifier design — 6

More on negative feedback and non-linearity distortion — a continuation of part 5

by Peter J. Baxandall, B.Sc.(Eng.), F.I.E.E., F.I.E.R.E.

Part 5 (December issue) discussed the theory of non-linearity distortion in an ideal feedback amplifier having a parabolic forward transfer characteristic. Attention is now turned to distortion in circuits using ordinary junction transistors, having exponential transfer characteristics. The concept of "inverse distortion" is introduced, leading to a useful distortion theorem.

THE CIRCUIT USED for obtaining the experimental results presented below is shown in Fig. 1 and is the same as for the f.e.t. tests in Part 5, except for two small modifications. The 1nF capacitor was found necessary to prevent high-frequency oscillation when full feedback was applied, and the resistive attenuator in the base circuit was added to reduce loop gain, for convenience, to a similar range of values to that applying to the f.e.t. version of the circuit. The measured current gain ( $\beta_{dc}$  or  $h_{FE}$  of the transistor used was 580 at an  $I_{dc}$  of 1.45mA.

Throughout the measurements the fundamental output voltage was kept constant at three volts peak, corresponding to a ratio of peak signal current to direct working current of 0.647 — the same conditions as for the f.e.t. tests in Part 5. The results are shown by the full-line curves in Fig. 2, and exhibit some fascinating features when compared with the earlier f.e.t. results. A great deal of thought, both of a formally analytical and also of a more intuitive type, has been devoted to trying to understand these features, and considerable enlightenment has resulted.

A junction transistor has the great virtue, at sufficiently low values of collector current, that it follows in practice, with high accuracy, the relationship

$$I_c = I_o \exp \frac{qV_{be}}{kT} \quad (1)$$

where  $I_c$  is collector current,  $V_{be}$  base-

\* Sometimes called bipolar junction transistors or b.j.t. because their operation involves both polarities of charge carrier. The usual type of f.e.t. is also a junction device, but it is unipolar because only one polarity of charge carrier is involved.

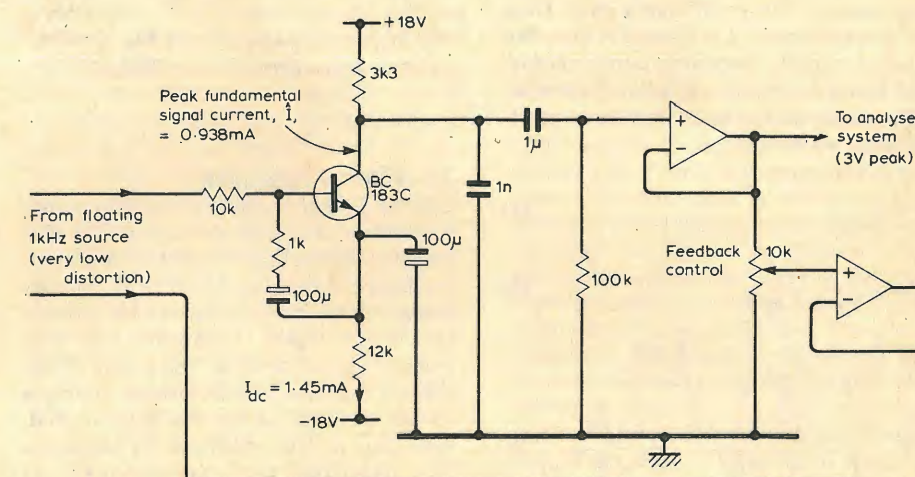


Fig. 1. Circuit used for distortion measurements.

to-emitter voltage, and the other symbols are constants. (The tendency always to regard junction transistors as current-operated devices, and the current gain as the basic parameter for design purposes, should be most strongly discouraged, in my opinion.)

A practical junction transistor would be expected to follow the above law much more closely than an f.e.t. would be expected to follow a parabolic law, so that there seemed good reason for thinking that the curious wiggles in the Fig. 2 curves might be theoretically explicable on the basis of equation 1.

### Determining transfer characteristic

For analysis the circuit may be simplified to that shown in Fig. 3, in which the transistor is assumed to follow equation 1. It may be shown that the incremental signal input and output voltages of the circuit are related by

$$v_{out} = -R_L I_{dc} \left[ \exp \frac{qV_{in}}{kT} \times \exp \frac{qV_{out}}{kT} - 1 \right] \quad (2)$$

where  $q$  is the electronic charge ( $1.60 \times 10^{-19}$  coulomb)  $k$  Boltzmann's constant ( $1.38 \times 10^{-23}$  joules/deg C) and  $T$  absolute temperature. To be able to calculate the harmonics in  $v_{out}$  when  $v_{in}$  in equation 2 is put equal to  $\hat{V}_{in} \sin \omega t$ , the relation must be expressed in the form of a power series:

$$v_{out} = a_1 v_{in} + a_2 v_{in}^2 + a_3 v_{in}^3 + a_4 v_{in}^4 + \dots \quad (3)$$

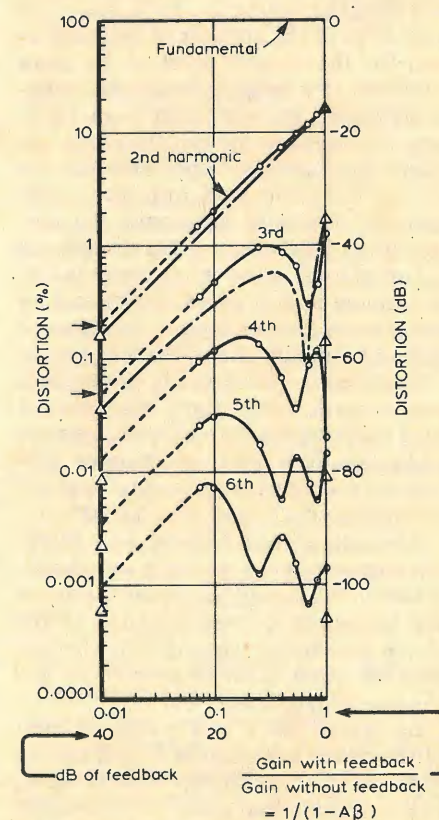


Fig. 2. Measured and calculated results for the Fig. 1 circuit.



The values of the coefficients  $a_1, a_2, a_3$  etc may be found by using Maclaurin's theorem, which says

$$a_1 = \left[ \frac{dv_{out}}{dv_{in}} \right]_{v_{in}=0}$$

$$a_2 = \frac{1}{2!} \left[ \frac{d^2 v_{out}}{dv_{in}^2} \right]_{v_{in}=0}$$

$$a_3 = \frac{1}{3!} \left[ \frac{d^3 v_{out}}{dv_{in}^3} \right]_{v_{in}=0}$$

By successively differentiating equation 2 and putting  $v_{in}=0$  in the resultant expressions, the coefficients may thus be determined. Unfortunately the algebra rapidly becomes cumbersome, and being no mathematician, I gave up after determining the first three coefficients, which are

$$a_1 = \frac{A}{1-A\beta} \quad (4)$$

$$a_2 = \frac{1}{2!} \frac{q}{kT} \frac{A}{(1-A\beta)^3} \quad (5)$$

$$a_3 = \frac{1}{3!} \left( \frac{q}{kT} \right)^2 A \left[ \frac{1}{1-A\beta} - \frac{3|A\beta|}{(1-A\beta)^2} + \frac{3|A\beta|^2}{(1-A\beta)^3} - \frac{|A\beta|^3 + 3|A\beta|}{(1-A\beta)^4} + \frac{3|A\beta|^2}{(1-A\beta)^5} \right] \quad (6)$$

In these equations  $\beta$  is positive and  $A = -g_m R_L$ , where  $g_m$  is the transistor mutual conductance when  $v_{in} = v_{out} = 0$  and the collector current is  $I_{dc}$ .

### Determining the harmonics

Knowing the value of  $v_{in} (= \hat{V}_{in} \sin \omega t)$ , as a function of the amount of feedback in use, for the output level of 3V peak adopted, the output harmonic magnitudes may be calculated from equation 3 on the assumption that only the square-law term is responsible for the second harmonic and only the cubic term for the third harmonic. Because the output level is large, this simplifying assumption leads to appreciable, though not unduly gross, errors, and for better accuracy the production of some second harmonic due to the presence of a fourth-power term needs to be taken into account, etc. A fairly high output level was adopted in the experiments to make the high-order harmonics sufficiently large for straightforward measurement, i.e. well over 0.0001%.)

The calculated second and third-harmonic curves are shown chain-dotted in Fig. 2, and lie somewhat below the measured curves because of the above simplifying assumption. The reasons for other detailed differences will become apparent later on.

In view of the + and - signs in front of the terms in equation 6, and on the supposition that the expressions for  $a_4, a_5$  etc. will contain even more terms of both signs, one can at least say that it is hardly surprising that the measured curves for the higher-order harmonics in Fig. 2 are of a more complex type.

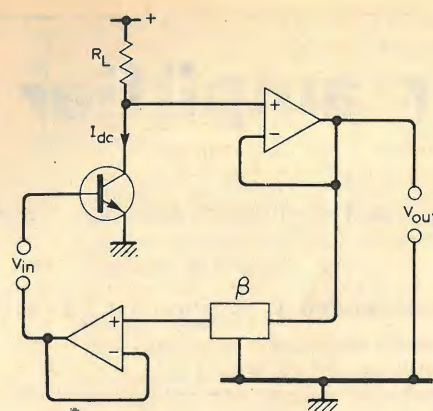


Fig. 3. Simplified version of Fig. 1 with d.c. bias arrangements omitted.

### Alternative approach

The method of analysis presented above basically involves determining the transfer characteristic for the complete feedback amplifier, and then calculating the harmonic magnitudes when a sine-wave input is handled via this transfer characteristic. The shape of the overall transfer characteristic changes as the amount of feedback is altered, resulting in the observed variation in the magnitudes of the various harmonics. It should be emphasized that no intermodulation concept is involved in this approach when the input to the complete circuit is a single sine-wave signal.

An alternative approach, which is very helpful in providing further insight, involves thinking simply in terms of the invariant transfer characteristic of the forward amplifying device. Intermodulation effects then do have to be taken into account, for the forward amplifying device receives inputs from both the sine-wave input signal and also via the  $\beta$ -network from the amplifier output, the last-mentioned contribution containing harmonics which intermodulate with the fundamental and with each other.

In particular, the second harmonic and fundamental intermodulate to produce a component at third-harmonic frequency, and careful consideration of the waveform polarities involved shows that this third-harmonic component is in antiphase with that produced by straightforward third-harmonic distortion of the fundamental. The null in the curve for total third-harmonic distortion thus occurs when the amount of feedback is such as to make these oppositely-phased third-harmonic components of equal magnitude. The fact that in the measured third-harmonic curve, a minimum rather than a perfect null is observed, is believed to be because slight phase errors in the experimental circuit prevented the two third-harmonic components from being exactly in antiphase.

A further intermodulation effect is that the fundamental and third harmonic intermodulate to produce a

second-harmonic component which, though of considerably smaller magnitude than that produced by straightforward second-harmonic distortion of the fundamental, nevertheless slightly modifies the shape of the second-harmonic curve.

The percentage of second harmonic generated within the transistor, at constant output, is proportional to the third-harmonic voltage fed back into the base circuit, but the percentage output distortion is reduced  $(1-A\beta)$  times relative to this by negative feedback. For working conditions to the left of the null in the third-harmonic curve, this intermodulation-generated second harmonic is in the same phase as that produced by straightforward second-harmonic distortion. Its magnitude at the amplifier output, with enough feedback to bring the working point onto the approximately constant-slope part of the third-harmonic curve, is such as to lift the position of the second-harmonic curve by a constant distance, and the calculated spacing is of the order shown.

It is instructive to compare the Fig. 2 curves with those of Fig. 7 in Part 5. There are two basic differences (a) the f.e.t. curves show no nulls or minima; and (b) the measured f.e.t. second-harmonic curve does not exhibit the departure from linearity evident in the Fig. 2 curve. The reason for (a) above is believed to be that, for the f.e.t. specimens used, the harmonic-distortion-generated third-harmonic component was in phase, rather than in antiphase, with the component generated by intermodulation. The high-order terms in the transfer characteristic for an f.e.t., unlike those for a junction transistor, seem to vary from one specimen to another - the one used for the Fig. 7 (Part 5) results had been selected for low third harmonic. It may well be that some other specimens would give curves with nulls, but this has not been investigated.

The reason for difference (b) above is simply that the signal level was too low to make the effect noticeable. Though the f.e.t. and the junction transistor were both worked at the same ratio of peak signal to direct working current, the f.e.t., because of its different type of transfer characteristic, gave less second-harmonic distortion in the absence of feedback - see equations 16 and 17 in Part 5. On turning up the signal level in the f.e.t. circuit for 4V rather than 3V peak output, an appreciable departure of the second-harmonic curve from linearity was observed.

### High-feedback theory

It is a characteristic of the Fig. 2 curves that all their complex features disappear when enough feedback is applied, and this fact suggests that maybe the high-feedback parts of the curves at the left could be calculated in a manner

devoid of the above complications. This indeed turns out to be the case, and it is thought that appreciation of this fact is of considerable engineering value, for in the majority of practical applications one is really only interested in the performance with plenty of feedback applied.

Any amplifier, without feedback, can in principle be made to give a perfectly sinusoidal output voltage, at a specified level, by feeding an appropriately distorted waveform to its input. With negative feedback applied, this same totally undistorted output voltage can be maintained if  $V_{in}$  (Fig. 4) is arranged to contain the necessary distorted error voltage, as above, plus some extra fundamental to cancel the fundamental being injected negatively into the input circuit via the  $\beta$ -network. (With undistorted output, the feedback voltage is, of course, also perfectly sinusoidal.) Thus, as  $\beta$  is increased,  $V_{in}$  has to supply a constant-amplitude harmonic spectrum plus an increased amount of fundamental. The magnitude of the required fundamental input, for the specified constant output voltage  $V_{out}$ , is given by the usual feedback formula.

$$\frac{V_{out}}{V_{in}} = \frac{A}{1-A\beta}$$

which for the present purpose is more conveniently arranged as

$$V_{in} = V_{out} \frac{1-A\beta}{A}$$

Since the harmonic part of the input is quite constant, the percentage input distortion is inversely proportional to the amount of fundamental input voltage, i.e. it is proportional at  $1/(1-A\beta)$ , and this applies at every harmonic frequency. It also applies whether the amount of feedback is large or small.

It is thus seen that the distortion situation for a feedback amplifier is really very much simpler when viewed on this basis of percentage input distortion for a pure output, than when considered on the more usual basis of the output distortion for a pure sinusoidal input. At this point the reader may well object that, while it may indeed be easier to consider the feedback mechanism on this basis, the concept is artificial and not related to the way amplifiers are used in practice. The utility of the approach, however, lies in the fact that, provided there is plenty of feedback, the distortions become practically identical whether expressed on a distorted-input/pure-output basis, or on the usual distorted-output/pure-input basis. Thus if the percentage distortion with no feedback is calculated on a pure-output/distorted input basis - which turns out to be relatively easy - then the distortion with plenty of feedback applied, expressed in the customary manner, is equal to the just-mentioned no-feedback percentage divided by  $(1-A\beta)$ , the output level

being kept constant. This applies both to total harmonic distortion and also to all individual harmonics of practical significance, provided only that the amount of feedback is sufficiently large. For the working conditions relevant to Fig. 2, or Fig. 7 of Part 5, it is evident that 20 to 26dB of feedback would be "sufficiently large."

It is now necessary to justify the statement that the distortion with plenty of feedback is practically the same whether expressed on a distorted-input/pure-output basis, or on a distorted-output/pure-input basis. With reference to Fig. 4, consider the state of affairs when  $V_{in}$  is of pure sine waveform, suitably adjusted in magnitude to maintain a constant output voltage no matter how much feedback there is. With no feedback,  $V'$  will be equal to  $V_{in}$  and will be sinusoidal,  $V_{out}$  being highly distorted. As the amount of feedback is increased,  $V_{out}$  becomes more and more nearly sinusoidal, which requires that the  $V'$  waveform must approximate more and more closely to that specific highly-distorted waveform, characteristic of the particular forward amplifier, which will make it deliver a perfectly sinusoidal output. The whole of the distortion in  $V'$  - call it  $V_{dist}$  - is supplied from the  $\beta$ -network, since  $V_{in}$  is pure. When the amount of feedback is large, the fundamental output from the  $\beta$ -network, injected into the input circuit, is very nearly equal in magnitude to  $V_{in}$ . Hence the percentage distortion in the output from the  $\beta$ -network, and therefore also in the amplifier output voltage, which feeds the  $\beta$ -network, is very nearly  $(V_{dist}/V_{in}) \times 100\%$ . If now, with this large amount of feedback applied, a slight harmonic content is introduced into the  $V_{in}$  waveform so as to make the output perfectly sinusoidal, neither the magnitude of  $V_{in}$ , nor the harmonic content of the  $V'$  waveform, will change by more than a tiny amount, so that the distortion will still be given quite closely by  $(V_{dist}/V_{in}) \times 100\%$ . Thus the larger the amount of feedback, the more nearly does the percentage output distortion for pure input become equal to the percentage input distortion for pure output.

Another argument to support the statement that the percentage distortions, with a large amount of feedback applied, are virtually the same when expressed on either basis mentioned above, is as follows. Referring to Fig. 4 again, suppose  $V_{in}$  contains the necessary harmonics to make  $V_{out}$  perfectly sinusoidal. Now, with these input harmonics still present, imagine that we add a further set of input harmonics, each of equal magnitude to, and in antiphase with, the corresponding harmonic already there. The result will be to cancel all the input harmonics, but introduce harmonics into  $V_{out}$ . If the harmonics thus introduced into  $V_{out}$  are simply the result of the faithful amplification of the additional set of

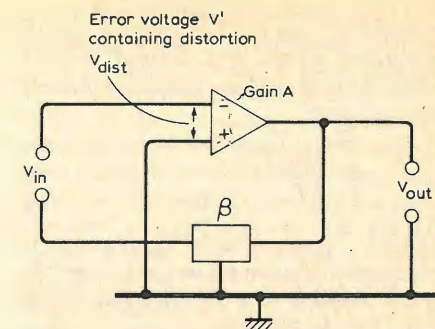


Fig. 4. Basic feedback amplifier configuration.

harmonics fed in, then it follows that the percentage distortions must be the same whether considered on a distortionless input or a distortionless output basis. Whether this is nearly enough the case for practical purposes depends on how low is the intermodulation distortion introduced by the complete feedback amplifier when fed with these small-amplitude additional input harmonics in the presence of a large fundamental input, and clearly the more feedback there is, the less significant will be any "false harmonics" introduced by intermodulation - intermodulation between the fundamental and the second harmonic might introduce some third harmonic, for example.

Thus, once again, the conclusion is reached that, provided there is enough feedback, harmonic-distortion percentages will be very nearly the same whether expressed on the normal pure-input basis, or on the inverse basis of input distortion for pure output.

### A distortion theorem

The ideas discussed above may be formulated as a distortion theorem, applicable to total and to individual harmonic distortion:

"The percentage harmonic distortion in the output of an amplifier having a large amount of feedback and a sine-wave input, is very nearly equal to the percentage input distortion for distortionless output without feedback, divided by  $(1-A\beta)$ ."

The usefulness of this theorem is dependent on knowing the input distortion required to give distortionless output without feedback, for common amplifying devices and circuits, but fortunately the theory required is relatively simple. Such distortion can be termed "inverse distortion."

### Junction transistor inverse-distortion theory

A simple single-ended junction-transistor stage will be considered first, the transistor being assumed to follow equation 1. When  $V_{be}$  is such as to cause  $I_c$  to vary sinusoidally,

$$I_{dc} + \hat{I}_c \sin \omega t = I_c \exp \frac{qV_{be}}{kT} \quad (7)$$



in which  $V_{be}$  has the appropriate special waveform which it is desired to find. When  $\hat{I}_{dc}\sin\omega t$  passes instantaneously through zero

$$I_{dc} = I_0 \exp \frac{qV_{dc}}{kT} \quad (8)$$

where  $V_{dc}$  is the value of  $V_{be}$  required to establish the mean collector current  $I_{dc}$  in the absence of a signal input. Equation 7 may be written

$$\log_e \left[ \frac{I_{dc} + \hat{I}_{dc}\sin\omega t}{I_0} \right] = \frac{qV_{be}}{kT}$$

from which may be derived

$$V_{be} = \frac{kT}{q} \left[ \log_e \frac{I_{dc}}{I_0} + \log_e \left( 1 + \frac{\hat{I}_{dc}\sin\omega t}{I_{dc}} \right) \right] \quad (9)$$

But from equation 8,

$$\log_e \frac{I_{dc}}{I_0} = \frac{qV_{dc}}{kT},$$

so that equation 9 becomes

$$V_{be} = V_{dc} + \frac{kT}{q} \log_e \left( 1 + \frac{\hat{I}_{dc}\sin\omega t}{I_{dc}} \right)$$

We now use the fact that

$$\log_e(1+x) = x - x^2/2 + x^3/3 - x^4/4 + \dots$$

which leads to

$$V_{be} = V_{dc} + \frac{kT}{q} \left[ \frac{\hat{I}_{dc}}{I_{dc}} \sin\omega t - \frac{1}{2} \left( \frac{\hat{I}_{dc}}{I_{dc}} \right)^2 \sin^2\omega t + \frac{1}{6} \left( \frac{\hat{I}_{dc}}{I_{dc}} \right)^3 \sin^3\omega t - \frac{1}{24} \left( \frac{\hat{I}_{dc}}{I_{dc}} \right)^4 \sin^4\omega t + \dots \right] \quad (10)$$

On the assumption that  $\hat{I}_{dc}$  is not so large that, for example, the second harmonic generated by the  $\sin^4\omega t$  term is large enough to cause serious error, equation 10 yields harmonic percentages as given in the middle column of Table 1. Since  $g_m = qI_{dc}/kT$  and  $\hat{I} = g_m \hat{V}_{in}$ , we may replace  $\hat{I}_{dc}$  by  $q\hat{V}_{in}/kT$ . At 290 K, which is approximately relevant to low-level stages at least,  $kT/q$  is 25mV. These facts enable the results in the right-hand column of Table 1 to be calculated.

**Table 1. Theoretical input distortion percentages for pure sinusoidal output from ideal junction transistor without feedback.**

Harmonic number	Distortion %	Distortion (%), alternative formulae for 290K ( $V_{in}$ in mV)
2	$25(\hat{I}_{dc}/I_{dc})^2$	$\hat{V}_{in}^2$
3	$8.33(\hat{I}_{dc}/I_{dc})^3$	$1.33 \times 10^{-2} \hat{V}_{in}^3$
4	$3.13(\hat{I}_{dc}/I_{dc})^4$	$2.00 \times 10^{-4} \hat{V}_{in}^4$
5	$1.25(\hat{I}_{dc}/I_{dc})^5$	$3.20 \times 10^{-6} \hat{V}_{in}^5$
6	$0.521(\hat{I}_{dc}/I_{dc})^6$	$5.33 \times 10^{-8} \hat{V}_{in}^6$

### Comparison with "normal" distortion

It is instructive to compare the Table 1 results with those giving the output distortion for an ideal sine-wave-driven junction transistor without feedback. Referring to equation 1, put  $V_{be} = V_{dc} + \hat{V}_{in}\sin\omega t$ , where  $V_{dc}$  is a direct bias voltage. This leads to

$$\frac{i_c}{I_{dc}} = \exp \frac{q\hat{V}_{in}\sin\omega t}{kT} - 1 \quad (11)$$

where  $i_c$  is the instantaneous signal component of collector current and  $I_{dc}$  the collector current when  $\hat{V}_{in}\sin\omega t = 0$ . This time the required mathematical fact is that

$$\exp x = 1 + x + \frac{x^2}{2!} + \frac{x^3}{3!} + \frac{x^4}{4!} + \dots$$

which gives

$$\frac{i_c}{I_{dc}} = \frac{q\hat{V}_{in}\sin\omega t}{kT} + \frac{1}{2} \left( \frac{q\hat{V}_{in}}{kT} \right)^2 \sin^2\omega t + \frac{1}{6} \left( \frac{q\hat{V}_{in}}{kT} \right)^3 \sin^3\omega t + \frac{1}{24} \left( \frac{q\hat{V}_{in}}{kT} \right)^4 \sin^4\omega t + \dots \quad (12)$$

The harmonic percentages may then be evaluated on the same basis as for Table 1, as functions of both  $\hat{V}_{in}$  and  $\hat{I}_{dc}$ , since

$$\hat{V}_{in} = \frac{\hat{I}_{dc}}{g_m} \times \frac{kT}{q}$$

Substituting this in equation 12 gives

$$\frac{i_c}{I_{dc}} = \frac{\hat{I}_{dc}}{I_{dc}} \sin\omega t + \frac{1}{2} \left( \frac{\hat{I}_{dc}}{I_{dc}} \right)^2 \sin^2\omega t + \frac{1}{6} \left( \frac{\hat{I}_{dc}}{I_{dc}} \right)^3 \sin^3\omega t + \frac{1}{24} \left( \frac{\hat{I}_{dc}}{I_{dc}} \right)^4 \sin^4\omega t + \dots \quad (13)$$

From equations 12 and 13 have been calculated the results given in Table 2. As before it is assumed that  $\hat{I}_{dc}$  is small enough to ensure that a negligible portion of the total second-harmonic generated arises from the presence of the  $\sin^4\omega t$  term, etc. However, since the terms in equation 10 fall off in magnitude with increasing order less rapidly than in equation 13, a given high value of  $\hat{I}_{dc}$  causes larger errors in the inverse-distortion figures of Table 1 than it does under the conditions of Table 2.

**Table 2. Theoretical output distortion percentages for pure sinusoidal input voltage to ideal junction transistor without feedback.**

Harmonic number	Distortion (%)	Distortion (%), alternative formulae for 290 K ( $\hat{V}_{in}$ in mV)
2	$25(\hat{I}_{dc}/I_{dc})^2$	$\hat{V}_{in}^2$
3	$4.17(\hat{I}_{dc}/I_{dc})^3$	$6.67 \times 10^{-3} \hat{V}_{in}^3$
4	$0.521(\hat{I}_{dc}/I_{dc})^4$	$3.33 \times 10^{-4} \hat{V}_{in}^4$
5	$0.0521(\hat{I}_{dc}/I_{dc})^5$	$1.33 \times 10^{-5} \hat{V}_{in}^5$
6	$0.00434(\hat{I}_{dc}/I_{dc})^6$	$4.44 \times 10^{-7} \hat{V}_{in}^6$

### Inverse distortion for parabolic device

For an amplifier having the general parabolic transfer characteristic given by equation 1 of Part 5, repeated here as equation 14, the formulae for input distortion for pure

$$V_{out} = Av' + \alpha(Av')^2 \quad (14)$$

sinusoidal output without feedback are

given in Table 3, middle column. For an ideal f.e.t. there is the restriction that the bottom of the parabola must lie on the zero-drain-current axis, as shown in Fig. 4 of Part 5, and it then follows that  $\alpha\hat{V}_{out}$  may be replaced by  $\frac{1}{4}(\hat{I}_{dc})^2$ , giving the formulae in the right-hand column of Table 3. (This substitution may also be made for  $\alpha V_{out}$  in Table 1 of Part 5 when applied to an ideal f.e.t.)

**Table 3. Theoretical input distortion for pure output for general parabolic device, and f.e.t. without feedback.**

Harmonic number	Distortion (percentage)	
	General parabolic device	Ideal f.e.t.
2	$50\alpha\hat{V}_{out}^2$	$12.5\hat{I}_{dc}/I_{dc}^2$
3	$50\alpha^2\hat{V}_{out}^3$	$3.12(\hat{I}_{dc}/I_{dc})^3$
4	$62.5\alpha^3\hat{V}_{out}^4$	$0.977(\hat{I}_{dc}/I_{dc})^4$
5	$87.5\alpha^4\hat{V}_{out}^5$	$0.342(\hat{I}_{dc}/I_{dc})^5$
6	$131\alpha^5\hat{V}_{out}^6$	$0.128(\hat{I}_{dc}/I_{dc})^6$

Comparing the right-hand column of Table 3 with the middle column of Table 1, the input harmonics for the f.e.t., at a given  $\hat{I}_{dc}$  are smaller and decay more rapidly with increasing order than for a voltage-driven junction transistor. However, in many practical feedback circuits, this apparent disadvantage of the junction transistor will be more than compensated by the fact that it has a much higher mutual conductance, giving a higher feedback loop gain and thus reducing all significant harmonics to a lower level than for the f.e.t.

With regard to the f.e.t. investigation of Part 5, dividing the figures determined from the right-hand column of Table 3 by 100 gives points on the left-hand vertical axis of Fig. 7 in Part 5 which coincide with the intercepts of the chain-dotted curves.

### Relationship to experimental results

The distortion theorem formulated above may be used to calculate quickly and easily, the approximate output distortion for a single junction transistor stage having, say, 40dB of feedback, for  $\hat{I}_{dc} = 0.647$  as used in the experiments with the Fig. 1 circuit. The no-feedback inverse-distortion figures are determined from the middle column of Table 1, and are divided by 100 to give the predicted distortion with feedback. The values obtained are indicated by triangles on the left-hand vertical axis of Fig. 2.

As already explained, the Table 1 formulae assume  $\hat{I}_{dc}$  is small enough for the amount of second harmonic produced by the 4th and 6th power terms in the power series to be ignored, etc. With  $\hat{I}_{dc}$  as high as 0.647, there is, however, an appreciable error due to this cause. Calculation shows that the amounts of inverse second harmonic arising from the  $\sin^4\omega t$  and  $\sin^6\omega t$  terms in equation 10 are approximately 21% and 4% of the amount due to the  $\sin^2\omega t$  term, the amounts produced by even higher-order terms being relatively negligible. Thus the true second-harmonic figure would be expected to

be about 26% higher than that calculated from Table 1, the error becoming rapidly smaller with reduction in signal level. This more accurate theoretical prediction is indicated by the upper arrow at the left of Fig. 2, and ties up well with the broken-line extrapolation of the measured curve.

At the zero-feedback end of the Fig. 2 curves the simple theoretical distortion values are given by the middle column of Table 2, for  $\hat{I}_{dc} = 0.647$ , and the values obtained are indicated by the triangles on the right-hand vertical axis of Fig. 2. As already stated, the errors under the Table 2 conditions, caused by working at a rather high signal level, are much less than for Table 1, but there are other causes of errors to be considered. Nevertheless, the calculated second-harmonic percentage agrees quite well with the experimental value. The shape of the second-harmonic curve can thus be explained in terms of the increasing effect of the high-order terms in the power series as the amount of feedback is increased — an alternative but equally sound explanation to that previously given involving intermodulation within the forward amplifier.

The theoretical zero-feedback points, marked by triangles, for harmonics higher than the second do not agree well with the measured values. The reason for this is believed to be that when the Fig 1 circuit is set for nominally zero feedback, a small but finite amount of negative feedback is effectively still in operation, mainly because of the presence of finite resist-

ance in the base circuit. If this resistance, including  $r_{bb'}$ , totals 1.2k $\Omega$ , and assuming  $\beta_{ac}$  or  $h_{fe}$  of 500, it is equivalent to a resistance of 2.4 $\Omega$  in the emitter lead, causing  $1/(1-\alpha\beta)$  to be effectively 0.88 when set for nominally 1.0. To allow for this, the extreme right-hand plotted points on all the experimental curves should be moved to the left to  $1/(1-\alpha\beta) = 0.88$ . The effect of the 2.4 $\Omega$  is negligible because of the logarithmic scale used in Fig. 2, except toward the right-hand side of the curves. With the curves thus shifted to the left, it seems reasonable to suppose that continuing the patterns of undulations already established, towards the  $1/(1-\alpha\beta) = 1.0$  axis, would bring the curves approximately to the theoretical values marked by triangles.

When allowance is made for the production of third harmonic by the  $\sin^5\omega t$ ,  $\sin^7\omega t$  and  $\sin^9\omega t$  terms in equation 10, the magnitude of the third-harmonic distortion voltage is increased by approximately 32%, raising the calculated value to that indicated by the lower arrow in Fig. 2, which again then ties up well with the broken-line extrapolation of the measured curve. The corresponding tedious calculations have not been done for the 4th, 5th and 6th harmonics, but it seems probable that they, too, would raise the levels of the points marked by triangles to give reasonable agreement with the broken-line, 45°, extrapolations of the measured curves.

There is a further small point which must now be mentioned. In Table 2, for

a junction transistor without feedback driven by a sine-wave voltage, the factor  $\hat{I}/I_{dc}$  appears.  $\hat{I}$  is the peak value of the fundamental current and  $I_{dc}$  is the value of the collector current at the moment when the input signal voltage goes through zero. It would also be the quiescent current, if the transistor were operated at fixed bias voltage, and the mean current with the signal applied would then be larger than  $I_{dc}$  because of the rectifying action. However, the mean current is prevented from rising significantly when the signal is present in the Fig. 1 circuit, owing to the virtually-constant current in the 12k $\Omega$  emitter resistor. This results in the distortion being higher than the simple theory predicts. The fact that the measured second-harmonic curve goes through the 16% point predicted by Table 2 at its top end is thus fortuitous. The effect just mentioned tends to raise the level of the point, whereas the fact that there is a little feedback in action, even when the control is set for nominally zero feedback, tends to lower it. Once there is plenty of feedback in action both these effects become negligible.

It can thus be concluded generally that provided plenty of feedback is assumed right at the beginning, the more awkward parts of the theory outlined in this article, though academically interesting, do not need to be taken into account for design purposes.  $\square$

(To be continued)

## Investigating black holes

The part played by radio and electronics in studying black holes — those strange phenomena of the universe in which all resistance to gravity seems to have been overcome — was brought out recently by Professor R. L. F. Boyd, of University College London, in a contribution to the report of the Science Research Council for the year 1977-78. He revealed that scientists at his own university had been using a unique photon counting system to study the velocity dispersion of stars in the giant elliptical active galaxy in Virgo, and that they had obtained results pointing to the existence at its centre of a black hole of more than  $10^9$  times the mass of the sun. There was here a strong hint that the mysterious energy source at the heart of quasars or in the highly agitated cores of Seyfert galaxies could be of the same nature. Using data from the Ariel V satellite, Leicester University, in collaboration with Sussex University and employing the Anglo-Australian Telescope, had identified 15 strong extragalactic x-ray sources as Seyferts. The x-ray power output of such sources exceeded that of a billion suns. Its correlation with infra-red and optical data showed that it originated close to the nuclei and was probably due to the collision of electrons of relativistic energy with low energy photons (the inverse Compton effect). It also seemed likely from the form of the spectral energy

distribution that the so far unresolved cosmic x-ray background was, at least in part, due to these very active galaxies.

Professor Boyd went on to say that quasars "whose unimaginably concentrated energy sources may also be black holes, are important radio objects." The Nuffield Radio Astronomy Laboratory at Manchester University had studied angular structures of radio sources and, in a collaboration with the USA using the very long base-line technique, had found a jet-like feature some 5,000 light years long extending from the very compact core of quasar 3C147. Strong beaming of energy from the core to the distant parts of certain radio galaxies was also suggested by studies at the Mullard Radio Astronomy Observatory at Cambridge University of the radio galaxy NGC 6251 which was found to have a very narrow jet 600,000 light years in length. In our own galaxy compact objects were also important, being formed when stars ran out of nuclear fuel and died — either slowly to become white dwarfs or violently in supernova explosions which led to neutron stars (of mass comparable to the sun but of diameter only 10-20 kilometres) or perhaps sometimes to black holes. The Mullard observatory, continued Professor Boyd, had found such an object at the centre of a vast interstellar shock wave which was a likely candidate for the core of the star which

exploded there. Manchester University, using several unique instruments on optical telescopes, notably the Anglo-Australian Telescope, had discovered two new supernova remnants and had obtained data of unequalled quality on structure and motion in the Monoceros remnant.  $\square$

### Technician on PO board

As part of a two-year experiment in industrial democracy, a factory technician, Mr Leonard Willett, has been appointed as a part-time member of the Post Office Board. He was nominated by the Post Office Engineering Union. Mr Willett joined the Post Office in 1935 and the Post Office Engineering Union in 1937 and has been a member of the union's National Executive Council since 1967. He became secretary of the Post Office London Factories Branch of the union in 1951, secretary of the Factories Unions Regional Council in 1953 and secretary of the Factories Council of Post Office Unions in 1972. Other part-time members of the PO Board serving till the end of the year are: Mr Derek Gladwin, Professor Michael Posner, Mrs J. E. Walsh, Mr Peter Walters and Lord Winstanley.  $\square$



continued from page 39

## Meteorological satellite one year old

The European Space Agency's meteorological satellite, Meteosat-1 which was launched on 23rd November, 1977, has completed its first year in orbit. After its launch the satellite reached its geostationary position at 0° longitude over the Gulf of Guinea on 7th December, 1977, and transmitted its first visible and infrared images of the Earth on the 9th and 11th of December.

This first year in orbit has shown how extremely well the onboard equipment is working and has satisfied ESA that it is carrying out "to perfection" its image-taking, dissemination and data-collection missions. During this time ground facilities have been set up at the European Space Operations Centre (ESOC) in Darmstadt, Germany, to handle the reception, processing, archiving and dissemination of the meteorological data from the satellite. These facilities gradually became operational and were ready to fulfil all the requirements for the first Global Atmospheric Research Programme (GARP) experiment which began on 1st December, 1978.

Meteosat is now transmitting high quality

visible and infrared images of the Earth every half hour and retransmitting about 300 formats, in various digital and analogue forms, to the primary and secondary data user stations (PDUSs and SDUSs). It also acts as a relay for the distribution of image data from the US satellite, GOES-1, which is situated over the Atlantic. These data are transmitted by the Centre de Météorologie Spatiale (CMS) at Lannion at the rate of about 20 formats per day. In addition, November '78 saw the start of the first transmission tests via Meteosat of image data from the US GOES-3 satellite, located over the Indian Ocean. Users of the four PDUSs and the forty or so SDUSs currently in service in the Meteosat coverage area can therefore receive data from three geostationary satellites covering approximately two-thirds of the globe.

Successful experiments have also been carried out between Meteosat and mobile data-collection platforms (DCPs). There are now about 15 operational DCPs, and another ten are expected to be in service early this year.

constructed in space from basic units and materials transported from Earth by launch vehicles such as the Space Shuttle.

The conference was organised by the Royal Aeronautical Society and the American Institute of Aeronautics and Astronautics provided a forum so that delegates from many different engineering and scientific disciplines could examine the interactions of aerospace operations and technology on the provision of energy for mankind. The idea for this meeting came from HRH Prince Charles, The Prince of Wales, who delivered the opening speech.

British Aerospace's Dynamics Group at Bristol, who are already committed to an involvement in solar arrays for spacecraft applications, recognise this activity as an expanding business. Under contract to the European Space Agency (ESA) they are designing a 33 square-metre, 4kW solar array to power NASA's space telescope. They are also developing a 6kW lightweight flexible, fold-out array for future communication satellites and proposing to augment the Space Shuttle power using solar power modules of up to 60kW. Arrays to provide up to 500kW have also been proposed as space power sources and these could form modules for 2MW space power station schemes.

## Soviet rocket designer dies

A November '78 issue of *Soviet News*, a journal published by the Press Department of the Soviet Embassy in London, reports that Professor Konstantin Bushuyev, who was the director of the international Soyuz-Apollo test project for the Soviet side, has died at the age of 64. An obituary, signed by Leonid Brezhnev and other Soviet leaders, notes the great contribution which he made to the exploration of outer space.

Professor Bushuyev worked on the development and building of piloted spaceships and automatic spacecraft for the exploration of near-Earth space, the Moon, Mars and Venus.

## Space solar power stations could provide energy for Earth

A British Aerospace engineer, in presenting a paper entitled "A review of some critical aspects of satellite power systems" at an energy and aerospace conference held in London during December, suggested the possibility of giant solar space power stations providing energy for domestic and industrial uses on Earth. The engineer, Ivor Franklin, project manager for Solar Power Systems, part of the British Aerospace Dynamics Group, said that if pilot studies to develop solar arrays into space power sources proved economically feasible, by the end of the century giant orbiting solar power stations could be in use. The stations would cover many square miles of space and would be

## Land mobile radio conference in '79

In 1975 the IERE held its first international conference on mobile communications. Since then, there have been so many developments in this field that the institution is now planning to hold another, on the same general theme, in September, 1979.

The aim of the conference will be to provide a forum for an in-depth discussion of the developing technologies and the ways in which they can best be used for land mobile radio.

The institution, which has been inviting offers of papers for the conference, informs us that while initially the call for abstracts of the papers was given an early deadline, in order to permit the most recent work to be incorporated, this was extended to late December. Papers in their final form will therefore now be required by 4th April, 1979, and not by February as was previously announced by the institution.

## Breadboard show, a success

Breadboard '78, "the first show of its kind for the electronics enthusiast" scored a major success in London, say the organisers, Trident Exhibitions Ltd. During its five-day run, over 10,000 enthusiasts and dealers visited the show and the exhibitors were delighted and "not a little overwhelmed" by the response.

Some of the exhibitors were able to sell hundreds of pounds worth of goods on their stands, while others, like Vero Electronics, who were not selling items off the stand, were able to make many new contacts amongst the dealers. In fact, making contact with the dealers seemed to be the main aim of many of the exhibitors at the show.

One exhibitor from Electrovalue was selling equipment on the stand and feeding orders into a computer terminal at the show, to be processed the same day at his company's Surrey headquarters. Another apparently satisfied exhibitor was the managing director of AP Products, he said that he had taken many orders and it was well worth the trip from Germany.

The show director, Angela Larcombe, of Trident, says that in view of the show's success, Breadboard '79 will have to be moved to a bigger venue and that dates and the new venue will be announced in the near future.

## Emergency network

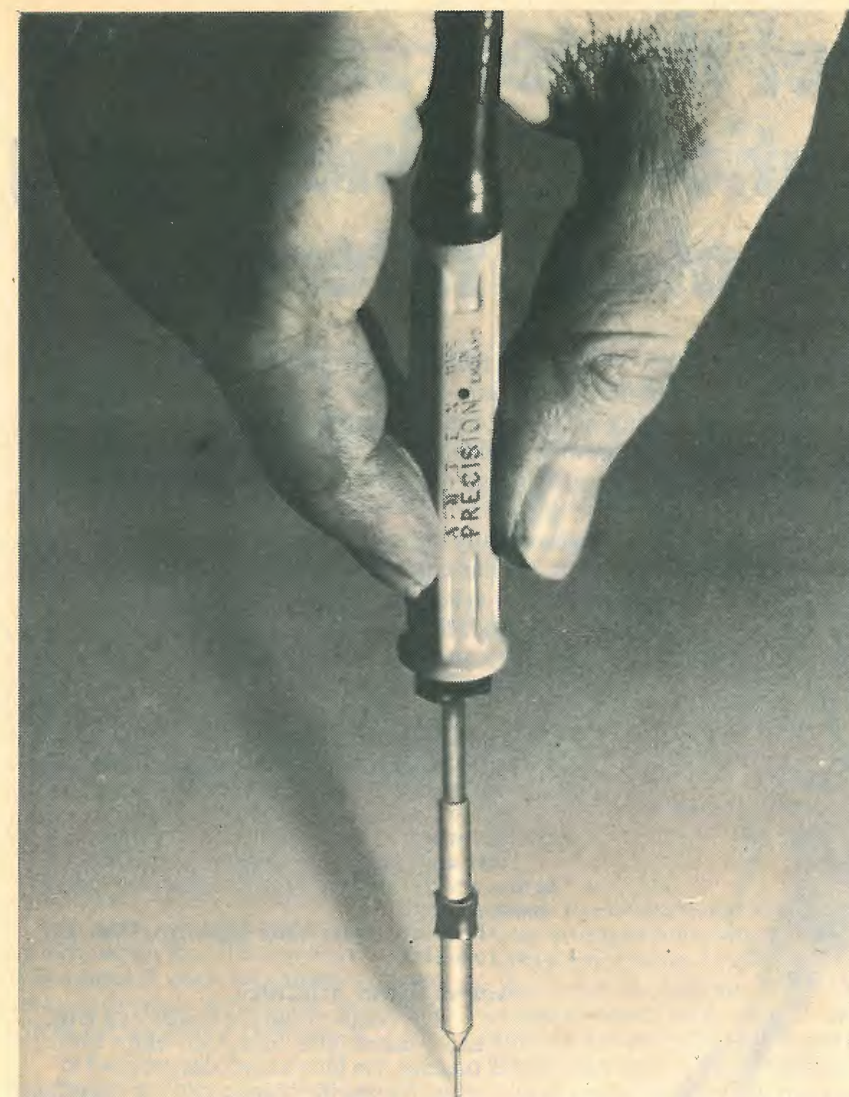
Some 75 members of the Radio Amateurs' Emergency Network (RAEN) provided round-the-clock communications assistance to the St John's Ambulance Service in Birmingham and the West Midlands, dealing with more than 500 emergency calls in the first two weeks. An industrial dispute had left ambulances without their usual radio communications and shifts of four RAEN operators were stationed at the Birmingham headquarters while eight two-man mobile teams accompanied ambulances.

## Long distance television

A new edition of Roger Bunney's long-distance television guide has now been published by Bernard Babini Ltd under the new title "long distance television reception (TV-DX) for the Enthusiast." The 120-page booklet (£1.45) is publication number BP52 and now includes a section on reception of transmission from satellites. Details are also given of a number of mast-head pre-amplifiers, receiver requirements for European standards as well as extensive information on receiving aerials and propagation modes for long-distance reception on the television bands.

## Register of consulting scientists

A 1978 edition (4th) of the Register of Consulting Scientists has recently been published. The publication, which was edited by E. A. G. Liddiard of the Fulmer Research Institute in Slough, Bucks., lists contract research organisations and other scientific and technical services.



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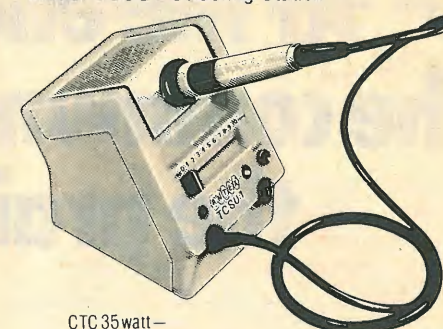
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# Electronic organ tone system — 4

Reproduction, coupling and tonal variations

by A. D. Ryder, M.A., Ph.D., F.I.E.E.

This article starts the description of optional additions. Although they are intended for the basic design, some of the circuits may be useful in other organ applications.

A VISIT to a reasonably modern church organ will help in deciding on the mechanical details. Whether an existing console is used or new parts are obtained, it is likely that key dimensions and spacing will be standard. The pedalboard should be correctly set laterally, with its middle D accurately in line with the middle D manual keys. Standard key-operating forces are 4 oz for the manuals and 4 lb for pedals because very light keyboards are hard to control. The most important playing requirement is that the contacts make at the same point for every key, and halfway down the stroke is about right. It is useful if each keyboard with its contacts is made detachable so that it can be set precisely on the bench.

## Sustain

To provide a slow decay of the tone after a key is released, the pull-down voltage in Fig. 14 may be raised towards ground. A wider range can be achieved by increasing the 100kΩ resistor and hence the maximum pull-down path of an n-p-n transistor with its base connected to ground. More elaborate shaping may of course be used, and the light loading presented by the gates simplifies the design. The muting time-constant in Fig. 34 must be adjusted to suit the envelope in use, and for low pull-down currents the 10kΩ base resistor of the second transistor in Fig. 32 will need to be increased to achieve correct KD operation.

## Alternative reference generator

Sets of approximations to the e.t. scale were published by R. Staplefeld in 1970. The GIM device uses a 9-bit set with a worst case error of about 0.1% or 1000 p.p.m., and the set includes two perfect fifths, which are a drawback for some of the options. The 4040 c.m.o.s. counter can be used to obtain the 12-bit set, which has a worst case error of 80 p.p.m., and no perfect fifths. In the absence of any frequency modulation (to be discussed later) this greater accuracy does appear to produce an improvement in sound quality. Table 10

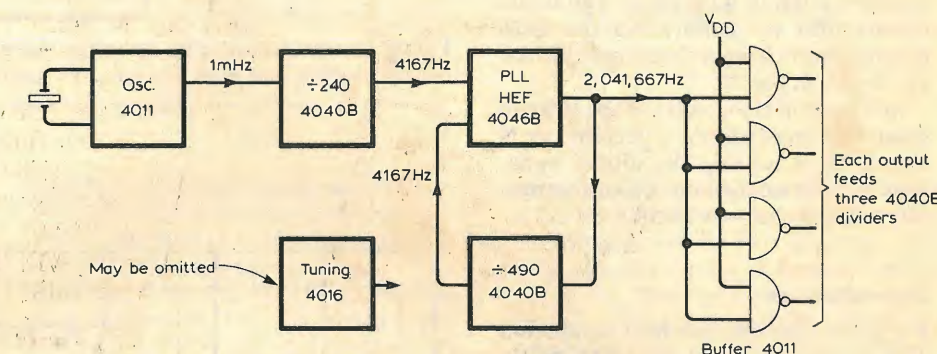


Fig. 37 Reference generator using 12-bit divisors.

shows the divisors and the departures from an exact scale with A 880Hz, for an input frequency of 2,041,667Hz.

In the block diagram of Fig. 37, the crystal oscillator is the same as in Fig. 4. The HEF4046B p.l.l. will operate at 2MHz with the same or similar component values as used for the standard 4046. Alternatively, a device such as the 561 may be used with appropriate circuit modification. In Fig. 4 the low-pass filter capacity should be increased to 22nF. For fixed tuning, the remaining i.c.s comprise a second 4011 used as a buffer, and 14 4040 counters. The generated reference frequencies are two octaves below those of table 2. Each gate card input capacitor is increased

from 1nF to 10nF and its p.l.l. comparator input is taken from pin 6 rather than pin 4 of the upper 4520 in Fig. 12. This gives an overall division of 240 instead of 60. In Fig. 10, the three resistors associated with the vibrato input are doubled, and the 100pF capacitor is increased to 4.7nF.

To achieve large divisions, a modified premature reset circuit is used, as shown in Fig. 38. The counter increments on negative-going clock edges and the count or clock low on which the AND node N goes high to charge the 10pF capacitor, is determined by the outputs connected via diodes to N. On the succeeding clock high the capacitor charge is shared with the R input stray capacitance and produces a reset pulse, during which the high at N is lost. The next clock low terminates the reset pulse and discharges the capacitor. The

Table 10 Divisors for alternative reference generator. The binary numbers are one less than the decimal because the reset pulse suppresses one count. The last two columns relate to an input frequency of 2,041,667 Hz (1MHz x 49/24). An asterisk shows the pin used for the output. Reference outputs have 22kΩ buffer resistors as in Fig 4.

Note	Decimal divisor	2048 (pin 1)	1024 (pin 15)	512 (pin 14)	256 (pin 12)	128 (pin 13)	64 (pin 4)	32 (pin 2)	16 (pin 3)	8 (pin 5)	4 (pin 6)	2 (pin 7)	1 (pin 9)	Output frequency	Error p.p.m.
C	3902	1	1	1	0	0	1	1	1	1	0	1		523.24	-29
C'	3683	1	1	1	0	0	1	1	0	0	1	1		554.35	-30
D	3476	1	1	0	1	1	0	0	1	0	0	1	1	587.36	+54
D'	3281	1	1	0	0	1	1	0	1	0	0	0	0	622.27	+25
E	3097	1	1	0	0	0	0	0	1	1	0	0	0	659.24	-23
F	2923	1	0	1	1	0	1	1	0	1	0	1	0	698.48	+39
F'	2759	1	0	1	0	1	1	0	0	0	1	1	0	740.00	+18
G	2604	1	0	1	0	0	0	1	0	1	0	1	1	784.05	+76
G'	2458	1	0	0	1	1	0	0	1	1	0	0	1	830.62	+14
A	2320	1	0	0	1	0	0	0	0	1	1	1	1	880.03	+33
A'	2190	1	0	0	0	1	0	0	0	1	1	0	1	932.27	-64
B	2067	1	0	0	0	0	0	0	1	0	0	1	0	987.74	-23
Xtal	240	0	0	0	0	1	1	1	0	1	1	1	1	4166.67	
P.I.I.	490	0	0	0	1	1	1	1	0	1	0	0	1		



outputs  $Q_1$  to  $Q_{12}$  represent counts of 1, 2, 4, up to 2048 as shown, and to give a division of  $M$ , the outputs connected to  $N$  total  $M-1$ . In table 10, each 1 represents a diode to  $N$ , and the reference signals are taken from outputs having a mark to space ratio nearest to unity.

Fig. 38 includes 4016 switches for the least significant digits, and with  $S_1$  and  $S_8$  closed as shown, the division is 490. Appropriate switching produces divisions from 485 to 495 which varies the p.l.l. output frequency, and so tuning of the organ from 1% below to 1% above standard pitch in 0.2% steps. The diode connections for generating the 4016 control inputs from a single-pole switch are shown in Fig. 39.

In construction, stray capacitance should be minimised, especially at  $N$  and  $R$ , by mounting the diodes vertically. Current consumption of the prototype generator was 72mA.

### Reproduction

For good reproduction hi-fi equipment may be satisfactory although much depends on the room characteristics. To obtain the benefit of reflected sound, the listening room should not be unduly damped by soft furnishings, and the h.f. speakers should not directly face the listener. Transient response is not of primary importance, and some low-order distortion can be tolerated. The main requirement is a reasonably smooth amplitude/frequency response from the loudspeakers. This can be compensated by adjustment of  $R_n$  values, so that individual notes do not stand out or disappear. The better types of small loudspeaker units, although they often have a resonance at around 100Hz, are fairly smooth at higher frequencies. The system adopted for the prototype used an active crossover at 270Hz with a pair of Bailey transmission line speakers, fitted with bass drivers only. The  $R_n$  values were adjusted, mainly in the lowest octave, to give the best regulation with the speakers in their final positions. The largest adjustment needed was -8dB at 55Hz.

A pipe organ has as many separate sound sources as there are pipes, and it is advantageous to provide separate h.f. channels as shown in Fig. 40, which also allows independent processing if required. Fig. 41 shows a low-pass filter using the Rauch configuration which can also serve as a mixer if the signal input sources are of low impedance. The 62k $\Omega$  resistors are each three times the calculated input resistor value, and the gain is reduced correspondingly. The filter is maximally flat and the overall passband gain is unity for each input. The high-pass filter in Fig. 42 also has unity gain and is third-order with a passive output section with the same corner frequency. Both filters operate from a single 12V supply and the working point is set by a suitably decoupled 6V bias. Fig. 43 shows how to adapt the

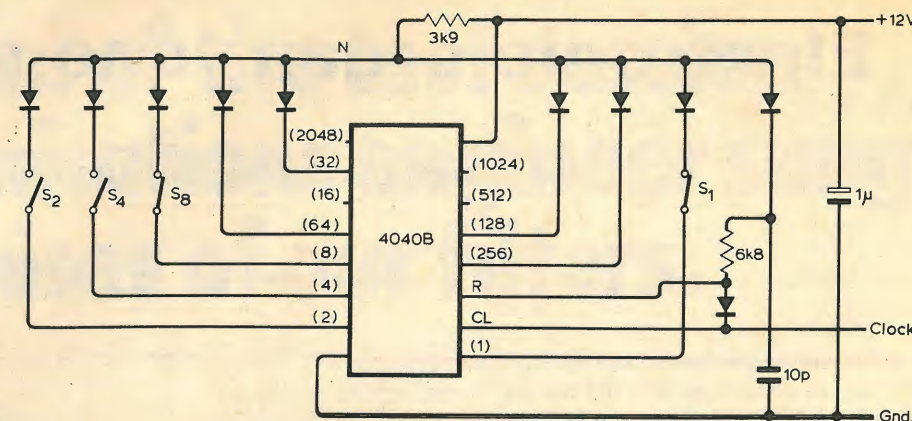


Fig. 38 Divider circuit. The switches represent 4016 sections used for the optional tuning facility, and apply to the p.l.l. comparator divider only.

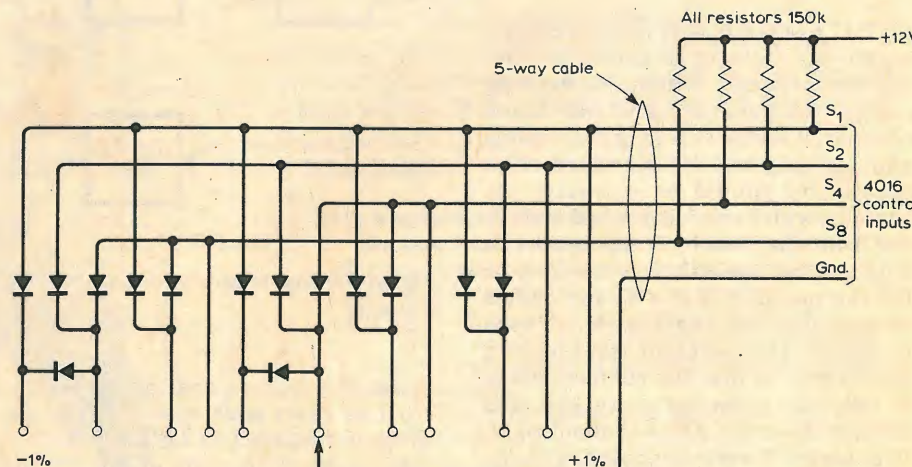


Fig. 39 Connections for an 11 position tuning switch. The 14 diodes are mounted on the switch wafer.

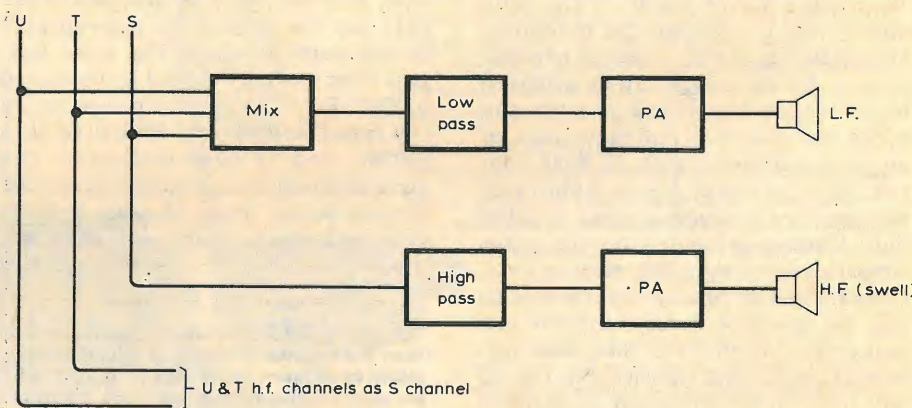


Fig. 40 Output channels. The prototype cross-over frequency is at 280Hz.

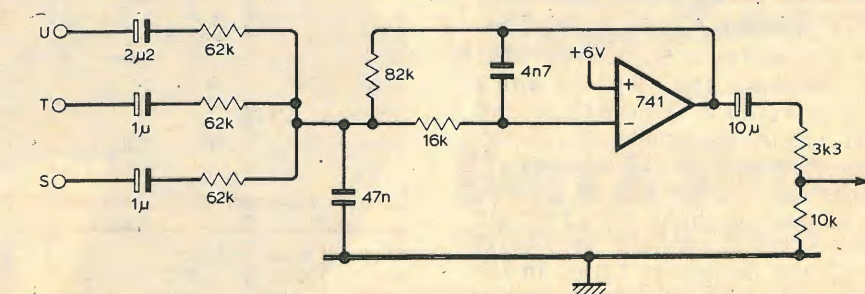


Fig. 41 Mixer and Rauch 12dB/octave l.p. filter. The 10k $\Omega$  resistor may be replaced by a potentiometer for channel balancing.

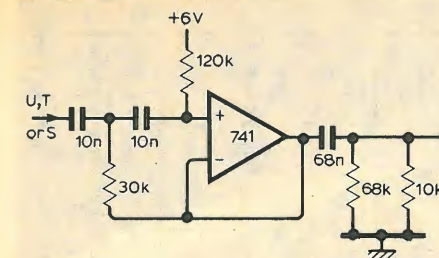


Fig. 42 Sallen and Key 18dB/octave h.p. filter. The output RC section is 3dB down at 270Hz to produce a maximally flat response with an active section  $Q=1$ . The 10k $\Omega$  resistor may again be replaced by a potentiometer for channel balancing.

circuits for individual requirements. The circuit values should be with 5%.

### Coupling

As discussed in part 1, coupling is useful both as a playing aid, and to extend the harmonic spectrum. In the prototype eight couplings were used, four from the pedal keyboard — unison (U1U), octave (U2U), great (T1U) and swell octave (S2U), and four from the great keyboard — unison (T1T), octave (T2T), swell (S1T) and swell twelfth (S3T). The unison-off couplers to be described are a necessary part of the design, but also have their own uses. For example, the pedal can be played at 8' pitch only or, using the swell octave, at 4' only on the great. If the 8' sound is taken from the swell only, the upper work provided by the great octave may use a different harmonic mixture as well as a separately-processed channel. The above list is not exhaustive, and other useful couplings could be provided, such as S2T, S2S, and sub-octave couplers such as S1/2, and S2/2T. A super octave S4T or S4U could be used for extra brilliance, but this would need to break back to the octave at G'4. The situation is similar to that of a small extension or unit organ which uses a limited number of ranks at several pitches, with the additional facility of coupling at reduced strength where required.

The most direct method of coupling is to use additional key-contacts, with diodes to prevent switching of unselected common lines when the same key is pressed on both keyboards. Semiconductor switching avoids the practical problems of multiple contacts, and reduces the interconnecting wiring. In Fig. 44, UCL1 is the contact operated by the C1 pedal, and UCK1, UCK2, etc are keying inputs on the C gate. (In the basic system, ICL1 is supplied from +5.6V and is only connected to UCK1. A particular coupling is selected by energising the appropriate common line, and its strength is controlled by the applied voltage. Allowing for the diode

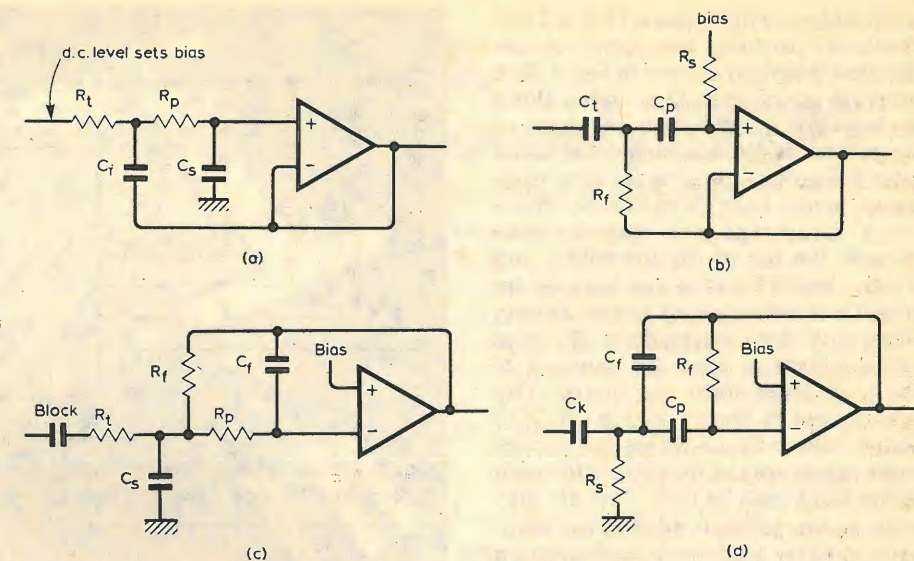


Fig. 43 Filter design using available capacitor values. (a) Non inverting Sallen and Key l.p.f. with unity gain. Select  $C_f/C_s = a$ , where  $a \geq 2$  for  $Q=0.707$ ,  $\geq 4$  for  $Q=1$ .  $\sqrt{X_f X_s} = R$ , say,  $R_t = R\sqrt{b}$  and  $R_p = R/\sqrt{b}$  or vice versa where  $b = (a-1 + \sqrt{a^2-2a})$  for  $Q=0.707$ , and  $b = \frac{1}{2}(a-2 + \sqrt{a^2-4a})$  for  $Q=1$ .

(b) Non inverting Sallen and Key h.p.f. with unity gain. Select  $C_t = C_p$ .  $X_t = X_p = R$ .  $R_f = R/2Q$  and  $R_s = 2Q.R$ .

(c) Inverting Rauch l.p.f. Select  $C_s/C_f = p$ , which predetermines design gain  $G$ .  $G = \frac{1}{2}(p-1)$  for  $Q=0.707$ ,  $G = \frac{1}{4}(p-4)$  for  $Q=1$ ,  $R_f = X_f/2Q$ ,  $R_t = R_f/G$ , and  $R_p = R_f/(1+G)$ .

(d) Inverting Rauch h.p.f. Select  $C_t/C_f = G$  and  $C_p = C_t$ . Then  $R_f = d.X_t$  and  $R_s = G.X_t/d$ , where  $d = (2G+1)/Q$ .

Fig. 44 Coupling from C1 pedal key.

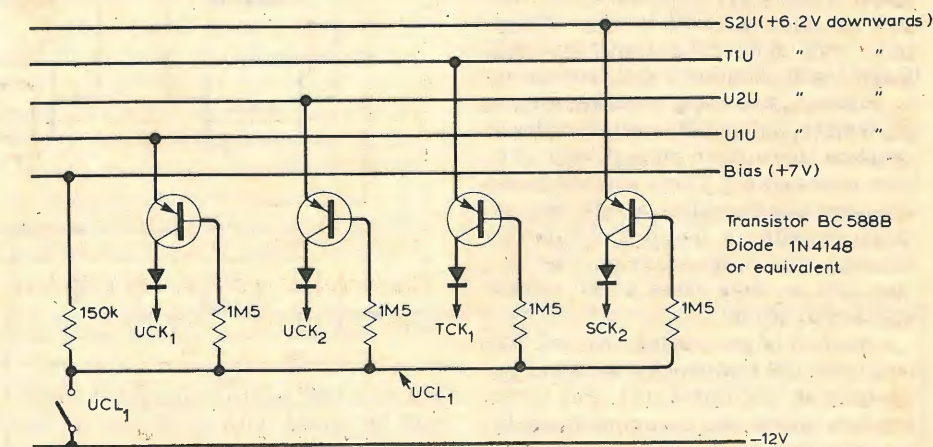


Table 11 Summary of couplers used in the prototype.

Coupler	U1U	U2U	T1U	S2U	T1T	T2T	S1T	S3T
Start	UCK1	UCK2	TCK1	SCK2	TCK1	TCK2	SCK1	SGK1
Ends	UGK3	UGK4	TGK3	SGK4	TCK6	TGK6	SGK6	SGK6
Gates	32	32	32	32	61	56	61	49
Groups	C-G	3 groups each	32		C	6 groups	61	
	G-B	2 groups each			C-B	5 groups each		



the prototype which uses a 17.9" x 7.05" Veroboard to form the plinth of the complete assembly shown in Fig. 1. The gates are arranged in 12 columns along the length, and in the same sequence as the gate cards. Within each column the gates are grouped in octaves with each group having four gates except where the T couplings are discontinued towards the top of the keyboard, and where the S3T gates are located by output position as noted above. Keying interconnections are made in 33 s.w.g. which is kept in place by loops of 24 s.w.g. soldered onto the board. The basic layout is shown in Fig. 46. The dotted lines indicate where the copper strips below are cut except for through connections such as bias. Cuts are also made at the points e, so that the transistor emitter leads can pass through without making contact. These leads are connected underneath by runs of 33 s.w.g. to the buses U1U, U2U, etc. It is convenient to use the outermost plain strips of copper on the veroboard for this purpose. The resistors are mounted vertically with appropriate L connections made to their upper ends. Terminal pins are used for K connection points, and most of the diodes can be accommodated with suitable track-cutting, in the area to the right of the transistors, the remainder being mounted on the copper side. If the swell keyboard is supplied with a fixed +5.6V, it is not necessary to use diodes between the swell keyboard contacts and swell K inputs. The T1U and S2U diodes are mounted with the U groups, and the connections to TK and SK pins are made with 33 s.w.g. on the copper side. In Fig. 45 the C column is at the left, and the pedal groups are towards the top.

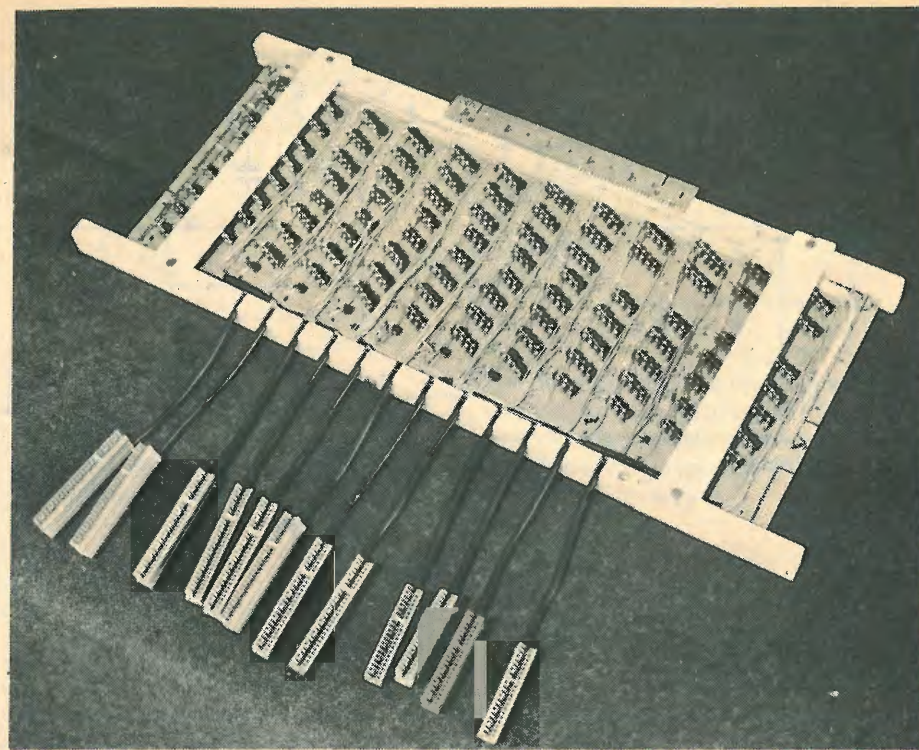


Fig. 45 Assembled coupling circuits from prototype. This section forms the plinth of Fig. 1, and shows the edge connectors used to bring keying signals to the gate cards.

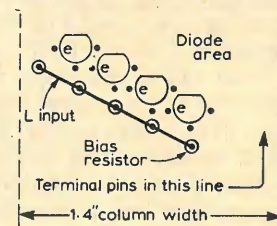
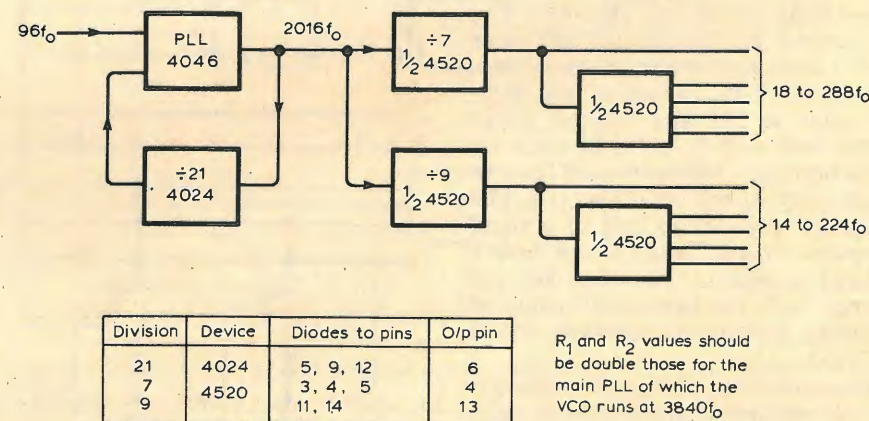


Fig. 46 Transistor and resistor layout of coupling groups viewed from the component side.

Fig. 47 Generation of the 7th and 9th harmonics.



### Tonal variations

Apart from what is needed for the basic system, the EO1 card provides eight spare input and output positions, 16 spare gate positions, and four spare i.c. positions, including tracking for an additional p.l.l. The EO2 card provides a complete spare filter section, and two spare connections. These unused positions can accommodate a selection of additions without using more cards, although wire connections may be needed, and in some cases a few tracks will have to be cut.

A method of generating 7th and 9th harmonics (60 frequencies each in Fig. 47), uses an additional p.l.l. and three counters which are accommodated in the spare gate-card positions. The 96f<sub>0</sub> output in Fig. 20 is used as a reference, and the non-binary divisions are obtained by premature reset as in Fig. 12. The  $R_n$  values and filter components may be the same as in the 6th and 8th harmonics respectively because a somewhat lower output is acceptable for these pitches. The 7th can be taken up to GK5, 10.98kHz, and the 9th to EK5, 11.8kHz. The 1.5f<sub>0</sub> output in Fig. 20

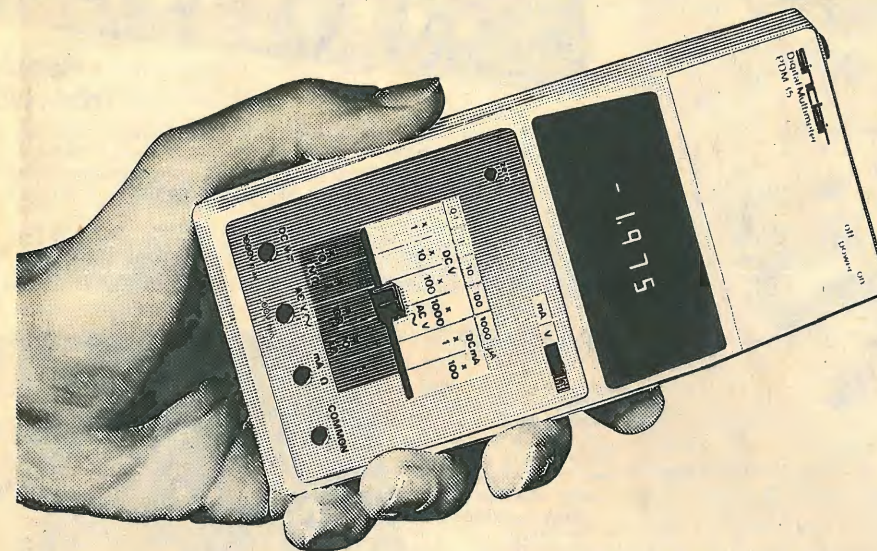
enables a 10% bus to be provided which may be mixed with a 16% bus in the lowest octave to generate a subjective 32' bass as sometimes found in pipe organs. Levels will need adjustment note by note, and additional low-pass filtering may be beneficial. Ideally, a separate speaker channel should be used for the quint.

In Fig. 26, if  $C_3$  is omitted the filter slope becomes 6 dB/octave. If the resonant frequency of the first stage is then raised to the working band, the

action becomes similar to that of the low-Q LCR filters used in subtractive designs, with  $R_6$  providing the main control of Q. Input  $R_n$  values will need to be graded differently, varying less widely, and more convenient levels will result from reducing  $R_3$  to say 47k $\Omega$ . Further tonal modification is possible by re-introducing a smaller value of  $C_3$  and/or reducing  $C_2$ . The resultant tone may be supplemented by specific harmonics from the sinewave buses.

To be continued

## The Sinclair PDM35. A personal digital multimeter for only £29.95 (+8% VAT)



### Now everyone can afford to own a digital multimeter

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The PDM35's 1% of reading compares with 3% of full scale for a comparable analogue meter. That makes it around 5 times more accurate on average.

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**DC Volts (4 ranges)**  
Range: 1 mV to 1000 V.  
Accuracy of reading: 1.0%  $\pm$  1 count.  
Note: 10 M $\Omega$  input impedance.  
**AC Volts (40 Hz-5 kHz)**  
Range: 1 V to 500 V.  
Accuracy of reading: 1.0%  $\pm$  2 counts.  
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Range: 1 nA to 200 mA.  
Accuracy of reading: 1.0%  $\pm$  1 count.  
Note: Max. resolution 0.1 nA.  
**Resistance (5 ranges)**  
Range: 1  $\Omega$  to 20 M $\Omega$ .  
Accuracy of reading: 1.5%  $\pm$  1 count.  
Also provides 5 junction-test ranges.  
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**Weight:** 6½ oz.

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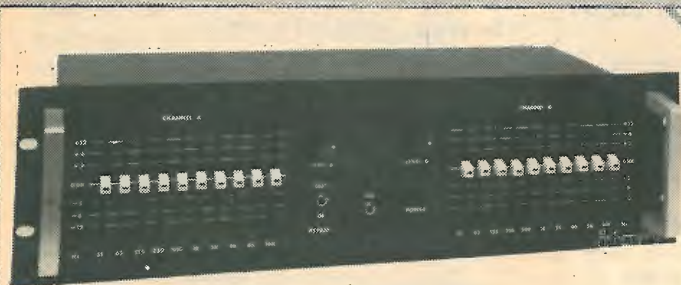
Please send me \_\_\_\_\_ (qty) PDM 35(s)  
@ £32.35 (incl. VAT) each..... £..... Name.....  
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# NEW PRODUCT BULLETIN



## ITAM Graphic Equalizer

A fully professional graphic equalizer at a reasonable price. Stereo, with 10 bands per channel, 19 inch rack mounting. £280 + VAT.

## OTARI DP4050 OCF IN CASSETTE DUPLICATOR

Now with 3% + 7% master capability and Ferrite heads. Duplicates 6 cassettes each run at 8 times speed. Over 80 stereo C60 per hour. Unquestionably the finest in cassette copier available from the world's largest duplicator manufacturer.



## ORBAN from USA

Dual channel multispring reverb unit. Each channel features four springs — far smoother than single spring systems. "Twang" and "boing" are virtually eliminated by incorporating a floating threshold limiter. Bass, mid-range EQ and bandwidth controls. The best compact reverb unit available.

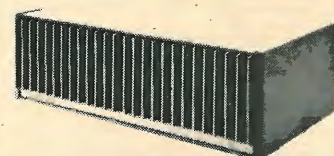
## ITAM 882

New Stereo Mixer, built to a specification not a price! Ultra low noise, —128dB. 8 inputs + 8 direct outputs, 2 outputs with VU meters. 2 limiters. XLR mic inputs (balanced). 3 band EQ + mid sweep. £395 + VAT



## AMPEX ATR-700

Now every studio can afford legendary Ampex performance and reliability. Fully professional specification including balanced inputs/outputs, Cannon connectors, variable tape speed, sel sync. The price will fit this year's budget, not next year's! Sole distribution by ITA.



## QUAD

The new 405 power amplifier is now in stock. 100 watts per channel — simply the best, for £££'s less! Immediate Delivery!

## Amateurs and plasma bubbles

A new theory to account for the 1977-78 transequatorial-type propagation experienced for the first time on the 144 and 432MHz bands has been put forward in Europe and the United States. J. Roettger, DJ3KR, in *Radio Communication* suggests that the basic mode is scatter propagation due to plasma bubbles which occur in equatorial spread-F conditions and rise to altitudes of 400 to 1000km. Joseph Reisert, W1JR, and Gene Pfeffer, KOJHH, in QST put forward a generally similar concept, although they have adopted the term FAI (field aligned irregularities) to describe this propagation mode. It is clear that the amateur work that resulted in these long-distance contacts on 144MHz (and a few reception reports on 432MHz) in South and Central America, in Europe and Africa, and in Japan and Australia have uncovered propagation modes previously unobserved and unexplained. The prospects in favourable countries for more late-evening long-distance contacts using these plasma bubble modes during the present peak of the sunspot cycle appear promising.

## Is amateur radio booming?

Although the number of American amateur licences is now at an all-time high of over 365,000, the American Radio Relay League with 170,000 members has recently reported a significant decline in the demand for its technical publications. As a result it has been forced to cut back on expenditure, including trimming some 20 people from its staff. The current inflation rate in the United States is thought to contribute materially to the difficulty ARRL is experiencing in breaking even, and the hardening yen/dollar exchange rate is also expected to have an effect on equipment sales this year.

However a rather different note is struck by *Electronics* staff writer Vincent Biancomano, WB2EZG, in an article "That boom you hear is ham radio." He forecasts an expanding market to meet the needs of the 28,000 additional amateurs licensed in 1978. Although, he says, the "old names" like Collins Radio, National Radio Co, Hallicrafter Co and Hammarlund "are all but gone," arisen to take their place are such firms as Trio-Kenwood and Yaesu Musen both of Japan and new American firms such as Dentron, Wilson Electronics Corporation (a new aerial firm), Lunar Electronics (v.h.f. amplifiers) etc. He notes that interest is shifting upwards to include u.h.f. as well as v.h.f. factory equipment.

He suggests that the gap between citizens' band radio and amateur radio is closing and that "those who would wish to preserve the spirit and self-image of amateur radio must now deal



with the consequences of popularity." Official FCC c.b. licences have reached the staggering total of 14-million-plus with more than one million licensees in Texas alone. Some 2.5-million new licences were issued during 1978.

Dr Dain Evans, G3RPE, president of RSGB, following a recent visit to talk to amateurs in the United States has returned convinced that it is high time that the Americans renewed their former interest in microwaves. Whereas, he points out, a few years ago American amateurs held five out of the six "world microwave records" they now hold only one, despite the almost ideal topography and high incidence of ducting. Although a number of American amateurs have built 30ft. dish aerials for moonbounce etc, he feels that they just do not succeed in putting it all together so far as co-ordinated microwave activity is concerned, and contrasts this with Europe where activity on 10GHz is rapidly increasing.

What may have been the first 10GHz amateur contact to be effected by tropospheric scatter was made between G3JVL, Hayling Island, and G3YGF/A at Oxford, a distance of 110km, last October. Both stations used s.s.b. by means of transverter techniques to obtain p.e.p. outputs of 6 and 15 watts.

## Here and there

Problems that have arisen over the past year in using the predicted orbital data for Oscar 8 are now believed to have been due mainly to errors caused by radar tracking of the nose cone of the rocket in mistake for the amateur communications satellite, and also due to the greater effect of "drag" caused by the lower orbit of Oscar 8 compared with Oscars 6 and 7. However by autumn 1978, Amsat-UK were confident that their revised estimates were sufficiently accurate to allow them to issue a revised list of orbital data.

"I know that c.w. is obsolete and dead. I was told this when I first started to learn the code in 1947. Nevertheless there are a large number of stations who apparently do not know that c.w. is

dead, for they persist in using it." — John H. Smith, VK3IQ.

Canada has now started issuing a new "Amateur Digital Radio Operator's Certificate" which gives the holder operating rights on 144MHz and above basically similar to the British Class B licence. The U.S.A. remains one of the few countries where no "no-code" licences are issued but has a novice licence requiring a code speed of only about 5 words per minute. Some American novices start early: Neil Rapp, WB9VPG (formerly WN9VPG) held an FCC novice licence at the age of 5 years, moving on to a general licence at six years of age.

## In brief

The South African 144MHz beacon station ZS6DN was positively heard and identified by SV1DH in Athens between 1715 to 1725 GMT on November 5, a distance of over 7100 km. . . . Tropospheric openings on November 6 brought 144MHz contacts between Scotland and Austria, the distance of some 1450km being among the best tropo contacts by British amateurs during 1978. . . . Sunspot activity declined sharply during most of November but should peak up again early in 1979. . . . Naomi Uemura, JG1QFW who made a remarkable solo trip to the North Pole during the summer of 1978 depended for much of his communications on amateur stations. . . . Charles D. Tandy, the Fort Worth millionaire who created the Tandy Corporation and associated Radio Shack chain from a faltering group of amateur radio stores in the Boston area, has died. . . . The RSGB 1979 National VHF Convention is to be held on March 10 at The Winning Post, Twickenham, and will include lectures on "Slow scan television" (Grant Dixon), "Tropospheric propagation" (Ray Flavell) and "Sporadic E" (Professor Martin Harrison) as well as the usual convention events. . . . A special 10th anniversary "Worked All Britain" award (the "10 x 10 Decade Award" is being offered for contacts made this year; any profits will go to the Radio Amateurs Invalid and Bedfast Club. . . . Some 550 amateurs attended the 1978 Scottish Amateur Radio Convention in Aberdeen. . . . The high cost of the British maritime mobile amateur (£16.40) is to be reviewed by the Home Office. . . . It is now hoped that the extremely high power radar transmissions (Pave Paws project) planned for near 432MHz may be moved elsewhere, the system operates on a number of discrete frequencies. . . . A total of 104 v.h.f. repeater licences has now been issued by the Home Office to the RSGB and a number of new repeaters have been brought into operation recently. . . . Larry Le Kashman, W2IOP, a former editor of CW and a prominent contest operator, has died.

PAT HAWKER, G3VA



1-7 Harewood Avenue, Marylebone Road, London NW1.  
Tel: 01-724 2497 Telex: 21879

FRANCE: Son Professionnel, 2 Rue des Tennerolles,  
92210 Saint Cloud (Paris). Tel 602 6815.

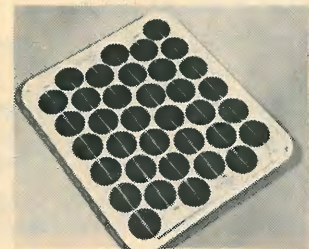


# NEW PRODUCTS

Professional readers are invited to enter codes on the reply-paid card bound in at pages 116/117

## Solar power panel

The MST300 solar power panel contains 36 silicon cells, each 3 inches in diameter, series connected to give an output of 1.1 amps at 14.4 volts. It measures 560mm x 480mm and is 130mm deep. Its aluminium construction serves as a heat sink and makes the module suitable for use in high ambient temperature zones. A recessed aluminium extrusion, into which the base plate and cover are fitted, provides a hermetic seal and prevents moisture from getting into the resin filled



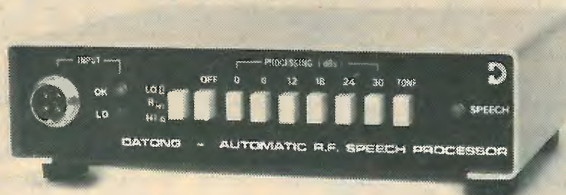
WW301

space containing the silicon cells. The panel has a protective cover of fibre reinforced polyester. Additional silicon cells can be incorporated as required by the customer. Ferranti Electronics Ltd, Fields New Road, Chadderton, Oldham OL9 8NP.

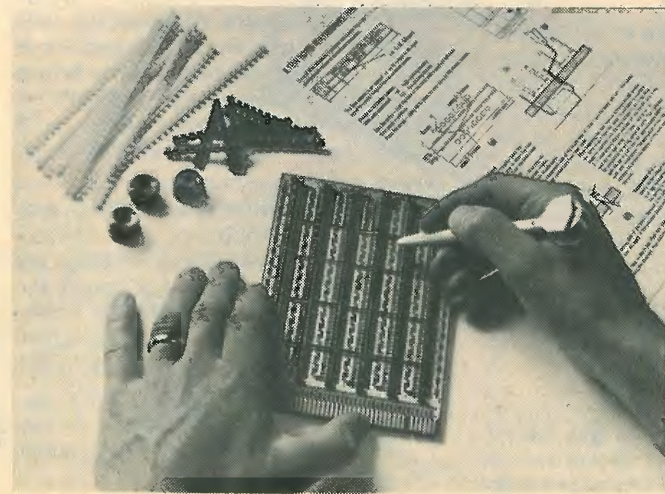
WW 301

## Speech processor

Intended for s.s.b., f.m. and a.m. radio systems, telephone and public address systems and similar applications, the model ASP speech processor is designed to improve the effectiveness of peak-power limited speech equipment, firstly by allowing an increase in the average radiated power while maintaining a constant peak power, and secondly by increasing the intelligibility of speech in noise. Taking an audio signal input, it generates a s.s.b. signal. This is clipped, filtered and demodulated back to audio, giving instantaneous compres-



WW302



WW303

sion of dynamic range "without harmonic distortion." An unusual feature is an audio a.g.c. system before the main processor. This adapts to changing input levels and maintains the degree of speech enhancement at push-button-selected values from 0 to 30 dB in 6dB steps. Datong Electronics Ltd, Spence Mills, Mill Lane, Bramley, Leeds LS13 3HE.

WW 302

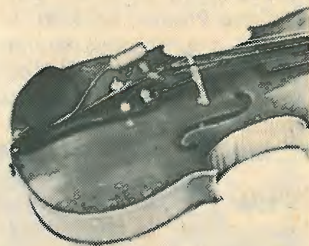
## Prototype circuit board system

The Roadrunner prototype wiring system uses a special wiring instrument and press-fixed or glue-fixed distribution strips. The wiring instrument feeds a quick soldering enamelled wire from interchangeable bobbins. Special features are a simple threading system which allows fast bobbin changes and a means for adjusting wire tension. The castellated distribution strips will retain a large number of wires in position without affecting the low profile of the finished boards. The strips have no posts to impede access when wiring. The system is normally supplied in kit form but individual components are available separately. A typical introductory kit costs £8.50. TJB Associates, Unit 116b, Blackdown Rural Industries, Haste Hill, Haslemere, Surrey GU27 3AY.

WW 303

## Musical instrument microphone

The SM17 moving-coil microphone is designed for attaching to musical instruments in recording studios and for on-stage professional sound reinforcement. Based on the makers' SM11 microphone, it is supplied with a 10ft cable and three types of mounting. The first is an expansion mount for fitting to the tailpiece of instruments of the violin family, while the second is an edge-mounting clip for attachment to a guitar sound-hole or to a suitable edge of a brass instrument. The third mounting is a complete lavalier assembly, with a cord and cable



clips. The microphone is claimed to give good isolation from other instruments and freedom from feedback. Price excluding v.a.t. is £45.60. Shure Electronics Ltd, Eccleston Road, Maidstone, Kent ME15 6AU.

WW 304

## L-band magnetron

The M5169 tunable magnetron is claimed to have better performance than that of fixed frequency tubes in current service for high power surveillance radars. A successor to the maker's M554/M586 series, it has been designed to be electrically and mechanically interchangeable with them. The design, which includes a rigid cathode structure and 15MHz tuning mechanism, offers a stable m.t.i. performance in microphonic environments, precise setting of radar operating frequency, and a.f.c.-tuning to crystal controlled local oscillator. Frequency range is 1260 to 1365 MHz, while output power is 2.6MW (typical). English Electric Valve Company Ltd, Waterhouse Lane, Chelmsford, Essex CM1 2QU.

WW 305

## Error rate tester

The bit error rate in digital communication, fibre optic and magnetic storage systems can be measured at data rates up to 150Mb/s by two new modules from Tau-Tron. The MN-301A transmitter and MB-301A receiver operate at rates from 1Mb/s to 150Mb/s and use pseudo-random data patterns as test signals. Two sequence lengths are used: 127 bits and 32767 bits. In addition to these, the transmitter will generate an alternating 1100 . . . . . test pattern. Errors can be injected

into the pseudo-random data internally at a rate of 2 errors per 100 bits of data or at an externally determined rate. Output data and clock amplitude is at 1 volt into 50 ohms and gives compatibility with e.c.l. Bit-by-bit error detection, using an internally generated reference pattern, is provided by the receiver. Synchronization may be manual or automatic. Bit error rate or cumulative errors may be shown by the four-digit l.e.d. display. The instrument has a variable display hold time and a b.c.d. printer output. Tau-tron Inc, 11 Esquire Road, North Billerica, Mass. 01862, USA.

WW 306

## Colour tv camera

The CY-8800E colour camera made by JVC uses three tubes, 2/3in Plumbicons or Saticons, with electromagnetic focus and deflection. It has a 1.5in adjustable electronic viewfinder and a 10-to-1 zoom lens with automatic iris control. Minimum illumination required is 300 lux, f/1.9. Signal-to-noise ratio of 49dB is achieved at 3,000 lux, f/4, with centre resolution in excess of 500 lines. For low light levels, the camera has a sensitivity switch with +6 dB and +12 dB steps. There is circuitry for horizontal and vertical contour correction, auto white balance and colour temperature adjustment. A filter system provides for different light levels, and an intercom system for studio use. The lens aperture can be controlled by a built-in level switch, giving 1/2 f-stop adjustments, and through a built-in video signal indicator. Other features are a colour bar generator, a battery warning indicator, a tally light and fast warm-up capability. On the hand grip are a recorder stop-start switch, an open/close switch for the zoom lens and an "aux" video switch. Power is supplied through an a.c. adaptor or external +12V d.c. source. Bell & Howell A-V Ltd, Alperton House, Bridgewater Road, Wembley, Middx HA0 1EG.

WW 307

## Pulse oscillator / amplifier

A high-power amplifier/pulsed oscillator is available for applications requiring r.f. pulsed or c.w. signals. The pulsed r.f. output is continuously variable up to 400 watts peak and covers the frequency range 0.18 to 190 MHz. Called the PG-650-C, it can amplify and modulate externally generated signals which are already phase, frequency or amplitude modulated. Four models are available. Other features are: continuously variable pulse durations from 1.5 to 20µs

or fixed at 0.5µs; r.f. phase stability of about 1ns with respect to trigger; small to "unmeasurable" interpulse noise; and a.c. line input regulation and positive or negative adjustable trigger level. Arenberg Ultrasonic Inc, 94 Green Street, Jamaica Plain, Mass. 02130, USA.

WW 308

## Video monitors

The PMC 50/S/RGB colour video monitor has a 20-inch screen and the PMC/35/S/RGB monitor a 14-inch screen. Both use red, green, blue video input signals with separate mixed syncs inputs and are intended primarily for computer graphics. They have phosphor-stripe cathode-ray tubes, capable of resolving 490 lines per picture width. Beam separation is claimed to be better than 0.5mm vertically and one stripe width horizontally. The monitors have wide-band gain-stabilized amplifiers (-1dB at 9MHz). Differential gain distortion is less than 2% and the black level stability is better than 1% for a 100% black to peak white change in picture level. Size stability of the raster is better than 5%. The monitors can be set for 525-line or 625-line scanning standards without modification. Raster geometry controls are accessible, and there is a built-in test signal source, providing six test patterns. Cotron Electronics Ltd, Rockland Works, Eagle Street, Coventry CV1 4GJ.

WW 309

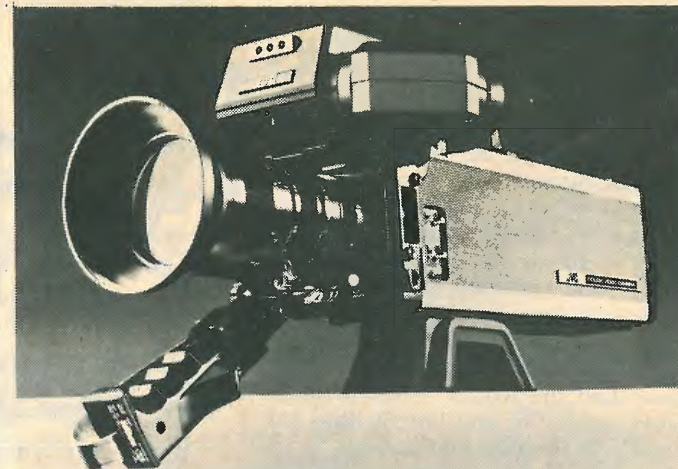
## Data acquisition device

A self-contained data-acquisition module will multiplex 16 channels of physical variables from sensors and convert them to digital information. Known as the Zeltex SMP1000, the device offers 12-bit resolution at rates of up to 50,000 channels per second. It can handle 16 single-ended or eight full differential inputs with a voltage range of ±10V. Housed in a case measuring 11.68 x 7.62 x 0.95 cm, the device has an input impedance of 100MΩ and an analogue/digital conversion time of 12µs. Inputs and outputs are compatible with t.t.l. circuitry. The operating temperature range is 0°C to +70°C. Walmore Electronics Ltd, 11-15 Betterton Street, Drury Lane, London WC2H 9BS.

WW 310

## Epoxy impregnant

Eccoseal W19-FR is a fire retardant epoxy impregnant. The low viscosity of the material (2-5 poise) makes it suitable for impregnating coils, transformers and other electronic components.



WW307



WW308

When tested in accordance with the underwriters' laboratory VL94 vertical burning test it meets the V-O specification in 1/16in thickness when cured with catalyst 11. It can also be used as a casting resin as it contains no solvents. Large castings can be produced in conjunction with filler A-21 or other types. It can be cured with catalyst 9 to give a room temperature cure or with catalyst 11 for high temperature properties. Emerson & Cuming (UK) Ltd, Colville Road, Acton, London W3.

WW 311

## Headset tester

A headset and audio accessories tester provides means for subjective evaluation of equipments having electromagnetic or carbon microphones with outputs in the range 1-300mV. Earphones or receivers are checked using speech side tone or an internal oscillator. Visual indication of microphone output level is given by a meter, which also allows more precise microphone measurements to be made using a small artificial voice supplied with the instrument. A total of six combinations of electromagnetic and carbon systems (3 each) can be tested on one

unit. Connections and sensitivities can be specified by the user. Isolation between microphone, receiver and exposed metal parts is shown by an l.e.d. indicator. The unit has its own self-checking facility and is designed for use by both skilled and non-technical people. Tone Dynamics Ltd, 110 Midland Road, Luton, Beds LU2 0BC.

WW 312

## Analogue function generator

Described misleadingly by its makers as a "linear microprocessor" the LH0094 packaged circuit generates an analogue output voltage  $E_o$  given by the equation:  $E_o = E_1 \times (E_2 / E_3)^m$ . The inputs  $E_1, E_2, E_3$  can be any voltages from 0.1V to 10V while  $m$ , a resistor ratio, sets the power in a range from 0.1 to 10. A thin film network included in the circuit allows squaring and square rooting by strapping pins. Accuracy is claimed to be 0.05%. Housed in a hermetic 16-pin package, the device operates from power supplies of ±5V to ±22V, and has a 10kHz bandwidth. National Semiconductor Ltd, 301 Harpur Centre, Horne Lane, Bedford MK40 1TR.

WW 313



## Fashion waggon

My first car was pretty basic. It had a speedometer and a petrol gauge and that was about it. As far as I was aware, all kinds of mayhem were going on under its bonnet, but the idea of concerning myself with temperatures, charging rates, r.p.m. and all that stuff didn't come high on my list of priorities. It did 60 m.p.h. downhill, with a following wind and what more could anyone want? (This is all a long time ago.) Then, of course, I entered the rat race and got hold of a year-old Zephyr, which had the works — temperature gauge, ammeter, and lots of little coloured lights that would have been invaluable if I'd had the instruction book.

I think my first grey hair showed up around that time. Now I knew that the engine had a temperature, I began to watch it, worrying when it approached 'H'. If the ammeter didn't do the right thing at the right time I knitted my brow. Driving became a race between the engine seizing solid and the battery going flat, neither of which contingency could have caused me a moment's concern without the instruments.

Now, the microcomputer is with us, of course, and all the more private and personal foibles of one's engine can be displayed. Cadillac's new Seville uses a computer to increase the driver's worry level to new heights, exposing to his horror-stricken gaze all manner of vital information without which he would have been driving peacefully along, humming nonchalantly. It is all available 'on request' by pushing a button.

Well, I suppose it's all right for the man who has everything; but the thought of anyone actually pushing a button to ask the computer for the inside story on how the oil is doing, or how long it is since one left home is an odd one. When are you supposed to consult the thing, I wonder. Personally, I'm usually fairly preoccupied with staying out of trouble and when the road is reasonably clear I listen to music. At no time, I can state with confidence, does it occur to me to do any more than glance quickly at the oil pressure, temperature and fuel level every ten miles or so, and that's during the break between movements. If I had a Cadillac, I would feel entitled to assume that everything would work correctly without supervision from me.

## Future uncertain

I'm a bit worried. Not about Jim, although heaven knows he can be very trying, but more about our inescapably microscopic future. Every time we get a note from a company to tell us about their latest bit of nonsense, it's become pocket-sized or even credit-card-sized. If it goes on like this, we'll be walking around with pockets bulging to such an extent that we'll need handbags to relieve the strain.



But, really, this fascination with smallness is a bit of a puzzle. A telly with a two-inch screen is a pretty good piece of engineering when you consider the low power needed and the size of the case, and the performance is very good, too, but watching a two-inch diagonal picture holds about as much attraction, for me at least, as shark fishing in the bath. Minute calculators, too, aren't a lot of use if your fingers are too big to press one key at a time. If you have to find a pencil to prod them with, you may as well do the calculation with it.

The way round all this, of course, is to go in for a programme of forced evolution — or genetic engineering, to give it its more with-it title. It is becoming fairly clear that all this machinery we are faced with is not going to take kindly to being messed about by clumsy, error-prone, unadaptable and altogether regrettable humans.

So, equally clearly, we are going to have to change. Twenty-six very small fingers on each hand might be a good first objective, to conform with computer keyboards; and if we could only buck our ideas up a bit to conform with a decent clock rate we could save an awful lot of time and expenses in interfacing. We can't let people stand in the way of progress; either they pull themselves together or they will have to go.

## Jack and his master, both

Using all the tact and diplomacy of a wounded buffalo, the IEE recently addressed itself to the economic and social problems of the day. The Institution aligned itself with views expressed by the CBI on the role of trade unions and delivered itself of the following Thought. "... the Institution commends to the Trade Union movement and industry as a whole the standards of service and responsibility to the community displayed by the professions as being now the standards appropriate to so powerful a movement."

There are at least three assumptions in that one sentence, any one of which ought to send the blood pressure of a reasonably observant trade unionist up to somewhere near the danger level. First, what right has the Institution, admirable a body as it is, to commend

anything to anyone? Second, the implication that the professions are exemplars of responsibility and service to the community is hardly likely to find a sympathetic response from the public, who are accustomed to being treated like inconvenient morons by the medical profession (which strikes when it wants to, like anyone else) and as part of a game of poker by the legal brotherhood (which doesn't strike because it is doing very nicely out of its clients as it is). Thirdly, assuming the IEE meant to say that the professions exhibit high standards, are they therefore implying that the unions don't?

Oh, dear! How very unfortunate. Perhaps if the Thought had been addressed a little more wholeheartedly to management as well as unions (the 'industry as a whole' is fairly clearly an afterthought) it would have carried more conviction. Management's grief at being restrained by Government from paying employees the rewards they would 'like to' must call for a good deal of acting ability in public, and an act as cynical as that is surely not in the tradition of professional behaviour.

## Calling the tune

Half the fun of listening to the radio, I've always thought, is choosing your programme and tuning it in. The pleasure of catching and savouring the ether-borne sounds — Freud might have called it aural eroticism — is greatly intensified by all the excitement and uncertainties of the chase. So what am I to make of the latest technological marvel they're now working on that will do away with most of this? It's called programme labelling. Of course, it's put forward in the name of progress — that brilliant arrow which points towards a marvellous collective future but forgets to indicate that each individual goes through a drastic molecular change on the way. But I suspect it's really just a plot on the part of the i.c. merchants, aided by the broadcasters and the desperate set makers, to sell us even more chips in our consumer electronics boxes. Or it might come from a neurotic obsession with tidiness, a bureaucratic notion that everything has to be identified with a type number and if it can't be understood by a computer it has no right to exist. Perhaps I go too far. But I really do object to the implication that you should be able to tick off in advance all the programmes you can expect yourself to like, according to specification, and then set a machine to pre-select them and automatically tune them in to trickle into your head like manufactured parts onto a conveyor belt. If it's not an insult to intelligence it will certainly put the mockers on spontaneity. One has a vision of the tape machines inexorably rolling at the transmitting end and the digital electronics ticking away obediently at the receiving end in superb synchrony. Do the people really matter?

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AC125	0.20	AV110	1.70*	BC184	0.11*	BD181	1.10	BF598	0.20*	MJE521	0.52	OC28	0.90	R2019	1.75*	IN917	0.06	2N2218	0.24	2N3904	0.13*
AC126	0.20	BA145	1.3*	BC212	0.15*	BD182	1.18	BFW10	0.65	MJE521	0.52	OC28	0.90	R2020	1.75*	IN918	0.06	2N2219	0.24	2N3905	0.13*
AC127	0.20	BA148	1.3*	BC213	0.15*	BD231	0.40	BFW11	0.65	MJE2955	1.25	OC29	2.00	R2021	1.75*	IN919	0.06	2N2220	0.18	2N3906	0.14*
AC128	0.20	BA154	0.09	BC214	0.15*	BD238	0.55	BFX24	0.82	MJE3855	0.75	OC30	1.50	R2022	1.75*	IN920	0.06	2N2221	0.18	2N4058	0.13*
AC141	0.25	BA155	0.10	BC237	0.05*	BDX10	0.91	BFX85	0.21	MPS102	0.30*	OC31	1.50	R2023	1.75*	IN921	0.06	2N2222	0.18	2N4059	0.13*
AC141K	0.35	BA156	0.09	BC238	0.12*	BDX32	0.20	BFX87	0.21	MPF103	0.30*	OC41	0.90	R2024	1.75*	IN922	0.06	2N2223	0.18	2N4060	0.10*
AC142	0.20	BAW62	0.05	BC301	0.25	BDY20	1.25	BFX88	0.21	MPF104	0.30*	OC42	0.75	R2025	1.75*	IN923	0.06	2N2224	0.18	2N4061	0.12*
AC142K	0.30	BAX13	0.06	BC302	0.24	BDY80	1.50	BFX89	0.26	MPF105	0.30*	OC43	2.25	R2026	1.75*	IN924	0.06	2N2225	0.18	2N4062	0.12*
AC17	0.20	BC076	0.10	BC307	0.10*	BF15	0.25	BFY51	0.26	MPSA06	0.24*	OC44	6.65	R2027	1.75*	IN925	0.06	2N2226	0.18	2N4063	0.12*
AC187	0.20	BC107	0.12	BC308	0.10*	BF152	0.18	BFY52	0.26	MPSA06	0.24*	OC45	0.55	R2028	1.75*	IN926	0.06	2N2227	0.18	2N4064	0.12*
AC188	0.20	BC108	0.12	BC327	0.20*	BF153	0.20	BFY64	0.26	MPSU01	0.36*	OC71	0.55	R2029	1.75*	IN927	0.06	2N2228	0.18	2N4065	0.12*
ACY17	0.85	BC109	0.13	BC328	0.18*	BF154	0.17	BFY90	1.25	MPSU06	0.46*	OC72	0.55	R2030	1.75*	IN928	0.06	2N2229	0.18	2N4066	0.12*
ACY18	0.80	BC113	0.12*	BC329	0.18*	BF155	0.17	BSX19	0.21	MPSU55	0.46*	OC73	1.00	TP12P555	0.67	ZG301	1.00	2N2925	0.22*	2N5458	0.35*
ACY19	0.75	BC118	0.12*	BC338	0.17*	BF160	0.16	BSX20	0.20	NKT401	2.00	OC74	0.65	TP13P505	0.56	ZG302	1.00	2N2926	0.14*	2N5459	0.35*
ACY20	0.70	BC115	0.14*	BCY30	1.00	BF167	0.20	BSX21	0.20	NKT403	1.73	OC75	0.55	ZS140	0.25*	2N404	1.50	2N3854	0.50	2N5017	6.50
ACY21	0.75	BC116	0.15*	BCY31	1.00	BF173	0.20	BT106	1.25	NKT403	1.73	OC76	0.55	ZS140	0.25*	2N404	1.50	2N3854	0.50	2N5017	6.50
ACY39	1.50	BC117	0.17*	BCY32	1.00	BF177	0.24	BTY79/400R		NKT404	1.73	OC77	1.20	ZS170	0.21*	2N696	0.25	2N3055	0.70	2N5026	12.00
AD149	0.70	BC118	0.10	BCY33	0.90	BF178	0.24		3.19	OA5	0.55	OC81	0.65	ZS178	0.54*	2N697	0.25	2N3440	0.60	2N5102	1.50
AD181	0.45	BC119	0.10	BCY34	0.90	BF179	0.24	BU205	1.75*	OA7	0.55	OC81Z	1.20	ZS271	0.23*	2N698	0.25	2N3440	0.60	2N5103	0.75
AD182	0.45	BC126	0.20*	BCY39	3.00	BF180	0.30	BU206	1.75*	OA7	0.55	OC82	0.65	ZS271	0.23*	2N698	0.25	2N3440	0.60	2N5103	0.75
AF106	0.45	BC135	0.14*	BCY40	1.00	BF181	0.30	BU208	2.00*	OA10	0.60	OC82	0.65	ZS271	0.23*	2N698	0.25	2N3440	0.60	2N5103	0.75
AF114	0.35	BC136	0.15*	BCY42	0.25	BF182	0.30	BY100	0.45	OA7	0.55	OC84	0.65	ZS271	0.23*	2N698	0.25	2N3440	0.60	2N5103	0.75
AF115	0.35	BC137	0.15*	BCY43	0.25	BF183	0.30	BY101	0.45	OA7	0.55	OC84	0.65	ZS271	0.23*	2N698	0.25	2N3440	0.60	2N5103	0.75
AF116	0.35	BC140	0.15*	BCY48	0.25	BF184	0.30	BY102	0.45	OA7	0.55	OC84	0.65	ZS271	0.23*	2N698	0.25	2N3440	0.60	2N5103	0.75
AF117	0.35	BC145	0.08*	BCY70	0.15	BF185	0.25	BY103	0.45	OA7	0.55	OC84	0.65	ZS271	0.23*	2N698	0.25	2N3440	0.60	2N5103	0.75
AF139	0.40	BC149	0.09*	BCY71	0.17	BF186	0.25	BY104	0.45	OA7	0.55	OC84	0.65	ZS271	0.23*	2N698	0.25	2N3440	0.60	2N5103	0.75
AF186	1.20	BC157	0.09*	BCY72	0.13	BF195	0.09*	Series		OA81	0.08	OC10	1.00	ZTX304	0.19*	2N1304	0.45	2N3707	0.13*		
AF239	0.45	BC158	0.08*	BCZ11	1.50	BF196	0.10*	Series		OA95	0.08	OC170	1.00	ZTX310	0.19*	2N1304	0.45	2N3707	0.13*		
AFZ21	2.75	BC167	0.12*	BD11	0.45	BF197	0.12*	CR51/05	0.60	OA95	0.08	OC170	1.00	ZTX310	0.19*	2N1304	0.45	2N3707	0.13*		
AS26	0.40	BC170	0.11*	BD121	1.20	BF200	0.27	CR51/05	0.60	OA95	0.08	OC170	1.00	ZTX310	0.19*	2N1304	0.45	2N3707	0.13*		
AS27	0.40	BC171	0.10*	BD123	1.20	BF224	0.20*	CR53/05	0.60	OA95	0.08	OC170	1.00	ZTX310	0.19*	2N1304	0.45	2N3707	0.13*		
				BD124	1.30	BF244	0.20*	CR53/40	0.75	OA202	0.08	OC200	1.50	ZTX310	0.19*	2N1304	0.45	2N3707	0.13*		
										OA202	0.08	OC200	1.50	ZTX310	0.19*	2N1304	0.45	2N3707	0.13*		
										OA202	0.08	OC200	1.50	ZTX310	0.19*	2N1304	0.45	2N3707	0.13*		
										OA202	0.08	OC200	1.50	ZTX310	0.19*	2N1304	0.45	2N3707	0.13*		
										OA202	0.08	OC200	1.50	ZTX310	0.19*	2N1304	0.45	2N3707	0.13*		
										OA202	0.08	OC200	1.50	ZTX310	0.19*	2N1304	0.45	2N3707	0.13*		
										OA202	0.08	OC200	1.50	ZTX310	0.19*	2N1304	0.45	2N3707	0.13*		
										OA202	0.08	OC200	1.50	ZTX310	0.19*	2N1304	0.45	2N3707	0.13*		
										OA202	0.08	OC200	1.50	ZTX310	0.19*	2N1304	0.45	2N3707	0.13*		
										OA202	0.08	OC200	1.50	ZTX310	0.19*	2N1304	0.45	2N3707	0.13*		
										OA202	0.08	OC200	1.50	ZTX310	0.19*	2N1304	0.45	2N3707	0.13*		
										OA202	0.08	OC200	1.50	ZTX310	0.19*	2N1304	0.45	2N3707	0.13*		
										OA202	0.08	OC200	1.50	ZTX310	0.19*	2N1304	0.45	2N3707	0.13*		
										OA202	0.08	OC200	1.50	ZTX310	0.19*	2N1304	0.45	2N3707	0.13*		
										OA202	0.08	OC200	1.50	ZTX310	0.19*	2N1304	0.45	2N3707	0.13*		
										OA202	0.08	OC200	1.50	ZTX310	0.19*	2N1304	0.45	2N3707	0.13*		
										OA202	0.08	OC200	1.50	ZTX310	0.19*	2N1304	0.45	2N3707	0.13*		
										OA202	0.08	OC200	1.50	ZTX310	0.19*	2N1304	0.45	2N3707	0.13*		
										OA202	0.08	OC200	1.50	ZTX310	0.19*	2N1304	0.45	2N3707	0.13*		
										OA202	0.08	OC200	1.50	ZTX310	0.19*	2N1304	0.45	2N3707	0.13*		
										OA202	0.08	OC200	1.50	ZTX310	0.19*	2N1304	0.45	2N3707	0.13*		
										OA202	0.08	OC200	1.50	ZTX310	0.19*	2N1304	0.45	2N3707	0.13*		
										OA202	0.08	OC200	1.50	ZTX310	0.19*	2N1304	0.45	2N3707	0.13*		
										OA202	0.08	OC200	1.50	ZTX310	0.19*	2N1304	0.45	2N3707	0.13*		
										OA202	0.08	OC200	1.50	ZTX310	0.19*	2N1304	0.45	2N3707	0.13*		
										OA202	0.08	OC200	1.50	ZTX310	0.19*	2N1304	0.45	2N3707	0.13*		
										OA202	0.08	OC200	1.50	ZTX310	0.19*	2N1304	0.45	2N3707	0.13*		
										OA202	0.08	OC200	1.50	ZTX310	0.19*	2N1304	0.45	2N3707	0.13*		
										OA202	0.08	OC200	1.50	ZTX310	0.19*	2N1304	0.45	2N3707	0.13*		
										OA202	0.08	OC200	1.50	ZTX310	0.19*	2N1304	0.45	2N3707	0.13*		
										OA202	0.08	OC200	1.50	ZTX310	0.19*	2N1304	0.45	2N3707	0.13*		
										OA202	0.08	OC200	1.50	ZTX310							

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As these circuits are capable of such an excellent performance we feel that it is not sensible to sacrifice this potential by designing a kit down to a price. We have therefore spent a little more on professional hardware allowing us to design a very advanced modular system. This enables a more satisfactory electrical layout to be achieved, particularly around the very critical input areas of the replay preamps. These are totally stable with this layout and require no extra stabilising components. Many other advantages also come from this system which has separate record and replay amps for each channel plugging in to a master board with gold-plated sockets. The most obvious is the reduction of crosstalk and interaction which could cause trouble on a single plane board, with our modular system the layout is compact but there is no component crowding. Testing is very easy with separate identical modules and building with the aid of our component-by-component instructions is childishly simple, but the finished result is a unit designed to normal domestic standards but to the best professional practice.

700C/2 High Quality Custom built steel Case. Complete with Brushed aluminium front plate, mains switch, record microswitch, turned record level knob, plastic cabinet feet, all bolts, nuts and mounting hardware. All necessary holes are punched and all surfaces are electroplated. Complete step-by-step assembly instructions are included. The cover is finished in an attractive black crackle surface. £16.50 + £2.06 VAT.

### LENCO CRV CASSETTE MECHANISM — Now fitted with Super Quality Sendust Alloy Head.

High Quality, robust cassette transport for Linsley-Hood recorder. Features fast forward, fast rewind, record, pause and full auto stop and cassette ejection facilities. Fitted with Record / play and erase heads and supplied complete with Data and extra cassette ejection spring for above horizontal use. Price £21.60 + £2.70 VAT.

Total cost of all parts £83.58.  
Special offer for Complete Kits £81.50 + £10.19 VAT.

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Set of components and data for optimising L-H Cassette circuits for use with this head, 50p + 6p VAT.

Standard Quality Stereo R/P Head, £4.50 + 56p VAT.

Economy Cassette Stereo R/P Head, £2.80 + 35p VAT.

4-track Cassette R/P Head, £7.40 + 93p VAT.

TEST CASSETTE to enable the user without instruments to easily set up the Head. Azimuth, tape speed and VU level, £1.50 inc. VAT.

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## TRANSFORMERS SAME-DAY DESPATCH

MAINS ISOLATOR				VAT 8% 12 or 24-VOLT			
Pri 120 or 240V Sec 120 or 240V				Separate 12V windings Pri 220-240V			
Centre Tapped and Screened				Ref	Amps	£	P&P
Ref.	VA (Watts)	£	P&P	Ref.	12v	24v	
07*	20	4.40	.79	111	0.5	0.25	2.20 .45
149	60	6.70	.96	213	1.0	0.5	2.64 .78
150	100	7.61	1.14	71	2	1	3.51 .78
151	200	11.16	1.14	18	4	2	4.03 .96
152	250	13.28	1.50	85	5	2.5	5.00 .79
153	350	16.43	1.84	70	6	3	5.35 .96
154	500	20.47	2.15	108	8	4	7.42 1.14
155	750	29.06	OA	72	10	5	8.12 1.14
156	1000	37.20	OA	116	12	6	8.99 1.32
157	1500	51.38	OA	17	16	8	10.72 1.32
158	2000	81.81	OA	115	20	10	13.98 2.08
159	3000	86.66	OA	187	30	15	17.05 2.08
*115 or 240 sec only. State volts required. Pri 0.220-240V.				226	60	30	36.14 OA

50 VOLT RANGE				30 VOLT RANGE			
Pri 220-240V. Sec. 0-20-25-33-40-50V.				Pri 220-240V			
Voltages available 5, 7, 8, 10, 13, 15, 17, 20, 25, 30, 33, 40 or 20V-0-20V and 25V-0-25V Screened				Sec. 0-12-15-20-24-30V			
				Voltages available 3, 4, 5, 6, 8, 9, 10, 12, 15, 18, 20, 24, 30V or 12V-0-12V and 15V-0-15V			
Ref.	Amps	£	P&P	Ref.	Amps	£	P&P
102	0.5	3.41	.78	112	0.5	2.64	.78
103	1.0	4.57	.96	79	1.0	3.57	.96
104	2.0	7.16	1.14	3	2.0	5.77	.96
105	3.0	8.56	1.32	20	3.0	6.20	1.14
106	4.0	15.06	1.50	21	4.0	7.99	1.14
107	6.0	14.62	1.64	51	5.0	9.87	1.32
118	8.0	20.26	2.08	117	6.0	11.17	1.45
119	10.0	24.98	OA	88	8.0	14.95	1.64
				89	10.0	17.25	1.84

60 VOLT RANGE				SCREENED MINIATURES Primary 240V			
Pri 220-240V				Ref.	mA	Volts	£ P&P
Sec. 0-24-30-40-48-60V. Voltages available 6, 8, 10, 12, 16, 18, 20, 24, 30, 36, 40, 48, 60V, or 24V-0-24V and 30V-0-30V				238	200	3-0-3	2.57 .55
Ref.	Amps	£	P&P	212	1A	0-6-0-6	2.85 .78
124	0.5	3.88	.96	1	100	9-0-9	2.14 .38
126	1.0	5.91	.96	235	330, 330	0-9-0-9	1.99 .38
127	2.0	7.60	1.14	207	500, 500	0-8-9, 0-8-9	2.77 .71
125	3.0	11.00	1.32	208	1A, 1A	0-8-9, 0-8-9	3.53 .78
123	4.0	12.52	1.84	236	200, 200	0-15, 0-15	1.99 .38
40	5.0	15.84	1.64	239	50MA	12-0-12	2.57 .38
120	6.0	18.06	1.84	214	300, 300	0-20, 2-20	2.80 .78
121	8.0	25.56	OA	221	700 (DC)	20-12-0-12-20	3.41 .78
122	10.0	29.55	OA	206	1A, 1A	0-15-20, 0-15-20	4.63 .96
189	12.0	34.06	OA	203	500, 500	0-15-27, 0-15-27	3.99 .96
				204	1A, 1A	0-15-27, 0-15-27	6.04 .96

HIGH VOLTAGE				AUTO TRANSFORMERS			
MAINS ISOLATING				Ref.	VA (Watts)	TAPS	£ P&P
Pri 200/220 or 400/440				113	15	0-115-210-240V	2.48 .71
Sec 100/120 or 200/240				64	75	0-115-210-240V	4.01 .96
VA	Ref.	£	P&P	4	150	0-115-200-220-240V	5.35 .96
60	243	6.70	1.32	66	300	"	7.75 1.14
350	247	16.43	1.84	67	500	"	10.99 1.64
1000	250	37.10	OA	84	1000	"	18.76 2.08
2000	252	61.81	OA	93	1500	"	23.28 OA
				95	2000	"	34.82 OA
				73	3000	"	59.21 OA
				80s	4000	0-10-115-200-220-240	76.86 OA
				57s	5000	"	89.50 OA

BRIDGE RECTIFIERS				CASED AUTO TRANSFORMERS			
100V 100A* £2.10				240V cable input USA 115V. Flat pin outlets.	P&P	Ref.	
200V 2A £4.5p				MINI-MULTIMETER			
400V 4A £5p				DC1000V, AC-1000V	15VA £4.95	90 113W	
400V 4A £8p				AC/DC-1000V/V	75VA £7.73	114 64W	
500V 12A* £2.35				DC-100mA, Res — 150K	150VA £10.01	114 4W	
*P&P 15p. VAT 12½% *VAT 8%				Bargain at £7.20	250VA £9.92	145 65W	
				VAT 8% P&P 62p	500VA £19.17	164 67W	
					750VA £23.41	176 83W	
					1000VA £27.88	230 84W	
					1500VA £26.02	OA 93W	
					2000VA £49.97	OA 95W	

TEST METERS				PANEL METERS			
AV08 Mk 5 £81.70				43mm x 43mm	82mm x 78mm		
AV07 £23.50				0-50µA	0-50µA	£6.70	
AV073 £45.20				0-500µA	0-500µA	£6.70	
AV08M5 MINOR £28.66				0-1mA	0-1mA	£6.70	
WEE MEGGER £66.90				0-30V	0-30V	£6.70	
TT169 (tests transistors in circuit) £34.75				VU Indicator Panel 90mm 250µA		£3.36	
U4315 budget meter (42 ranges) 20kΩ V/DC 1000V AC/DC (9 ranges) 2.5A AC/DC 500K resistance, in robust steel case with leads, full instructions. £15.85				VU Indicator Edge 54mm x 14mm µA FSD		£2.60	
EM272 316Kv £53.70				Carriage 65p VAT 8%			
DA116 Digital £102.00				Purbeck Oscilloscope Transformer. 250-0-250 at 80mA, 12.9-0-12.9 at 300mA, 6.3 at 600mA. £6.50. P&P 90p.			
Megger BM7 (Battery) £44.15							
MM5 £27.56							

MINIATURE TRANSFORMER				0.36-48 twice to give 72v or 92v.			
0-CT-15V							
Ref.	Amp	Price	P&P	2A	12.14	PP £1.40	4A 18.17 PP £1.90
171	500MA	2.09	.45	3A	14.70	PP £1.50	5A 26.64 PP £2.20
172	1A	2.96	.78				
173	2A	3.59	.78				
174	3A	3.75	.86				
175	4A	5.73	.96				
ABS PLASTIC BOXES				ELECTROSIL metal oxide low noise resistors.			
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PB3 120mm x 100 x 45 .87p				L	W	H	Price
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				B3	4	2¼	1½ 62p
				P&P29p. VAT 8% other sizes available			
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## 74 SERIES TTL IC's

Type	Price	Type	Price	Type	Price	Type	Price	Type	Price
7400	£0.07	7427	£0.21	7472	£0.19	74107	£0.22	74185	£0.25
7401	£0.09	7428	£0.25	7473	£0.22	74108	£0.25	74186	£0.25
7402	£0.09	7430	£0.06	7474	£0.22	74111	£0.25	74187	£0.25
7403	£0.09	7432	£0.20	7475	£0.27	74118	£0.25	74188	£0.25
7404	£0.09	7433	£0.28	7476	£0.22	74119	£1.10	74175	£0.20
7405	£0.09	7437	£0.20	7480	£0.40	74121	£0.22	74176	£0.25
7406	£0.22	7438	£0.20	7481	£0.20	74122	£0.35	74177	£0.25
7407	£0.12	7440	£0.16	7482	£0.65	74123	£0.30	74180	£0.20
7408	£0.12	7441	£0.45	7483	£0.55	74126	£0.30	74181	£1.25
7409	£0.12	7442	£0.38	7484	£0.82	74141	£0.50	74182	£0.55
7410	£0.09	7443	£0.08	7485	£0.65	74145	£0.54	74184	£1.00
7411	£0.15	7444	£0.08	7486	£0.22	74150	£0.85	74186	£0.60
7412	£0.22	7445	£0.64	7489	£1.00	74151	£0.45	74191	£0.60
7413	£0.22	7446	£0.60	7490	£0.30	74153	£0.45	74192	£0.65
7414	£0.05	7447	£0.45	7491	£0.60	74154	£0.80	74193	£0.60
7416	£0.22	7448	£0.52	7492	£0.32	74155	£0.44	74194	£0.55
7417	£0.27	7450	£0.09	7493	£0.29	74156	£0.44	74195	£0.55
7420	£0.09	7451	£0.09	7494	£0.70	74157	£0.44	74196	£0.50
7421	£0.09	7452	£0.09	7495	£0.45	74160	£0.55	74197	£0.50
7422	£0.35	7453	£0.09	7496	£0.48	74161	£0.80	74198	£1.00
7423	£0.20	7454	£0.09	74100	£0.80	74162	£0.80	74199	£1.00
7425	£0.18	7460	£0.09	74104	£0.35	74163	£0.60	74279	£1.00
7426	£0.21	7470	£0.24	74105	£0.35	74164	£0.65		

## CMOS IC's

Type	Price	Type	Price	Type	Price	Type	Price
CD4000	£0.12	CD4017	£0.65	CD4031	£1.60	CD4055	£1.00
CD4001	£0.13	CD4018	£0.70	CD4035	£0.90	CD4066	£1.15
CD4002	£0.13	CD4019	£0.35	CD4037	£0.78	CD4069	£0.15
CD4006	£0.30	CD4020	£0.80	CD4040	£0.78	CD4070	£0.15
CD4007	£0.14	CD4021	£0.75	CD4041	£0.68	CD4071	£0.15
CD4008	£0.30	CD4022	£0.75	CD4042	£0.68	CD4072	£0.15
CD4009	£0.30	CD4023	£0.13	CD4043	£0.78	CD4081	£0.15
CD4010	£0.32	CD4024	£0.65	CD4044	£0.78	CD4082	£0.16
CD4011	£0.13	CD4025	£0.13	CD4045	£1.15	CD4510	£0.80
CD4012	£0.14	CD4026	£1.00	CD4046	£0.95	CD4511	£0.80
CD4013	£0.35	CD4027	£0.45	CD4047	£0.75	CD4516	£0.85
CD4014	£0.70	CD4028	£0.60	CD4049	£0.35	CD4518	£0.85
CD4015	£0.70	CD4029	£0.40	CD4050	£0.35	CD4520	£0.85
CD4016	£0.35	CD4030	£0.55	CD4054	£0.95		

## BOOKS BY BABANI

The following books are offered at 10% off their normal retail price

		Normal Price	Sale Price
BP2	Handbook of Radio, TV & Industrial Transmitting Tubes & Valve Equivalents	£0.60	£0.54
BP6	Engineers & Mechanists Reference Tables	£0.40	£0.36
BP7	Radio & Electronic Colour Codes & Data	£0.15	£0.14
BP10	Modern Crystal & Transistor Set Circuits for Beginners	£0.95	£0.82
BP14	Second Book of Transistor Equivalents & Substitutes	£1.10	£0.99
BP15	Constructors Manual of Electronic Circuits for the Home	£0.50	£0.45
BP22	78 Electronic Novelty Circuits	£0.75	£0.68
BP23	First Book of Practical Electronic Projects	£0.75	£0.68
BP24	52 Projects Using IC's (for equivalents)	£0.75	£0.68
BP26	Radio Amateur Handbook for Long Distance Reception and Transmission	£0.85	£0.77
BP27	Giant Chart of Radio Electronic Semiconductor and Logic Symbols	£0.80	£0.54
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BP35	Handbook of IC Audio Pre-amplifier & Power Amplifier Construction	£1.10	£0.99
BP36	50 Circuits Using Germanium, Silicon & Zener Diodes	£0.75	£0.68
BP37	50 Projects Using Relays, SCR's & Triacs	£1.10	£0.99
BP39	50 (FET) Field Effect Transistor Projects	£1.25	£1.13
BP43	How to Make a Walkie-Talkie	£1.25	£1.13
BP44	IC 555 Timer Projects	£1.45	£1.31
BP47	Mobile Electronics Handbook	£1.35	£1.22
BP49	Popular Electronic Projects	£1.45	£1.31
BP60	Circuit Design and Construction Manual	£0.75	£0.68
BP66	AF-RF Resonance Frequency Chart for Constructors	£0.15	£0.14
BP202	Handbook of Integrated Circuits (ICs) Equivalents and Substitutes	£0.75	£0.68
BP205	First Book of Hi-Fi Loudspeaker Enclosures	£0.75	£0.68
BP213	Electronic Circuits for Model Railways	£0.85	£0.77
BP214	Audio Enthusiasts Handbook	£0.85	£0.77
BP215	Shortwave CIRCUITS & Gear for Experimenters and Radio Hams	£0.85	£0.77
BP216	Electronic Gadgets and Games	£0.85	£0.77
BP217	Solid State Power Supply Handbook	£0.85	£0.77
BP219	Solid State Novelty Projects	£0.85	£0.77
BP220	Build Your Own Solid State Hi-Fi and Audio Accessories	£0.85	£0.77
BP222	Solid State Short Wave Receivers for Beginners	£0.85	£0.86
BP223	50 Projects Using IC CA3130	£0.85	£0.86
BP224	50 CMOS IC Projects	£0.85	£0.86
BP225	A Practical Introduction to Digital IC's	£0.85	£0.86
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### VPS30 Variable Regulated Stabilised Power Supply Module

Incorporating a short circuit protection and current limiting:

Voltage Regulation	2-30v
Regulated Current	0-2A
AC Input Maximum	25v

Eliminates the use of batteries and thus saves £s — can be used time and time again. **ONLY £7.60 + V.A.T.**

**BRAND NEW** ITT 923 Silicon Diodes 200mA 200v.

100 off	£2.00
500 off	£9.00
1,000 off	£15.00
10,000 off	£130.00

**Single-sided Fibre-glass P.C. Board. S143.**  
12" x 3 1/2" approx. 2 pcs  
£0.60

## THYRISTORS

No. THY1A/50	1 Amp.	50 volt	TO3	18p
No. THY1A/400	1 Amp.	400 volt	TO3	32p
No. THY3A/50	3 amp.	50 volt	TO64	25p
No. THY3A/200	3 Amp.	200 volt	TO64	32p
No. THY3A/400	3 Amp.	400 volt	TO64	40p
No. THY5A/50	5 Amp.	50 volt	TO66	40p
No. THY5A/400	5 Amp.	400 volt	TO66	40p
No. THY5A/600	5 Amp.	600 volt	TO66	50p
No. C106/4	6 Amp.	400 volt	TO220	42p

S84	8 Amp. 400 volt	TO220 Plastic	80p
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ITT	V413 equit	12p
BR100	D32	12p

16201	18 Electrolytics	4.7µF — 10µF
16202	18 Electrolytics	10µF — 100µF
16203	18 Electrolytics	100µF — 680µF

All 3 at SPECIAL PRICE of £1.20\*

16160	24 Ceramic Caps	22pF — 82pF
16161	24 Ceramic Caps	100pF — 390pF
16162	24 Ceramic Caps	470pF — 3300
16168	24 Ceramic Caps	4700pF — 0.047µF

All 4 at SPECIAL PRICE of £1.60\*

Order No.	604W.	100 ohm — 820 ohm
16213	604W.	1K — 8.2K
16214	604W.	10K — 82K
16215	604W.	100K — 820K
16216	604W.	100K — 820K

All 4 at SPECIAL PRICE of £1.60

16217	404W.	100 ohm — 820 ohm
16218	404W.	1W — 8.2K
16219	404W.	1K — 8.2K
16220	404W.	100K — 820K

All 4 at SPECIAL PRICE of £1.60

16217	404W.	100 ohm — 820 ohm
16218	404W.	1W — 8.2K
16219	404W.	1K — 8.2K
16220	404W.	100K — 820K

All 4 at SPECIAL PRICE of £1.60

Order No.	38309K TO3	£1.20
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Positive	No. MVR7805	µA7805	TO220	55p
	No. MVR7812	µA7812	TO220	55p
	No. MVR7815	µA7815	TO220	55p
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µA723C TO99 38p 72723 14 pin DIL 38p

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No. 16178	5 x Mains Slide Switches	40p*
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	29 - Audio lead 5 pin plug to 5 pin DIN plug—Mirror Image	70p*
	130 - 5-metre lead 2 pin DIN plug to 2 pin DIN inline socket	45p*

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	130 - 5-metre lead 2 pin DIN plug to 2 pin DIN inline socket	45p*

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	130 - 5-metre lead 2 pin DIN plug to 2 pin DIN inline socket	45p*



**TRANSMITTER/RECEIVER MK.123.**

Very compact Army unit for use in range 2.5 to 20 Mc/s. Receiver section 7 valves inc. RF stage and BFO provides O/P for 4K ohm phones. 3 bands with direct calibration 2.5 to 5, 5 to 10 and 10 to 20 Mc/s. Tx section 2.5 to 20 Mc/s in 3 ranges O/P 15-25 watts over range C.W. only. 2 valves crystal osc and 5B/251M P.A. will match into the following A.C. amp. 25, 100, 500 and 1500 ohms as int. tune up meter, reqs. crystals type FT243 in range 2.5/10Mc/s. internal Morse key fitted with provision for ext. key. Mains P.U. self-contained suitable for 115 or 200/250V 50c/s overall size inc. Rx, Tx and P.U. 30x9x14 Cm weight 4Kg also supplied ext. inverter unit 12v DC. for mobile use. Supplied in clean condition with 80 page handbook giving circ details etc. (no details of 12v P.U.). Price £54.

**INDICATOR UNIT AN/ALA-6.**

Special purpose airborne indicator 115v 400c/s I/P as int. EHT supply and heaters reqs. ext. HT of 250 and -150 volts. 13 min. valves plus 3WP1 CRT 3" flat face with P.1 trace, this unit normally works with Rx and directional aerial system to give trace similar to CRDF indicator, requires ext. sine cosine resolver to generate scan, in clean condition in case size 10x8x21" with circ. Price £25.

**U.H.F. CAVITIES.**

Equipment spares for UPX-6 equipment for use with 2C39/7289 valves will tune over range 990/1040 Mc/s with int. fittings circ supplied. Price £6.50 also Rx section tunable preselector 1080/1130 Mc/s 4 section with 1N21 mixer diode for 60Mc/s I.F. new with circ. £4.50.

**ELECTROSTATIC VOLTMETERS.**

Range 0 to 15kV AC or DC usable scale 3 to 15kV complete in wood carrying case. size 8x9x6" tested. £10.80.

**INFRA RED LAMPS.**

Sealed beam units new spares. 115v 500 watts size 7" dia. 4 1/2" deep American G.E. okay for paint drying etc. Price 2 for £5.40.

**PHOTOMULTIPLIER TUBES.**

Dumont type 7065 end viewing S.11 response new boxed. Price £15.

**TAPE RECORDERS.**

Made for use in language lab equipment standard 240v I/P uses BSR type TD.10 3 speed deck 5" spools, two chan. trans. amps with separate O/Ps can be used for stereo provision for record and playback. P.U. and circ. boards are mounted below tape deck approx. overall size 12x11x7" intended to work phones. Supplied in clean condition may be less knobs and ind. lamps, some circ. details supplied. No ext. case. Price £13, also valve unit TD.2 deck. Price £8.50.

**MOTOR DRIVE AND CONTROLLER.**

Removed from Radar Simulators 24v DC P.M. motor driving into 2 stage gearbox the O/P shaft turns a 360° Ind also a sine cosine pot, approx. speed at 24v 1 RPM controller enables speed to be controlled from 0 to Max. in both directions, supplied with mains trans, rect, trans control amp and circ. Price £10.50.

**METER UNIT.**

Part of test set contains meter 5-0.5 Ua 3" dia. linear scale 5-0.5 contained in neat carrying case size 10x8x7 1/2" also contains pots swts. etc. price £8.50.

**APX-6 IFF TRANSPONDERS.**

Airborne unit for use in 1Km/s band contains tunable cavities for 2C42/6 series valves 60 Mc/s IF strip, blower, high voltage modulator parts, etc. 115v 400 c/s I/P note supplied less all valves ext. soiled cases with circ. Price £10.80.

**RECTIFIER UNIT.**

General purpose Army unit 200/250v I/P provides 2 DC O/Ps of 12v DC at 3 amps each can be connected for 12v 6 amp or 24v 3 amp will do 4 amps, in case size 17x8x7" with circ were used for 24v teleprinters. Price £10.80.

**A. H. SUPPLIES**

122 HANDSWORTH ROAD

SHEFFIELD, S94AE

Phone 444278 (0742)

**PULSE GENERATORS.**

Services type CT500 transistorised unit made by Wayne Kerr 200/250v I/P size 17 x 9 x 11" basic freq. 100 c/s to 12 kc with additional range to 60Kc with reduced facilities, pulse width. 5 to 7 Us, delay 5 to 1500 Us. O/P 100Uv to 10v neg. or pos. provision for ext. sync. or mod. 60Kc-range O/P var 5/20v. In good condition, regret no information. Price £22.

**E.H.T. UNITS.** Brandenburg EHT units 200/250v I/P O/P 12 to 15kV DC stab at 300 Ua. new condition with circ. £25.

**X.BAND ATTEN.** Variable atten W.G. 16 size about 40 Db approx. 6" long new. Price £4.50.

**HELIPOT DIALS.** Standard 10 turn type to fit 1/4" shaft, size 1 1/4" dia. Price £1.50, or with 100K helipot £2.

**DOVE BELLS.** American pattern chrome with red base, 11 1/2" dia. 6" deep, these are spring operated with solenoid control were used as supply failure warning solenoid control nom. 220v AC can be adapted for most AC or DC voltages, new unused American manufacture, chrome may be blemished due to storage. Price £25.

**AUTO TRANSFORMERS.** Nom taps 115/200/220/240v at 2 Kva C core new cond. few only £22.

**MISC ITEMS.** Dish Aerials approx. 7ft. dia. 15" deep, double skin ali. £54 ea., two only. Q band Tx Rx ass. K band Rx special with trav wave tube, etc. Digital Recorder H.P. type 562A with print out 11 digit with electronics £65. Marconi OA1094 Spec Analyzers 3/30 Mc/s with handbook £65. SKYLARK equipment telemetry Tx, Magnetometers, Flux Gates, 3 phase Inv. P.U. motor swts, etc.

**IMAGE INTENSIFIER.** Medical 9" with optics, transis TV system, 19" monitor, cables, etc. Price £750 nett.

Above prices include carriage and V.A.T. Goods ex. equipment unless stated new. see for list 21 or enquiry. Shop open Tuesdays through Saturdays.

# 15—240 Watts!

## HY5

Preamplifier

The HY5 is a mono hybrid amplifier ideally suited for all applications. All common input functions (mag Cartridge, tuner, etc.) are catered for internally, the desired function is achieved either by a multi-way switch or direct connection to the appropriate pins. The internal volume and tone circuits merely require connecting to external potentiometers (not included). The HY5 is compatible with all I.L.P. power amplifiers and power supplies. To ease construction and mounting a P.C. connector is supplied with each pre-amplifier.

**FEATURES:** Complete pre-amplifier in single pack — Multi-function equalization — Low noise — Low distortion — High overload — two simply combined for stereo.

**APPLICATIONS:** Hi-Fi — Mixers — Disco — Guitar and Organ — Public address.

**SPECIFICATIONS:**

**INPUTS:** Magnetic Pick-up 3mV, Ceramic Pick-up 30mV, Tuner 100mV, Microphone 10mV, Auxiliary 3-100mV, input impedance 47kΩ at 1kHz.

**OUTPUTS:** Tape 100mV, Main output 500mV R.M.S.

**ACTIVE TONE CONTROLS:** Treble ± 12dB at 10kHz, Bass ± at 100Hz.

**DISTORTION:** 0.1% at 1kHz, Signal/Noise Ratio 68dB.

**OVERLOAD:** 38dB on Magnetic Pick-up, SUPPLY VOLTAGE ± 16.50V

Price £6.27 + 78p VAT. P&P free.

HY5 mounting board B1 48p + 6p VAT P&P free.

## HY30

15 Watts into 8Ω

The HY30 is an exciting New kit from I.L.P., it features a virtually indestructible I.C. with short circuit and thermal protection. The kit consists of I.C., heatsink, P.C. board, 4 resistors, 6 capacitors, mounting kit, together with easy to follow construction and operating instructions. This amplifier is ideally suited to the beginner in audio who wishes to use the most up-to-date technology available.

**FEATURES:** Complete kit — Low Distortion — Short, Open and Thermal Protection — Easy to Build.

**APPLICATIONS:** Updating audio equipment — Guitar practice amplifier — Test amplifier — Audio oscillator.

**SPECIFICATIONS:**

**OUTPUT POWER:** 15W R.M.S. into 8Ω, **DISTORTION:** 0.1% at 15W.

**INPUT SENSITIVITY:** 500mV, **FREQUENCY RESPONSE:** 10Hz-16kHz — 3dB.

**SUPPLY VOLTAGE:** ± 18V.

Price £6.27 + 78p VAT. P&P free.

## HY50

25 Watts into 8Ω

The HY50 leads I.L.P.'s total integration approach to power amplifier design. The amplifier features an integral heatsink together with the simplicity of no external components. During the past three years the amplifier has been refined to the extent that it must be one of the most reliable and robust High Fidelity modules in the World.

**FEATURES:** Low Distortion — Integral Heatsink — Only five connections — 7 Amp output transistors — No external components.

**APPLICATIONS:** Medium Power Hi-Fi systems — Low power disco — Guitar amplifier.

**SPECIFICATIONS:** INPUT SENSITIVITY 500mV.

**OUTPUT POWER:** 25W RMS in 8Ω LOAD IMPEDANCE 4-16Ω, **DISTORTION:** 0.04% at 25W at 1kHz.

**SIGNAL/NOISE RATIO:** 75dB, **FREQUENCY RESPONSE:** 10Hz-45kHz — 3dB.

**SUPPLY VOLTAGE:** ± 25V, SIZE 105.50x25mm.

Price £8.18 + £1.02 VAT. P&P free.

## HY120

60 Watts into 8Ω

The HY120 is the baby of I.L.P.'s new high power range, designed to meet the most exacting requirements including load line and thermal protection, this amplifier sets a new standard in modular design.

**FEATURES:** Very low distortion — Integral Heatsink — Load line protection — Thermal protection — Five connections — No external components.

**APPLICATIONS:** Hi-Fi — High quality disco — Public address — Monitor amplifier — Guitar and organ.

**SPECIFICATIONS:**

**INPUT SENSITIVITY:** 500mV

**OUTPUT POWER:** 60W RMS into 8Ω, **LOAD IMPEDANCE:** 4-16Ω, **DISTORTION:** 0.04% at 60W at 1 kHz.

**SIGNAL/NOISE RATIO:** 90dB, **FREQUENCY RESPONSE:** 10Hz-45kHz — 3dB, **SUPPLY VOLTAGE:** ± 35V.

Size: 114 x 50 x 85mm.

Price £19.01 + £1.52 VAT. P&P free.

## HY200

120 Watts into 8Ω

The HY200, now improved to give an output of 120 Watts, has been designed to stand the most rugged conditions, such as disco or group while still retaining true Hi-Fi performance.

**FEATURES:** Thermal shutdown — Very low distortion — Loadline protection — Integral Heatsink — No external components.

**APPLICATIONS:** Hi-Fi — Disco — Monitor — Power Slave — Industrial — Public address.

**SPECIFICATIONS:**

**INPUT SENSITIVITY:** 500mV.

**OUTPUT POWER:** 120W RMS into 8Ω, **LOAD IMPEDANCE:** 4-16Ω, **DISTORTION:** 0.05% at 100W at 1kHz.

**SIGNAL/NOISE RATIO:** 96dB, **FREQUENCY RESPONSE:** 10Hz-45kHz — 3dB, **SUPPLY VOLTAGE:** ± 45V.

Size 114 x 100 x 85mm.

Price £27.99 + £2.24 VAT. P&P free.

## HY400

240 Watts into 4Ω

The HY400 is I.L.P.'s "Big Daddy" of the range producing 240W into 4Ω! It has been designed for high power disco or public address applications. If the amplifier is to be used at continuous high power levels a cooling fan is recommended. The amplifier includes all the qualities of the rest of the family to lead the market as a true high power hi-fidelity power module.

**FEATURES:** Thermal shutdown — Very low distortion — Load line protection — No external components.

**APPLICATIONS:** Public address — Disco — Power slave — Industrial.

**SPECIFICATIONS:**

**OUTPUT POWER:** 240W RMS into 4Ω, **LOAD IMPEDANCE:** 4-16Ω, **DISTORTION:** 0.1% at 240W at 1 kHz.

**SIGNAL/NOISE RATIO:** 94dB, **FREQUENCY RESPONSE:** 10Hz-45kHz — 3dB, **SUPPLY VOLTAGE:** ± 45V.

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820, 1000, 1200, 1500, 1800

2200, 2700, 3300, 3900, 4700, 5600

6800, 8200, 10000, 12000, 15000

22000, 27000, 33000, 39000, 47000, 56000

68000, 82000, 100000, 120000, 150000

220000, 270000, 330000, 390000, 470000, 560000

680000, 820000, 1000000, 1200000, 1500000

2200000, 2700000, 3300000, 3900000, 4700000, 5600000

6800000, 8200000, 10000000, 12000000, 15000000

22000000, 27000000, 33000000, 39000000, 47000000, 56000000

68000000, 82000000, 100000000, 120000000, 150000000

220000000, 270000000, 330000000, 390000000, 470000000, 560000000

680000000, 820000000, 1000000000, 1200000000, 1500000000

2200000000, 2700000000, 3300000000, 3900000000, 4700000000, 5600000000

6800000000, 8200000000, 10000000000, 12000000000, 15000000000

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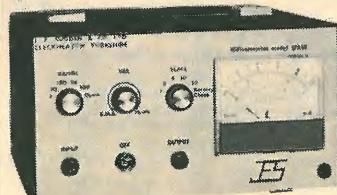
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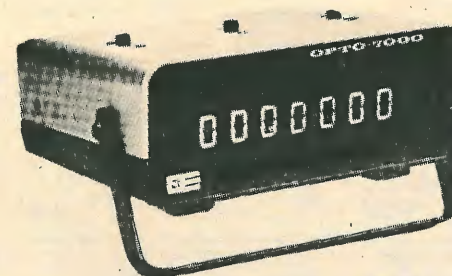
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**Frequency Range:** 10 Hz to 60 MHz (65 MHz Typical)  
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(10 MHz to 700 MHz Typical)

**Input Impedance:** 1 megohm shunted by 20 pF (60 MHz input)  
50 ohm (600 MHz input)

**Input Protection:** 1 megohm/60 MHz input - 100V up to 10 MHz  
50V up to 60 MHz  
50 ohm/600 MHz input - 2V max.

**Gate Times:** 100 milliseconds (1/10 second)  
1 second

**Resolution:** \* 1 Hz (10 Hz to 6 MHz) with switch (S4) Option  
10 Hz (10 Hz to 60 MHz)  
100 Hz (10 MHz to 600 MHz)

**Sensitivity:** <10 mV to 60 MHz  
25 mV to 150 MHz  
50 mV 450 MHz typ.  
(<75 mV Guaranteed)

**Time Base:** Quartz Crystal, 5.24288 MHz, TCXO, first order linear compensation

**Counter Accuracy:** ±1 count, Temperature stability and aging

**Temp. Stability:** .08PPM/°C (± 1 PPM 20° to .40°C, Typ.)

**Aging:** <2 PPM/year

**Display:** 7, .4" Red LED Digits

**Decimal Point:** Auto Placement

**Connectors:** BNC type.

**Power Requirement:** 1.5 Watts  
7.5 - 15V AC/DC <250 ma

**Batteries:** \* 4 - AA Ni-Cad, Constant Current Charger

**Size:** 1-3/4"H x 4-1/4"W x 5-1/4"D

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\*Optional - not included with basic unit

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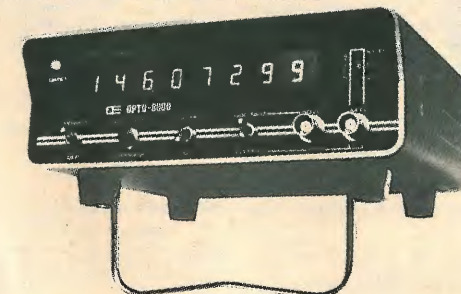


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### SPECIFICATIONS

**Frequency Range:** 10 Hz to 600 MHz in two over-lapping ranges.  
1 Megohm input - 10 Hz to 60 MHz  
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**Gate Times:** 100 milliseconds (1/10 second)  
(Switch Selectable) 1 second

**Resolution:** 1 Hz to 60 MHz, 10 Hz to 600 MHz

**Display:** 8 LED digits with floating decimal point, Flashing LED gate period indicator

**TIME BASE**

**Frequency:** 5.242880 MHz

**Type:** TCXO (Temperature Compensated Crystal Oscillator)  
less than .1 ppm from 17° to 40° C  
less than 1.0 ppm from 0° to 60° C

**Aging:** less than 1.0 ppm first year. (less than 1.0 ppm first three years typical)  
less than 0.5 ppm per year thereafter

**Initial set accuracy:** less than 0.5 Hz at 25° C

**External adjustment range:** ± 10 ppm

**Impedance:** 1 Megohm shunted by 20pF, DC coupled (10 Hz to 60 MHz)  
50 Ohm, AC coupled (20 MHz to 600 MHz)

**Connectors:** BNC type

**Typical Sensitivity:** 2 - 10 MV RMS below 150 MHz  
10 - 50 MV RMS below 600 MHz

**Max Input Signal:** 1 Meg input - 100V peak to peak up to 10 MHz  
50V peak to peak up to 60 MHz  
50 Ohm input - 2V peak to peak max.

**Power:** 220-250V AC nominal, 50 Hz, 2 Watts  
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**Dimensions:** 3-1/4" H x 7-1/4" W x 6-3/4" D

**Weight:** Approximately 2.5 pounds

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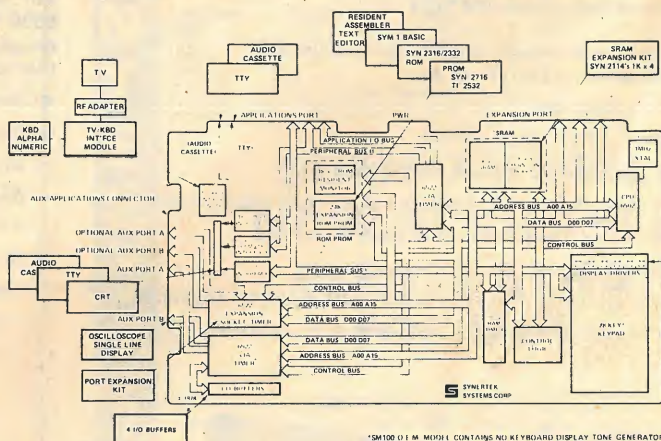


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EC52	0.40	KT66	4.00	UB01	0.80	6AQ5W	0.85	7V4	0.80	75C1	0.70
EC81	0.55	KT88	5.75	UBA5C80	0.60	6AS6	0.80	902	0.70	76	0.70
EC82	0.50	EL802	1.50	U424	0.75	6B6	0.75	902	0.70	76	0.70
EC83	1.15	ML6	1.00	UBF85	0.75	6AUB	0.70	10C2	0.60	80	0.80
EC84	0.45	OA2	0.55	UBF89	0.50	6AV6	0.50	10F18	0.60	85A2	0.70
EC85	0.50	OR2	0.60	UB11	1.00	6AXGT	0.80	10P13	0.60	85A2/B	0.70
EC86	1.25	PACB90	0.40	UBL21	0.75	6AXSGT	1.00	11E2	1.00	803	0.70
EC88	0.80	YB55	0.50	UCB04	0.80	6B6	0.75	12A6	0.60	725A	0.70
EC189	0.80	PC86	0.85	UCB85	0.50	6BA6	0.40	12A7E	0.65	807	0.70
ECF80	0.50	PC88	0.75	UCF80	0.80	6BE6	0.50	12A77	0.55	813	0.70
ECF82	0.45	PC900	1.25	UCH81	0.80	6BG6G	1.00	12AU7	0.50	829	0.70
ECF83	0.75	PC84	0.85	UCL82	0.75	6B16	1.10	12AV6	0.60	832A	0.70
ECF84	0.95	PC89	0.85	UF41	0.80	6C4	0.40	12E1	0.55	866A	0.70
ECF85	1.50	PC189	0.85	UF80	0.45	6BR7	2.30	12BA6	0.50	931A	0.70
ECF86	0.85	PCF80	0.80	UF85	0.55	6BW6	2.80	12BE6	0.60	95A	0.70
ECF87	0.45	PCF82	0.40	UL41	0.75	6BW7	1.00	12BM7	0.60	95B	0.70
ECF88	0.85	PCF84	0.85	UM84	0.80	6C4	0.40	12C4	0.55	956	0.70
ECF89	0.85	PCF86	0.85	UM80	0.60	6C6	0.55	12E1	0.45	957	0.70
ECF90	0.85	PCF200	0.90	UM84	0.40	6CH6	3.00	12J5GT	0.40	1625	0.70
ECF91	0.85	PCF801	0.90	UY82	0.55	6CL6	0.75	12KTGT	0.60	1625	0.70
ECF92	0.85	PCB802	0.85	UY85	0.50	6CY5	0.80	12K8GT	0.70	5183	0.70
ECF93	1.50	PCF805	1.80	VR105/30	1.80	6E48	0.80	12SC7	0.55	5842	0.70
ECF94	0.70	PCF806	0.85	VR105/30	1.80	6FB6G	0.90	12SH7	0.70	5933	0.70
ECF95	2.90	PCF808	1.80	Z66	1.25	6F8G	0.75	12SJ7	0.55	6057	0.70
ECF96	0.75	PC200	0.80	X66	0.80	6F12	0.85	12SQ7	0.55	6064	0.70
ECF97	0.80	PC181	0.80	X61M	1.80	6F14	0.80	12V4	0.40	6064	0.70
ECF98	1.50	PC182	0.65	Z800U	3.00	6F15	0.50	13D6	0.60	6065	0.70
ECF99	0.85	PC184	0.70	Z801U	3.50	6F17	1.00	14S7	1.00	6087	0.70
ECF100	0.65	PC186	0.70	Z900T	1.80	6F24	0.90	19AQ5	0.75	6080	0.70
ECF101	0.85	PC1805/8	0.85	11A	0.40	6G3	0.40	19C4	0.75	1046	0.70
ECF92	0.75	0.75	11A	0.30	6H6	0.85	19G6	0.50	6148B	0.70	
ECF95	0.55	P0500	3.25	11A	0.55	6J4WA	1.75	19H5	1.70	6380	0.70
ECF183	0.85	PL1200	1.35	15A	0.40	6J5	0.75	20D1	0.60	8020	0.70
ECF184	0.80	PL126	1.80	155	0.40	6JST	0.85	20F2	0.80		
ECF180	2.00	PL801	0.75	14A	0.40						
ECF200	0.75	PL82	0.50	14A	0.40						
EH90	0.80	PL83	0.50	1X28	1.10						
EL32	3.00	PL84	0.65	2X21	0.50						
EL24	2.20	PL84	0.65	2X25	0.50						
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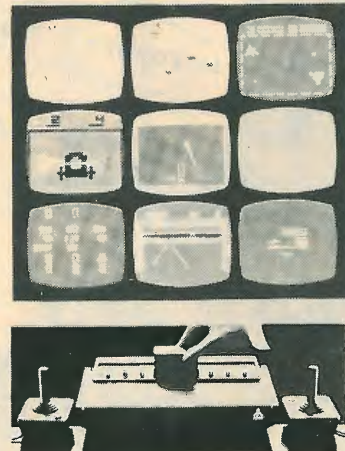
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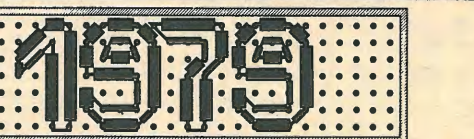
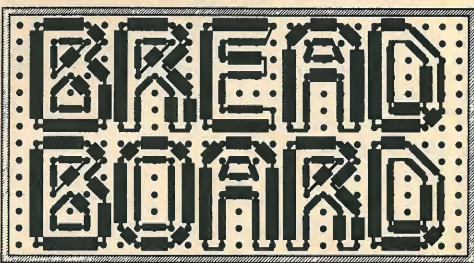
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11 pin plug in type		
12v AC	3 CO	85p
24v AC	3 CO	85p
24v DC	3 CO	85p
10v DC	3 CO	85p

### AC WKG BLOCK CAPACITORS BY FAMOUS MANUFACTURERS

MFD	Volts	Price
Q25	1500vAC	50p
Q75	440vAC	50p
1	470vAC	60p
1.25	360vAC	65p
2	400vAC	75p
2.4	360vAC	75p
2.5	360vAC	75p
2.7+0.1	700vAC	£1.25
3	440vAC	£1.00
3.5	250vAC	£1.00
4	250vAC	£1.00
5	360vAC	£1.25
6	440vAC	£1.50
7.2	440vAC	£1.50
8.4	250vAC	£1.75
15	250vAC	£1.75
PP up to 2.5 MFD 25p, 2.7 to 15MFD 50p+8% on total.		

### BLOCK PAPER CAPACITORS

6 MFD 300v wkg. AC. £1 pp 25p. 6 MFD 350v DC wkg. 75p pp 25p. 1 MFD 1,500v DC wkg. 60p pp 25p. 0.1 MFD 3,000v DC wkg. 50p pp 25p.

### BENSON'S SOLENOIDS





## NRDC-AMBISONIC UHJ



### SURROUND SOUND DECODER

The **first ever** kit specially produced by Integrex for this British NRDC backed surround sound system which is the result of 7 years' research by the Ambisonic team. W.W. July, Aug., '77.  
The unit is designed to decode not only UHJ but virtually all other 'quadrophonic' systems (Not CD4), including the new BBC HJ 10 input selections.  
The decoder is linear throughout and does not rely on listener fatiguing logic enhancement techniques. Both 2 or 3 input signals and 4 or 6 output signals are provided in this most versatile unit. Complete with mains power supply, wooden cabinet, panel, knobs, etc.

Complete kit, including licence fee **£49.50 + VAT**  
or ready built and tested **£67.50 + VAT**

## NEW S5050A STEREO AMP

50 watts rms-channel. 0.015% THD. S/N 90 dB, Mags/n 80 dB.

Tone cancel switch. 2 tape monitor switches.

Complete kit only **£63.90 + VAT**.

## Wireless World Dolby<sup>TM</sup> noise reducer

Trademark of Dolby Laboratories Inc.



Featuring:

- switching for both encoding (low-level h.f. compression) and decoding
- a switchable f.m. stereo multiplex and bias filter.
- provision for decoding Dolby f.m. radio transmissions (as in USA).
- no equipment needed for alignment.
- suitability for both open-reel and cassette tape machines.
- check tape switch for encoded monitoring in three-head machines.

### Typical performance

Noise reduction better than 9dB weighted.  
Clipping level 16.5dB above Dolby level (measured at 1% third harmonic content)

Harmonic distortion 0.1% at Dolby level typically 0.05% over most of band, rising to a maximum of 0.12%

Signal-to-noise ratio: 75dB (20Hz to 20kHz, signal at Dolby level) at Monitor output

Dynamic Range >90dB

30mV sensitivity.

Complete Kit **PRICE: £43.90 + VAT**

Also available ready built and tested

**Price £59.40 + VAT**

Calibration tapes are available for open-reel use and for cassette (specify which)

**Price £2.40 VAT\***

Single channel plug-in Dolby<sup>TM</sup> PROCESSOR BOARDS (92 x 87mm) with gold plated contacts are available with all components

**Price £9.00 + VAT**

Single channel board with selected fet

**Price £2.75 + VAT\***

Gold Plated edge connector

**Price £1.75 + VAT\***

Selected FETs **65p** each + VAT, **110p** + VAT for two, **£2.10** + VAT for four.

Please add VAT @ 12½% unless marked thus\*, when 8% applies (or current rates)

We guarantee full after-sales technical and servicing facilities on all our kits, have you checked that these services are available from other suppliers?



\*Please send SAE for complete lists and specifications

Portwood Industrial Estate, Church Gresley,  
Burton-on-Trent, Staffs DE11 9PT  
Burton-on-Trent (0283) 215432 Telex 377106

# INTEGREX LTD.

# INTEGREX

## S-2020TA STEREO TUNER/AMPLIFIER KIT

**SOLID MAHOGANY CABINET**

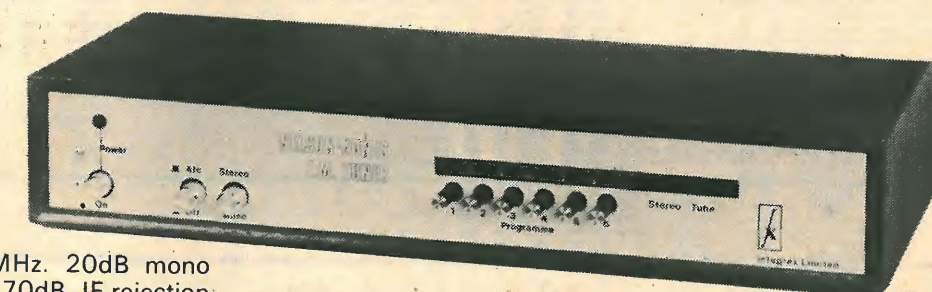
A high-quality push-button FM Varicap Stereo Tuner combined with a 24W r.m.s. per channel Stereo Amplifier.



**Brief Spec.** Amplifier Low field Toroidal transformer, Mag. input, Tape In/Out facility (for noise reduction unit, etc.), THD less than 0.1% at 20W into 8 ohms. Power on/off FET transient protection. All sockets, fuses, etc., are PC mounted for ease of assembly. Tuner section uses 3302 FET module requiring no RF alignment, ceramic IF, INTERSTATION MUTE, and phase-locked IC stereo decoder. LED tuning and stereo indicators. Tuning range 88—104MHz. 30dB mono S/N @ 1.2µV. THD 0.3%. Pre-decoder 'birdy' filter. **PRICE: £59.95 + VAT**  
Nelson-Jones Mk. 2 Stereo FM Tuner Kit. Price: **£69.95 + VAT**.

## NELSON-JONES MK. I STEREO FM TUNER KIT

A very high performance tuner with dual gate MOSFET RF and Mixer front end, triple gang varicap tuning, and dual ceramic filter/dual IC IF amp.



**Brief Spec.** Tuning range 88—104MHz. 20dB mono quieting @ 0.75µV. Image rejection — 70dB. IF rejection — 85dB. THD typically 0.4%.  
IC stabilized PSU and LED tuning indicators. Push-button tuning and AFC unit. Choice of either mono or stereo with a choice of stereo decoders.

Compare this spec. with tuners costing twice the price.



Sens. 30dB S/N mono @ 1.2µV  
THD typically 0.3%

Tuning range 88—104MHz  
LED sig. strength and stereo indicator

**Mono £36.40 + VAT**  
**With ICPL Decoder £40.67 + VAT**  
**With Portus-Haywood Decoder £44.20 + VAT**

## STEREO MODULE TUNER KIT

A low-cost Stereo Tuner based on the 3302 FET RF module requiring no alignment. The IF comprises a ceramic filter and high-performance IC Variable INTERSTATION MUTE. PLL stereo decoder IC. Pre-decoder 'birdy' filter. Push-button tuning

**PRICE: Stereo £33.95 + VAT**

## S-2020A AMPLIFIER KIT

Developed in our laboratories from the highly successful "TEXAN" design. PC mounting potentiometers, switches, sockets and fuses are used for ease of assembly and to minimize wiring. Power 'on/off' FET transient protection.



**Typ Spec.** 24+24W r.m.s. into 8-ohm load at less than 0.1% THD. Mag. PU input S/N 60dB. Radio input S/N 72dB. Headphone output. Tape In/Out facility (for noise reduction unit, etc.). Toroidal mains transformer.

**PRICE: £35.95 + VAT**

**BASIC NELSON-JONES TUNER KIT £15.70 + VAT**

**PHASE-LOCKED IC DECODER KIT £4.47 + VAT**

**BASIC MODULE TUNER KIT (stereo) £18.50 + VAT**

**PUSH-BUTTON UNIT £6.00 + VAT**

**PORTUS-HAYWOOD PHASE-LOCKED STEREO DECODER KIT £8.80 + VAT**







## ELECTRONIC KITS OF DISTINCTION FROM

## DE LUXE EASY TO BUILD LINSLEY-HOOD 75W AMPLIFIER



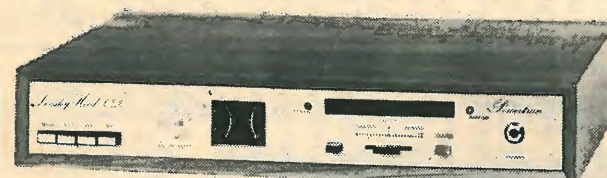
Pack	Price	Pack	Price
1. Fibreglass printed circuit board for power amp	£1.15	11. Fibreglass printed-circuit board for power supply	£0.85
2. Set of resistors, capacitors, pre-sets for power amp	£2.50	12. Set of resistors, capacitors, secondary fuses, semiconductors for power supply	£5.40
3. Set of semiconductors for power amp	£6.50	13. Set of miscellaneous parts including DIN skts., mains input skt., fuse holder, interconnecting cable, control knobs	£6.20
4. Pair of 2 drilled, lined heat sinks	£1.10	14. Set of metalwork parts including silk screen printed fascia panel and all brackets, fixing parts	£8.20
5. Fibreglass printed-circuit board for pre-amp	£1.90	15. Handbook	£0.30
6. Set of low noise resistors, capacitors, pre-sets for pre-amp	£4.10	16. High Quality Teak Veneer cabinet 18.3" x 12.7" x 3.1"	£10.70
7. Set of low noise, high gain semiconductors for pre-amp	£2.40		
8. Set of potentiometers (including mains switch)	£3.50		
9. Set of 4 push-button switches, rotary mode switch	£5.40		
10. Toroidal transformer complete with magnetic screen/ housing primary: 0 117-234 V; secondary: 33-0-33 V, 25-0-25 V	£12.95		

## PACK PRICES FOR STANDARD KIT

Designed in response to demand for a tuner to complement the world-wide acclaimed Linsley-Hood 75W Amplifier, this kit provides the perfect match. The Wireless World (Skingley and Thompson) published original circuit has been developed further for inclusion into this outstanding slimline unit and features a pre-aligned front end module, excellent a.m. rejection and temperature compensated varicap tuning, which may be controlled either continuously or by push-button pre-selection. Frequencies are indicated by a frequency meter and sliding LED indicators, attached to each channel selector pre-set. The PLL stereo decoder incorporates active filters for "birdy" suppression and power is supplied via a toroidal transformer and integrated regulator. For long term stability metal oxide resistors are used throughout.

AVAILABLE AS SEPARATE PACKS — PRICES IN OUR FREE CATALOGUE

## LINSLEY-HOOD CASSETTE DECK



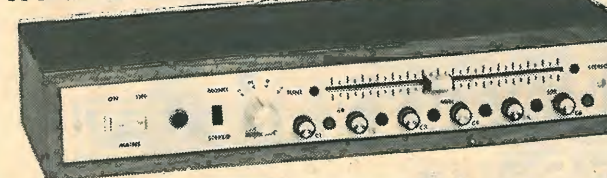
SPECIAL PRICE FOR COMPLETE KIT £79.60

Published in Wireless World (May, June, August 1976) by Mr. Linsley-Hood, this design, although straightforward and relatively low cost, nevertheless provides a very high standard of performance. To permit circuit optimization separate record and replay amplifiers are used, the latter using a discrete component front-end designed such that the noise level is below that of the tape background. Pushbutton switches are used to provide a choice of equalization time constants, a choice of bias levels and also an option of using an additional pre-amplifier for microphone use. The mechanism used is the Goldring-Lenco CRV, a unit distinguished in its robustness and ease of operation. Speed control and automatic cassette ejection are both implemented by electronic circuitry. This unit which is powered by a toroidal transformer and uses metal oxide resistors throughout offers an excellent match for the Wireless World Tuner and the Linsley-Hood 75 Watt Amplifier. Circuit changes as published in February, 1978, follow-up article are included in the kit AT NO EXTRA COST! A higher performance head (Matsushita WY 436 AZ head as recommended in the follow-up article) is offered as an optional extra but this will be automatically supplied FREE OF CHARGE with all orders for complete kits!

## T20+20 AND T30+30 20W, 30W AMPLIFIERS



## WWII TUNER



SPECIAL PRICE FOR COMPLETE KIT £47.70

AVAILABLE AS SEPARATE PACKS — PRICES IN OUR FREE CATALOGUE

Following the success of our Wireless World FM Tuner Kit this cost reduced model was designed to complement the T20+20 and T30+30 amplifiers and the cabinet size, front panel format and electrical characteristics make this tuner compatible with either.

AVAILABLE AS SEPARATE PACKS  
PRICES IN OUR FREE CATALOGUE

## SPECIAL PRICE FOR COMPLETE KIT

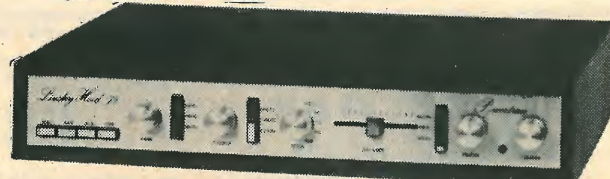
£99.30

The standard model of our kit for Mr. Linsley-Hood's 75 watt design has for a long time offered exceptional performance at a very modest cost with high quality high power ready-built units of comparable quality generally being over three times the price.

Features of the amplifier include very low distortion (less than 0.01%), 75W rms per channel power output, rumble filter, variable slope scratch filter, variable transition frequency tone controls, tape monitoring facilities and individually adjustable inputs. This model is based on 5 circuit boards which not having the controls mounted on them can, if desired, be effectively used separately in high performance audio systems not based on our metalwork.

Our new De Luxe model uses 14 boards which interconnect with gold plated contacts and are designed to have the potentiometers and switches mounted upon them. This system almost eliminates internal wiring, making installation, after their assembly, delightfully straightforward, and as each board can be easily removed in seconds from the chassis, checking and maintenance is so simple that even newcomers to electronics will be able to cope competently with the kit. Additional features of our new model are inclusion of the latest circuit improvements, generously sized heat sinks for heavy duty use, even in tropical climates, and metal oxide resistors throughout for long-term stability and reliability.

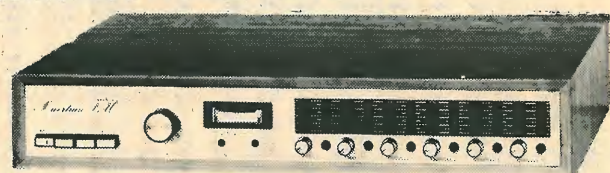
## STANDARD LINSLEY-HOOD 75W AMPLIFIER



SPECIAL PRICE FOR COMPLETE KIT

£79.80

## WIRELESS WORLD FM TUNER



SPECIAL PRICE FOR COMPLETE KIT

£70.20

Pack	Price	Pack	Price
1. Stereo PCB (accommodates 2 rep. amps, 2 meter, amps, bias/erase osc. relay)	£3.35	10. Set of capacitors, rectifiers, I.C. voltage regulator P.C.B. for power supply (Powertran design)	£2.80
2. Stereo set of capacitors, M.D. resistors, potentiometers for above	£7.95	11. Set of miscellaneous parts, including sockets, fuse holder, fuses, interconnecting wire, etc.	£3.40
3. Stereo set of semiconductors for above	£8.50	12. Set of metalwork including silk screened fascia panel, internal screen, fixing parts, etc.	£7.10
4. Miniature relay with socket	£2.90	13. Construction notes	£0.25
5. PCB, all components for solenoid, speed control circuits	£3.80	14. High Quality Teak Veneer cabinet 18.3" x 12.7" x 3.1"	£10.70
6. Goldring-Lenco mechanism as specified	£18.50		
7. Function switch, knob	£1.90		
8. Dual VU meter with illuminating lamp	£6.95		
9. Toroidal transformer with E.S. screen prim. 0-117V, 234V, Sec. 15V	£4.90		

One each of packs 1-14 inclusive are required for complete stereo cassette deck. Total cost of individually purchased packs £83.00

Matsushita WY 436 AZ head (optional extra) £4.50  
(free with complete kit)

Designed by Texas engineers and described in Practical Wireless, the Texan was an immediate success. Now developed further in our laboratories to include a Toroidal transformer and additional improvements, the slimline T20+20 delivers 20W rms per channel of true Hi-Fi at exceptionally low cost. The easy to build design is based on a single F/Glass PCB and features all the normal facilities found on quality amplifiers including scratch and rumble filters, adaptable input selector and headphones socket. In a follow-up article in Practical Wireless further modifications were suggested and these have been incorporated into the T30+30. These include RF interference filters and a tape monitor facility. Power output of this model is 30W rms per channel.

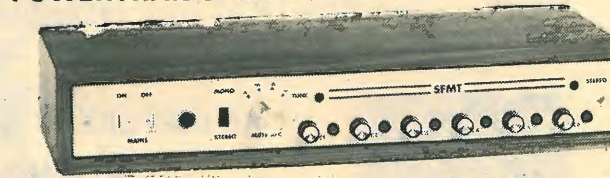
## SPECIAL PRICES FOR COMPLETE KITS

T20+20 KIT PRICE £33.10

T30+30 KIT PRICE £38.40

AVAILABLE AS SEPARATE PACKS — PRICES IN OUR FREE CATALOGUE

## POWERTRAN SFMT TUNER



PRICE FOR COMPLETE KIT £35.90

AVAILABLE AS COMPLETE KIT ONLY

This is a simple, low cost design which can be constructed easily without special alignment equipment but which still gives a first-class output suitable for feeding any of our very popular amplifiers or any other high quality audio equipment. A phase-locked-loop is used for stereo decoding and controls include switchable a.f.c., switchable muting and push-button channel selection (adjustable by controls on the front panel). This unit matches well with the T20+20 and T30+30 amplifiers.

## 200+200 watt AMPLIFIER

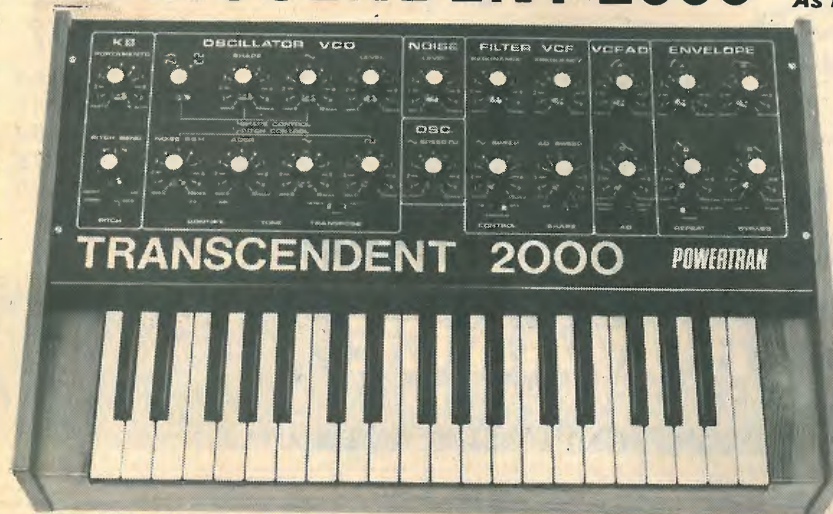
As featured in Electronics Today International

400W rms continuous — 800W peak!  
0.03% THD at FULL power!  
PLUS all the following features too!

- \* Each channel totally independent with its own stabilised power supply driven by custom designed TOROIDAL transformers!
- \* Inherent reliability — monster heat sinks for cool running at the hottest venues — electronic open and short circuit protection!
- \* Ultra low feedback (an incredible low 14dB overall!), super high slewing rate (20V/μs), 200W rms continuous to 4 ohm from EACH channel, input sensitivity 0.775V (0dB).
- \* Professional quality components, sturdy 19" rack mounting chassis complete with sleeve and feet for free standing work too.
- \* Easy to build — plenty of working space with ready access to all components, minimal wiring, extensive instruction suitable for both experienced constructors and newcomers to electronics.
- \* Value for money — quality and performance comparable with ready-built amplifiers costing over £600!

## TRANSCENDENT 2000 SINGLE BOARD SYNTHESIZER

As featured in Electronics Today International



The kit includes fully finished metalwork, fully assembled solid teak cabinet, filter sweep pedal, professional quality components (all resistors either 2% metal oxide or ½% metal film!) and it really is complete — right down to the last nut and bolt and last piece of wire! There is even a 13A plug in the kit — you need buy absolutely no more parts before plugging in and making great music! Virtually all the components are on the one professional quality fibre glass PCB printed with component locations. All the controls are mounted directly on the main board, all connections to the board are made with connector plugs and construction is so simple it can be built easily in a few evenings by almost anyone capable of neat soldering! When finished you will possess a synthesizer comparable in performance and quality with ready built units selling for between £500 and £700!

COMPLETE KIT ONLY

£172.00 + VAT!

Comprehensive handbook supplied with all complete kits! This fully describes construction and tells you how to set up your synthesizer with nothing more than a multi-meter and a pair of ears!

## CHROMATHEQUE 5000 5-CHANNEL LIGHTING EFFECTS SYSTEM

This versatile system featured as a constructional article in ELECTRONICS TODAY INTERNATIONAL has 5 frequency channels with individual level controls on each channel. Control of the lights is comprehensive to say the least. You can run the unit as a straightforward sound-to-light or have it strobe all the lights at a speed dependent upon music level or front panel control setting or use the internal digital circuitry which produces some superb random and sequencing effects. Each channel handles up to 500W and as the kit is a single board design wiring is minimal and construction very straightforward.

Kit includes fully finished metalwork, fibreglass PCB, controls, wire, etc. — Complete right down to the last nut and bolt!

COMPLETE KIT ONLY  
£49.50 + VAT

## MPA200 100W MIXER/AMPLIFIER

READ ALL ABOUT IT IN  
FEB. 1979  
ELECTRONICS TODAY  
INTERNATIONAL!

Wireless World Designs: Full kits are not available for the projects below but PCBs and component sets are stocked. Further details of these and other packs are in our Free Catalogue.

30W Bailey Amplifier	£1.00	Linsley-Hood Low Distortion Oscillator	£1.65	E. F. Taylor Pre-Amplifier	£1.45
BAIL Pk. 1 F/Glass PCB	£1.00	LDO Pk. 1 Fibreglass PCB	£1.65	EFTP Pk. 1 Fibreglass PCB (stereo)	£1.45
BAIL Pk. 2 Resistors, Capacitors	£2.35	LDO Pk. 2 MO Resistors, capacitors	£2.80	EFTP Pk. 2 MO Res. caps. (stereo)	£3.20
BAIL Pk. 3 Semiconductors	£4.70	LDO Pk. 3 Semiconductors	£3.90	EFTP Pk. 3 Semiconductors (stereo)	£4.20

Details of Stuart Tape Recorder and SQ Quadraphonic Decoders are in FREE CATALOGUE

## EXPORT A SPECIALITY!

Our Export Department can readily despatch orders of any size to any country in the world. Some of the countries to which we sent kits last year are shown in this advertisement. To assist in estimating postal costs our catalogue gives the same as for U.K. customers but no Value Added Tax charged. Postage charged at actual cost plus 50p documentation and handling. Please send payment with order by Bank Draft, Postal Order, International Money Order or cheque drawn on an account in the U.K. Alternatively for orders over £500 we will accept Irrevocable Letter of Credit payable at sight in London.

Value Added Tax not included in prices  
UK Carriage FREE

PRICE STABILITY: Order with confidence! Irrespective of any price changes we will honour all prices in this advertisement until February 28th, 1979, if this month's advertisement is mentioned with your order. Errors and VAT rate changes excluded.

U.K. ORDERS: Subject to 12½% \* surcharge for VAT (i.e. add ½ to the price). No charge is made for carriage, or current rate if charged.

SECURICOR DELIVERY: For this optional service (U.K. mainland only) add £2.50 (VAT inclusive) per kit.

SALES COUNTER: If you prefer to collect your kit from the factory, call at Sales Counter (at rear of factory). Open 9 a.m.-4.30 p.m. Monday-Thursday.

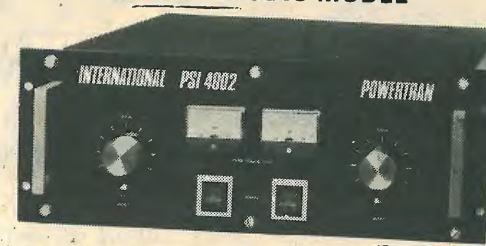
QUALITY: All components are brand new first grade full specification guaranteed devices. All resistors (except where stated as metal oxide) are low noise carbon film types. All printed circuit boards are fibreglass, drilled roller tinned and supplied with circuit diagrams and construction layouts.

FOR FURTHER INFORMATION PLEASE WRITE OR  
TELEPHONE FOR OUR FREE CATALOGUE

## POWERTRAN ELECTRONICS

PORTWAY INDUSTRIAL ESTATE  
ANDOVER HANTS SP10 3NN

ANDOVER  
(0264) 64455



Cabinet size 17.2" x 6.7"

COMPLETE KIT ONLY £196.90 + VAT

READ THE REVIEW  
IN SOUND INTERNATIONAL DEC 78!

## TRANSCENDENT 2000 SINGLE BOARD SYNTHESIZER

As featured in Electronics Today International

The kit includes fully finished metalwork, fully assembled solid teak cabinet, filter sweep pedal, professional quality components (all resistors either 2% metal oxide or ½% metal film!) and it really is complete — right down to the last nut and bolt and last piece of wire! There is even a 13A plug in the kit — you need buy absolutely no more parts before plugging in and making great music! Virtually all the components are on the one professional quality fibre glass PCB printed with component locations. All the controls are mounted directly on the main board, all connections to the board are made with connector plugs and construction is so simple it can be built easily in a few evenings by almost anyone capable of neat soldering! When finished you will possess a synthesizer comparable in performance and quality with ready built units selling for between £500 and £700!

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## POWERTRAN ELECTRONICS

PORTWAY INDUSTRIAL ESTATE  
ANDOVER HANTS SP10 3NN

ANDOVER  
(0264) 64455



# AUTOMATIC STANDBY INVERTERS

These units incorporate a transistorised inverter circuit, internal control relay that switches on a trickle 12v battery charger while the public power supply is maintained and automatically switches on the inverter restoring a 240v AC mains supply in the event of public power failure.

All units are assembled in strong durable instrument cases with ventilation slots, grade one new components are used throughout.

Features include: Five second charge to invert switchover period. Trickle charge rated 12v dc 500ma. SILICON POWER TRANSISTORS. DC Line Fused/AC supply fused. Neon-panel charge indicator. Inverter operating neon indicator, 13-amp type inverter outlet socket. Regulated 50Hz, smoothed and filtered inverter output waveform.

Installation is made simply by plugging in the AC lead to a normal domestic power socket, connect the two dc leads to a 12v car type battery and finally connect any electrical item you want to operate if there is a power cut into the unit's front panel inverter outlet socket.

ACE/1 — Cased assembly 6 1/2" x 4 1/2" x 4 1/2". 240v AC 50Hz 150 watts out.	£38.50
ACE/2 — Cased assembly 6 1/2" x 4 1/2" x 4 1/2". 240v AC 50Hz 200 watts out.	£46.73
ACE/3 — Cased assembly 8 1/2" x 5 1/2" x 5 1/2". 240v AC 50Hz 250 watts out.	£55.90
ACE/4 — Cased assembly 10 1/2" x 6 1/2" x 6 1/2". 240v AC 50Hz 300 watts out.	£70.87
ACE/5 — Cased assembly 10 1/2" x 6 1/2" x 6 1/2". 240v AC 50Hz 400 watts out.	£91.00
ACE/6 — Cased assembly 12 1/2" x 7 1/2" x 7 1/2". 240v AC 50Hz 450 watts out.	£108.75
ACE/7 — Cased assembly 16 1/2" x 9 1/2" x 12 1/2". 240v AC 50Hz 700 watts out.	£158.90

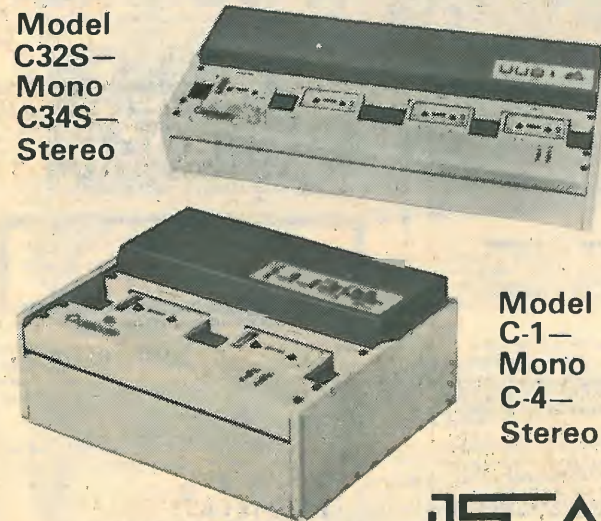
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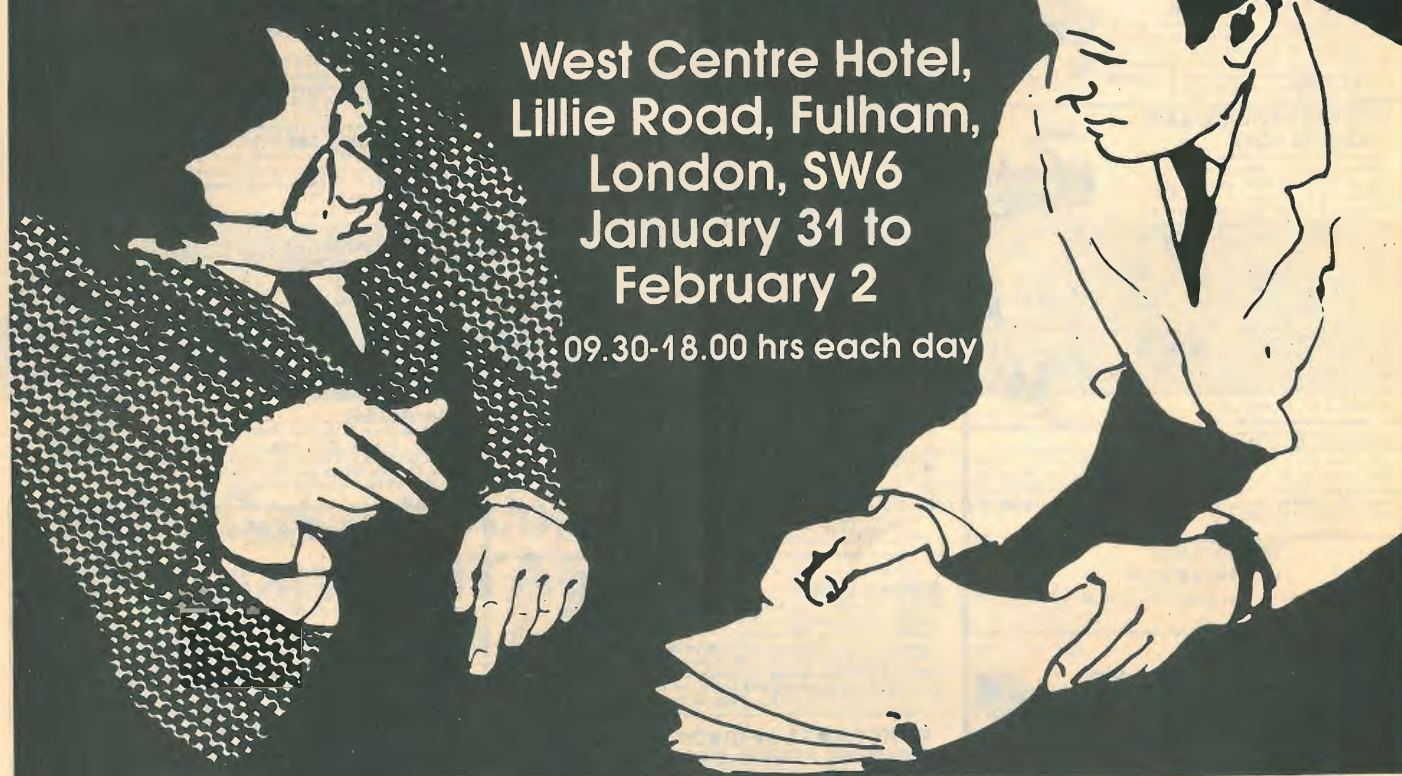


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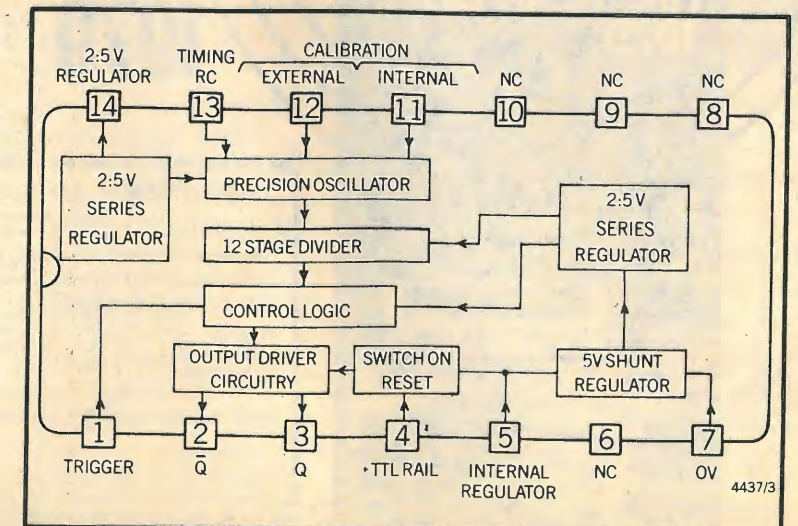
ZN1034 is a precision timer that can stand alone. It needs no special circuitry or special power supplies.

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And with a current consumption of only 5mA it can be used in battery powered equipment, where it gives an output drive capability of 25mA.

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7407	16p	74183	200p	4007	17p
7408	16p	74184	200p	4008	17p
7409	16p	74185	200p	4009	17p
7410	16p	74186	200p	4010	17p
7411	24p	74187	200p	4011	17p
7412	24p	74188	200p	4012	17p
7413	30p	74189	200p	4013	17p
7414	30p	74190	200p	4014	17p
7415	30p	74191	200p	4015	17p
7416	27p	74192	200p	4016	17p
7417	27p	74193	200p	4017	17p
7418	27p	74194	200p	4018	17p
7419	27p	74195	200p	4019	17p
7420	27p	74196	200p	4020	17p
7421	40p	74197	200p	4021	17p
7422	22p	74198	200p	4022	17p
7423	34p	74199	200p	4023	22p
7424	34p	74200	200p	4024	22p
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7426	34p	74202	200p	4026	22p
7427	34p	74203	200p	4027	22p
7428	34p	74204	200p	4028	22p
7429	34p	74205	200p	4029	22p
7430	34p	74206	200p	4030	22p
7431	34p	74207	200p	4031	22p
7432	34p	74208	200p	4032	22p
7433	34p	74209	200p	4033	22p
7434	34p	74210	200p	4034	22p
7435	34p	74211	200p	4035	22p
7436	34p	74212	200p	4036	22p
7437	34p	74213	200p	4037	22p
7438	34p	74214	200p	4038	22p
7439	34p	74215	200p	4039	22p
7440	34p	74216	200p	4040	22p
7441	34p	74217	200p	4041	22p
7442	34p	74218	200p	4042	22p
7443	34p	74219	200p	4043	22p
7444	34p	74220	200p	4044	22p
7445	34p	74221	200p	4045	22p
7446	34p	74222	200p	4046	22p
7447	34p	74223	200p	4047	22p
7448	34p	74224	200p	4048	22p
7449	34p	74225	200p	4049	22p
7450	34p	74226	200p	4050	22p
7451	34p	74227	200p	4051	22p
7452	34p	74228	200p	4052	22p
7453	34p	74229	200p	4053	22p
7454	34p	74230	200p	4054	22p
7455	34p	74231	200p	4055	22p
7456	34p	74232	200p	4056	22p
7457	34p	74233	200p	4057	22p
7458	34p	74234	200p	4058	22p
7459	34p	74235	200p	4059	22p
7460	34p	74236	200p	4060	22p
7461	34p	74237	200p	4061	22p
7462	34p	74238	200p	4062	22p
7463	34p	74239	200p	4063	22p
7464	34p	74240	200p	4064	22p
7465	34p	74241	200p	4065	22p
7466	34p	74242	200p	4066	22p
7467	34p	74243	200p	4067	22p
7468	34p	74244	200p	4068	22p
7469	34p	74245	200p	4069	22p
7470	34p	74246	200p	4070	22p
7471	34p	74247	200p	4071	22p
7472	34p	74248	200p	4072	22p
7473	34p	74249	200p	4073	22p
7474	34p	74250	200p	4074	22p
7475	34p	74251	200p	4075	22p
7476	34p	74252	200p	4076	22p
7477	34p	74253	200p	4077	22p
7478	34p	74254	200p	4078	22p
7479	34p	74255	200p	4079	22p
7480	34p	74256	200p	4080	22p
7481	34p	74257	200p	4081	22p
7482	34p	74258	200p	4082	22p
7483	34p	74259	200p	4083	22p
7484	34p	74260	200p	4084	22p
7485	34p	74261	200p	4085	22p
7486	34p	74262	200p	4086	22p
7487	34p	74263	200p	4087	22p
7488	34p	74264	200p	4088	22p
7489	34p	74265	200p	4089	22p
7490	34p	74266	200p	4090	22p
7491	34p	74267	200p	4091	22p
7492	34p	74268	200p	4092	22p
7493	34p	74269	200p	4093	22p
7494	34p	74270	200p	4094	22p
7495	34p	74271	200p	4095	22p
7496	34p	74272	200p	4096	22p
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7498	34p	74274	200p	4098	22p
7499	34p	74275	200p	4099	22p
7500	34p	74276	200p	4100	22p

TRANSISTORS			BFX30	34p	TIP35A	225p	2N3820	50p	DIODES			TRIACS	
AC126	25p	BFX45	30p	TIP35C	290p	2N3823	70p	'BY127	12p	PLASTIC		3A 400V	60p
AC127/8	20p	BFX66/7	30p	TIP36A	270p	2N3866	90p	'0A47	9p			3A 500V	65p
AC176	25p	BFX88	30p	TIP36C	340p	2N3903/4	18p	'0A81	15p			6A 400V	75p
AC187/8	25p	BFW10	30p	TIP41A	65p	2N4036	65p	'0A85	15p			6A 500V	85p
AF167/8	30p	BFY50	22p	TIP41C	78p	2N4036	65p	'0A90	9p			6A 500V	85p
AD149	70p	BFY51/2	22p	TIP42A	70p	2N4058/9	12p	'0A91	9p			6A 500V	85p
AD161/2	45p	BFY56	33p	TIP42C	82p	2N4060	12p	'0A95	9p			6A 500V	85p
BC107/8	11p	BFY90	90p	TIP295	70p	2N4081/2	18p	'0A99	9p			12A 400V	100p
BC109	11p	BUL70	700p	TIP305S	45p	2N4234/4	12p	'0A202	10p			12A 500V	105p
BC117	20p	BRY39	45p	'TIS43	34p	2N4125/6	22p	'1N914	4p			16A 400V	110p
BC147/8	9p	BSX19/20	20p	'TIS93	30p	2N4289	20p	'1N916	4p			16A 500V	110p
BC149	10p	BU104	225p	'TIX108	12p	2N4401/3	27p	'1N9148	4p			12B00D	120p
BC157/8	10p	'BU105	190p	2N102	13p	2N4427	90p	1N4001/2	5p				
BC158	11p	'BU108	250p	2N108	15p	2N4871	60p	1N4037/4	6p				
BC169C	12p	BU109	225p	2N109	15p	2N5087	27p	1N4005	6p				
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BC177/8	17p	'BU208	200p	2N457A	250p	2N5127	27p	1N5013/3	14p				
BC179	18p	'BU406	145p	2N696	35p	2N5179	27p	1N5047/7	19p				
BC182/3	10p	MJ132	10p	2N697	25p	2N5191	85p	'1N5920	10p				
BC184	11p	MJ491	200p	2N698	45p	2N5194	90p			THYRISTORS			
BC187	30p	MJ2501	225p	2N706A	20p	2N5245	40p			1A 50V		70p	
BC212/3	11p	MJ2955	100p	2N708A	20p	2N5298	55p			1A 400V		65p	
BC214	12p	MJ3001	225p	2N818	45p	2N5401	50p			1A 600V		85p	
BC277/8	30p	MJE285	100p	2N930	18p	2N5457/8	40p			3A 400V		90p	
BC516/7	50p	MJE3055	70p	2N1131/2	20p	2N5459	40p			8A 600V		140p	
BC547/8	16p	'MPF102	45p	2N1613	25p	2N5460	40p			12A 400V		160p	
BC548C	16p	'MPF103/4	40p	2N1711	25p	2N5485	44p			16A 100V		160p	
BC549C	18p	'MPF105/6	40p	2N1912	60p	2N5627	45p			16A 400V		160p	
BC559C	16p	MPSA06	30p	2N1260	120p	2N6247	19p			16A 600V		220p	
BC598	16p	'MPSA06	30p	2N219A	20p	2N6254	130p			BT106		110p	
BC599C	16p	MPSA06	30p	2N222A	20p	2N6290	85p			C106D		45p	
BCY70	18p	'MPSA12	50p	2N2369A	16p	2N6292	85p			'MCR101		36p	
BD131/2	22p	'MPSA56	32p	2N2478	9p	2N6312	120p			2N3525		120p	
BD132	50p	'MPSA56	32p	2N2646	50p	3N140	100p			2N4454		140p	
BD135/6	54p	'MPSU56	78p	2N2904/5	25p	3N141	110p			2N5060		34p	
BD139	56p	OC28	130p	2N2906A	24p	3N201	110p			2N5064		40p	
BD140	60p	OC35	130p	2N2907A	30p	3N204	100p						
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BD246	200p	'R2010B	200p	2N3053	20p	40360	40p						
BF200	32p	TIP29A	40p	2N3054	65p	40361/2	45p						
BF244B	35p	TIP29C	55p	2N3055	48p	40364	120p						
BF256B	70p	TIP30A	48p	2N3442	140p	40408	70p						
BF257/8	32p	TIP30C	60p	2N3553	240p	40409	65p						
BF259	36p	TIP31A	58p	2N3565	30p	40410	65p						
'BFR39	30p	TIP31C	62p	2N3634/4	48p	40411	300p						
'BFR40	30p	TIP32A	68p	2N3702/3	12p	40594	97p						
'BFR41	30p	TIP32C	82p	2N3704/5	12p	40595	105p						
'BFR42	30p	TIP33A	90p	2N3707/8	14p	40603	58p						
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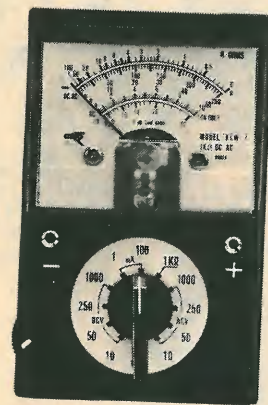


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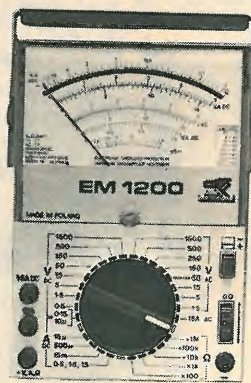
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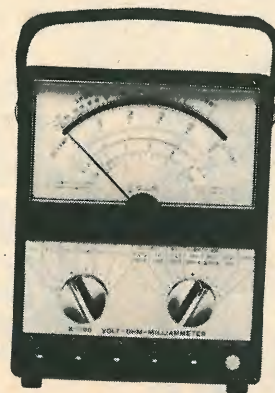
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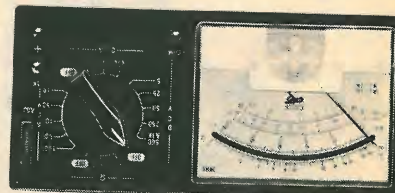
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**TMS3114 DUAL MOS 128 bit static shift reg.** OC-2.5mhz £1-75 each 4 for £5-00.  
**TMS 4050 4096 x 1 dynamic ram** £4-75 each 8 for £30-00.  
**NE555 10 for £2-40**  
**GE424 zero voltage switch, triac SCR.** relay driver TO5 can £1-00 each 7 for £5-00.  
**CA3011 20MHZ wideband amp** TO99 case 60p each 2 for £1-00.  
**FSA 2719 8 diodes IN4148-1N914** type in 16 D.I.L. pack all independently connected 35p each 4 for £1-00.  
**FPQ3725 4 NPN 50v 500ma silicon** transistors in 14 D.I.L. pack 65p each 2 for £1-00.  
**TEXAS LOW PROFILE I.C. SOCKETS**  
14 DIL 14p each 9 for £1-00  
22 DIL 25p each 5 for £1-00  
24 DIL 30p each 4 for £1-00  
**ZNS109 R.F. power output 400ma 2.5** watts up to 1200 mhz TO5 70p each 2 for £1-00.  
**AF279 low noise P.N.P. germanium** up to 780 mhz 30p each 4 for £1-00.  
**2N4304-WN720 F.E.T. transistor.** 35p each 4 for £1-00.

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**Scotch 215 Long Play Tape** Brand new "1" low noise on 1200' 5 1/2" spools £1-50 each 4 for £5-00 + P.P. 45p.  
**1000 Watt fully shrouded auto transformers** term. block connections £13-75 + P.P. £2-25.  
**Bulbs 24v 14 watt white frosted S.B.C. 8 for £1-00.**  
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**Heavy Duty Flat Insulated Earth Braid** 80-100 amp braided tinned copper in heavy clear P.V.C. sheath 40p per metre £5-00 for 15 metres P.P. 80p per metre.  
**BULGIN** miniature 6 way male chassis mount socket and matching shrouded free plug. 60p each 2 for £1-00.  
**Red L.E.D.'s** full spec. 0.2" 12p each 10 for £1-00.  
**Dynamic Stick Mikes 600 1/2"** with built in on/off switch complete with lead and min. jack plug £1-00 each.  
**TO5 HEATSINKS "Thermalloy"** black anodised press on aluminium finned type. 15p each 10 for £1-00.

**2N3001 30 volt 350 ma S.C.R. TO18** 20p each 6 for £1-00.  
**LM380N-SL6051 14 D.I.L. 2 watt** A.F. amp. 75p each 8 for £5-00.  
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Developer	30p	204mm x 114mm
Ferric Chloride	40p	204mm x 228mm
		408mm x 228mm
		467mm x 305mm

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Approx. 3.18mm thick sq. ft.	£1.25	£1.50
Approx. 2.00mm thick sq. ft.	£2.00	£2.25
Approx. 1.00mm thick sq. ft.	£1.50	£1.75

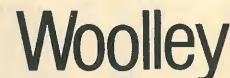
Single sided Copper-clad Paxolin, 10 sheets 245mm x 150mm £2.50  
Clear Acetate Sheet for making master, 260mm x 260mm 12p

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Small enough to be built inside the cabinets of many amplifiers.

Complete kit and board £30 including PSU & mains transformer. DESIGNER APPROVED

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The Stabilizer is a versatile frequency shifter for howl reduction on both speech and music. It offers variable shifts either up or down between 1 and 10 Hertz so allowing choice of the optimum shift for the particular acoustics and sound sources involved in each installation.

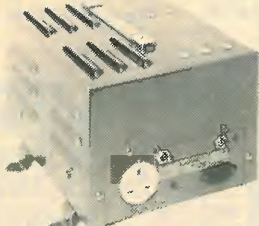
A shifter not only allows more usable gain (4-8dB) but also gives a greater stability margin between the onset of ringing and actual howling. With a shifter this is something between 3 and 5 dB, whereas a conventional system will go from 'ringing' to howling with a gain increase of 1 or 2 dB.

Available is a boxed unit with either balanced or unbalanced signal lines or rack mounting version offering studio quality 'SHIFT' control, duplicated jack and XLR connectors and a smart anodised finish with engraved front panel. Stabilizers include a signal overload LED, a 24 Hz high pass filter to remove VLF signals before connection to power amplifiers and a metal shrouded mains transformer to achieve low noise levels.

Stereo Disc Amplifier 2 \* 10 Outlet Distribution Amplifier 2 \* Peak Programme Meter movements, single and twin, and PPM2 IEC288-10A Drive Circuits \* Peak Deviation Meter \* Programme and Deviation Chart Recorders.

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**VM77E Transistorised Voltmeter** 10Hz-6MHz. 1mV-300V I/P Z 10M Ohms **£130.00**

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**91H RF Valve Voltmeter** 20KHz-1200MHz. 100V V-300V **£415.00**

### BRADLEY

**112RF Millivoltmeter** 3mV-300mV. 1-300MHz. Battery operated **£375.00**

### BRUEL & KJAER

**2409 Electronic Voltmeter** 2Hz-200KHz. 10mV-1000V True RMS **£245.00**

### HEWLETT PACKARD

**400F AC Voltmeter** 20Hz-4MHz 100V V-300V **£235.00**  
**411A RF Millivoltmeter** 500KHz-1GHz. 10mV-10V **£395.00**

### PHILIPS

**PM.2503 Electronic Multimeter** AC/DC Voltage and Current **£90.00**

### ROHDE & SCHWARZ

**U.R.V. RF Voltmeter** 1KHz-2400MHz with 50 Ohms insertion unit 20mV-10V **£235.00**

### SIGN

**AM324 AC Voltmeter** 15Hz-500KHz 1mV-300V **£60.00**

## Analogue Multimeters

### AVO

Model 7 **£40.00**  
Model 8 **£55.00**  
Test Set No. 1 **£65.00**  
Precision Avo **£45.00**  
Leads, Prods, clips for Avo's **£4.95**

### PHILIPS

**PM 2412 AC/DC Volts & amps** 40K Ohms/V **£60.00**

## Analysers

### GENERAL RADIO

**1911A Recording Sound Vibration Analyser** 2.5Hz-25kHz. 1/3 and 1/10 octave bandwidth. Recorder spec: 7Hz-200kHz. 1mVac sensitivity 20-40 or 80dB range. **£2,100.00**

### HEWLETT PACKARD

**310A Wave Analyser** 1kHz-2.5MHz. 10V V-100V voltage range. >70dB dynamic range. B.F.O. and recorder O/P. Has built-in AM, LSB and USB detector. **A SUPERB UNIT OF THE HIGHEST QUALITY** **£1200.00**  
**332A Distortion Analyser** 5Hz-600KHz. 0.1%-100% also AM detector **£495.00**  
**333A Distortion Analyser** 5Hz-600KHz. 0.1%-100% also Auto Null **£515.00**

### RACAL

**409 Modulation Meter** 3-1500MHz. AM & FM **£345.00**

### ROGERS/SIGN

**DM.344 Distortion Meter** 20Hz-20KHz **£230.00**

### TEKTRONIX

**1L10 Spectrum Analyser Plug In** 1-36MHz **£750.00**  
**1L30 Spectrum Analyser Plug In** 925MHz-10.5GHz **£1200.00**  
**1L40 Spectrum Analyser Plug In** 1.5-12.4GHz **£1275.00**

### T.E.S.

**D.566B Distortion Meter** 10Hz-1MHz. 0.03 to 100% **£305.00**

## MARCONI INSTRUMENTS

**T.F. 791D Deviation Meter** 4-1024MHz. Dev 10Hz/125KHz **£195.00**

### UREI

**Sonipulse 100A Acoustical Analyser** 40-16KHz. 27 x 1/3 octave bandpass filters. Microphone not included **£875.00**

## Attenuators

### MARCONI INSTRUMENTS

**TF.1073A Series Step Attenuator** DC-100MHz. 0-100dB 50 or 75 Ohms **£75.00**

**TF.2162 Step Attenuator** DC-1MHz. 0-111dB 600 Ohms. **£120.00**

### ROHDE & SCHWARZ

**RBD Attenuator** BN 33662/60 DC-600MHz. 20dB. 60 Ohms **£40.00**  
**DPR Step Attenuator** BN 18042/60 DC-300MHz. 0-100dB 60 Ohms **£90.00**

### ROHDE & SCHWARZ

**RBD UHF Attenuator** BN33661/50 DC-2.4GHz. 10dB. 50 Ohms **£55.00**

A copy of our trading conditions is available on request

**ELECTRONIC BROKERS LIMITED**

49-53 Pancras Road, London NW1 2QB

Tel. 01-837 7781. Telex: 298694

Hours of Business: 9 a.m.-5 p.m.  
Mon.-Fri.: closed lunch 1-2 p.m.

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WW—101 FOR FURTHER DETAILS

## Bridges

### MARCONI INSTRUMENTS

**TF.2701 In Situ Universal Bridge** **£395.00**

**TF.1245 'Q' Meter** Supplied with **TF.1246 Oscillator** 40KHz-50MHz **£625.00**

### WAYNE KERR

**B.221 (CT.530) Univ. Bridge** 0.1% accuracy **£275.00**

**Q.221 Low Impedance Adaptor** for use with B.221 **£75.00**

**B.521 (CT.375) Univ. Bridge** 1% accuracy **£120.00**

## Calibration Equipment

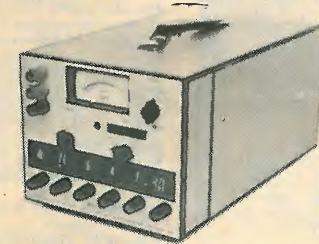
### HEWLETT PACKARD

**740B DC Voltage Source & Differential Voltmeter** **£850.00**

**741B DC Voltage Source & AC/DC Differential Voltmeter** **£975.00**

### FLUKE

**332A DC Voltage Calibrator** 0-1111.1110V in 3 ranges. 1ppm resolution. 0.003% calibration accuracy. O/P current 0-50mA **£1350.00**  
**931B True RMS Differential Voltmeter** 10Hz-1MHz. 0.01V-1100V **£1,050.00**  
**883AB. AC/DC Differential Voltmeter** 1mV-1100V **£975.00**



### TEKTRONIX

**184 Time Mark Generator** **£275.00**  
**2901 Time Mark Generator** **£400.00**  
**2101 5ns Pulse Generator** **£525.00**

**109 Pulse Generator.** Repetition rate 550-720Hz 250 ps risetime. Output 0-55V into 50 ohms **£320.00**

## Digital Counters

### ADVANCE

**TC.14 Frequency Counter** DC-250MHz. 9 digit **£295.00**

**TC.15 + TC.15P1 Counter & Plug In** DC-500MHz. 9 digit. 10mV sensitivity **£495.00**

**TC.17 or TC.1/A Time Counter** Freq. period. Period Average, Pulse width count DC-80MHz **£195.00**

**TC.21 Time Counter** Freq. Time, Period, Count Pulse width 2Hz-10MHz **£195.00**

Calibrated & Guaranteed 12 Months



**Calibrated & Guaranteed 12 Months**

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## The Test Equipment People

**Calibrated & Guaranteed 12 Months**



TC.22 Time Counter DC-100M+12 £275.00

### FLUKE

1900A opt. 01 5Hz-80MHz. 25mV Sensitivity with battery option £215.00  
1941A Industrial Counter/Totaliser 5Hz-40MHz £120.00

1900A Multifunction Counter. 6 digit LED display. Autoranging. 5Hz-80MHz. 25mV sensitivity £175.00  
1910A-01 Multifunction Counter. 7 digit LED display, autoranging. 5Hz-125MHz. 15mV sensitivity. Battery/Mains operated £285.00

1910A as 01 model but no battery opt. £189.00

1911A-01 Multifunction Counter. 7 Digit LED Display. Auto ranging. 5Hz-250MHz. 15-30mV sensitivity. Battery/mains operated £380.00

1912A Multifunction Counter. 7 Digit LED display. Auto ranging 5Hz-520MHz. 15-25mV sensitivity £390.00

1920A-06 Communications Counter 9 digit LED display. 5Hz-520MHz 15-25mV Sensitivity. Burst facility. A.F. resolution multiplier £490.00

1920A-13 Communications Counter As 1920A-06 but 5Hz-1GHz, and no A.F. resolution multiplier £525.00

1925A Multifunction Counter 9 digit LED display. 5Hz-125MHz. 15-25mV sensitivity. Adjustable trigger level. EMI Proof case £405.00

1953A-07 Universal Timer Counter 9 digit LED display DC-520MHz. 30mV sensitivity variable trigger level £675.00

1953A-15-16 Universal Timer Counter 9 digit LED display. DC-125MHz. 30mV sensitivity. Variable trigger level IEEE interface, rear inputs £850.00

1953A-15 as with opt. 16 but no rear input £825.00

### PHILIPS

PM.6612 Timer Counter 10Hz-80MHz 9 digit display £405.00  
PM.6615 Timer Counter 10Hz-1GHz. 10mV sensitivity £795.00

PM.6620 Timer Counter DC-45MHz. 50mV sensitivity £395.00

PM.6661 Frequency Counter 10Hz-80MHz. 20mV sensitivity. 8 digit £185.00

PM.6630A Timer Counter DC-160MHz. 8 digit Display £600.00

PM.6645 Frequency Counter 30Hz-512MHz. 5mV sensitivity £710.00

## Digital Voltmeters & Multimeters

### ADVANCE

DRM6 True R.M.S. DVM 4 1/2 digit, scale 1999. 10mV-1kV 10µV resolution. Frequency range DC-1MHz £295.00

### FLUKE

8020A DMM. 3 1/2 digit LED Display. DC volts and current. Resistance and diode test. Battery operated £99.00

8030A-01 DMM. 3 1/2 digit LED Display. DC Voltage and current. True rms AC Voltage and current. Resistance and diode test. Rechargeable battery supply £135.00

**Carriage and packing charge extra on all items unless otherwise stated**

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Tel: 01-837 7781

8040A-01 DMM 4 1/2 digit LED display. DC voltage and current. True rms AC Voltage and current. Resistance Rechargeable battery supply £220.00  
8600A-01 DMM 4 1/2 digit LED display. DC-AC voltage and current. Resistance. Mains or battery operated £335.00  
8600A as 01 model but no battery opt. £290.00

### HEWLETT PACKARD

3490A DM 5 1/2 digit, scale length 120000. AC Volts 1V-1kV 10µV resolution. DC Volts 100mV-1kV. 1µV resolution. Resistance 100 Ohms-10M Ohms. 1M Ohms resolution. Full auto-ranging and variable sample rate. Self check facility £595.00

34702A DMM C/W 34740A Display 4 1/2 digit. AC/DC & Ohms £295.00

### PHILIPS

PM2424 DMM 4 digit £300.00  
PM.2443 DC DVM 4 1/2 digit, scale length 19999 £430.00

PM2513 D.M.M. 3 1/2 digit £90.00  
PM.2513A DMM 3 1/2 digit, scale length 1999 £95.00

### S.E. LABORATORIES

SM210 DC DVM 4 digit, scale length 19999. 100mV-1kV. 10µV resolution £250.00

SM214 AC-DC DVM 5 1/2 digit, scale length 10999. SC-DC Volts 1.1V-1.1kV. 10µV resolution £300.00

### SCHLUMBERGER

A243 Digital Voltmeter 5 1/2 digit. 1µV resolution Autoranging DC & AC (mean) & Ohms £675.00

### WESTON

4449 3 1/2 digit D.M.M. AC-DC volts and current, resistance £79.50

## Miscellaneous

### E.N.I.

500L R.F. Amplifier 2-500MHz. 20dB gain. 300mW o/p £315.00

### FLUKE

412B H.V. Power Supply 0-2100V. Resolution 5mV o/p current 5-40mA £365.00

### BRUEL & KJAER

2607 Measuring Amplifier 2Hz-200kHz. 10µV-300V range. Average times 0.1S-300S. A, B, C and D weighting. Output signal 10V rms for F.S.D. on meter £1,200.00

## Oscilloscopes

### COSSOR

CDU 150 Dual Trace Oscilloscope DC-35MHz. 5mV-20V/div. Full delayed sweep. Long persistence CRT £450.00

4000 Dual Trace Oscilloscope DC-50 MHz. 5mV-10V/div. Full delayed sweep. Unused £495.00

### DYNAMCO

7100 Dual Trace Portable Oscilloscope with 1Y2 and 1X2 modules DC-30 MHz. 10mV-20V/div. Full delayed sweep £350.00

7500 Dual Trace Portable Oscilloscope DC-40MHz. 10mV-20V/div. Full delayed sweep. Unused £495.00

## HEWLETT PACKARD

1848 Storage Scope Rack style, variable persistence, c/w 1808A Dual Channel Vertical Amp. DC-75MHz. 1825 Time base and Delay Generator. UNUSED CONDITION - BARGAIN £1,800.00

1707B Dual Trace Portable Oscilloscope DC-75MHz. 10mV-5V/div. Full delayed sweep £895.00

422 Dual Trace Battery Portable Oscilloscope DC-15MHz. 10mV-20V/div. Full delayed sweep £750.00

422 Dual Trace Battery Portable Oscilloscope DC-15MHz. 10mV-20V/div. Full delayed sweep £750.00

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422 Dual Trace Battery Portable Oscilloscope DC-15MHz. 10mV-20V/div. Full delayed sweep £750.00

### PHILIPS

PM.3240 Dual Trace Portable Oscilloscope DC-50MHz. 5mV-2V/div. Full delayed sweep. From £950.00

PM3010 Miniature Scope DC-5MHz Dual Trace. Battery/Mains operation. Light weight 1.8 Kg £325.00

PM3230 Dual Beam Scope DC-10MHz. 20mV/div. TV frame and Sync separator £155.00

### SOLARTRON

CD 1400 Dual Trace Oscilloscope with 2 off CX 1441 and 1 off CX 1448 modules. DC-15MHz £155.00

### TEKTRONIX

531A Bench Oscilloscope with Dial trace vertical Plug-in unit CA. DC-13.5MHz. Sensitivity 50mV-20V/div. £290.00

647A Bench Oscilloscope with Dual trace vertical Plug-in unit 10A2A and delayed time base plug-in unit 11B2A DC-100MHz. Sensitivity 10mV-20V/div. £1,200.00

585A Bench Oscilloscope with Dual trace vertical Plug-in unit 82 DC-80MHz. Sensitivity 10mV-50V/div. £775.00

547 Bench Oscilloscope with dial trace vertical Plug-in unit 1A1 DC-50MHz. Sensitivity 5mV-20V/div. £775.00

545B Bench Oscilloscope with Dual trace vertical Plug-in unit CA DC-24MHz. Sensitivity 50mV to 20V/div. £425.00

432 Portable Scope Dual Trace DC-125MHz. 1mV/Div. SUPERB CONDITION. QUANTITIES AVAILABLE £495.00



661 Sampling Scope c/w 5T3 and 4S1 Dual Trace and accs. £585.00  
CA Plug In for 530, 540 and 580 series £60.00

T932 Dual Trace Portable oscilloscope DC-35MHz. 2mV-10V/div. Sweep speed 0.5S-10nS/div. With trigger hold off £550.00

7313 Split Screen Bistable Storage scope c/w 7A18 and 7B53A modules. DC-25MHz. 5mV-5V/div. Full delayed sweep. C.R.T. readout. 4.9cm/µS writing speed. Auto erase can be converted to 4 trace unit with addition of another 7A18 module. EXCELLENTLY PRICED AT ONLY £2450.00

7A22 Differential Plug In Unit. DC-1MHz 10µV-10V/div. Selectable upper and lower 10µV joints DC offset. 100,000:1 cm/m. SUPERB VALUE £450.00

7A26 Dual Trace Plug In Unit. DC-200MHz 5mV-5V/div £610.00

7D11 Digital Delay Plug In Unit. Delay 10µS. 0.1Hz-10MHz. TTL O/P will drive 30 gates £250.00

5715 Pulse Generator Similar spec. to 5712 but with variable rise and fall times £575.00

5438 Bench Oscilloscopes with Dual trace vertical Plug-in unit CA £350.00

575 Transistor Curve Tracer £400.00

556 Dual Beam Scope (Mainframe) DC-50MHz dependent on choice of Plug-ins £325.00

555 Dual Beam Scope (Mainframe) DC-33MHz wide choice of Plug-ins £300.00

422 Dual Trace Battery Portable Oscilloscope DC-15MHz. 10mV-20V/div. Full delayed sweep £750.00

422 Dual Trace Battery Portable Oscilloscope DC-15MHz. 10mV-20V/div. Full delayed sweep £750.00

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422 Dual Trace Battery Portable Oscilloscope DC-15MHz. 10mV-20V/div. Full delayed sweep £750.00

### ELEQUIPMENT

57A Dual Trace Portable Oscilloscope. DC-25MHz. 10mV-50V/div. Full delayed sweep £450.00

## Oscilloscope Probes

890 X1 Probe Kit DC-20MHz. 1.5 mtr cable 40pF I/P cap. 500V DC max. workable BRAND NEW £9.00

391 X10 Probe Kit. DC-80MHz. 1.2 mtr cable I/P Z 10M Ohms paralleled by 18pF. Compensation 15-50pF. BRAND NEW £11.00

395 X1 and X10 Switched Probe Kit. DC-15MHz in X1. DC-80MHz in X10. I/P Z 10M Ohms paralleled by 10.8pF in X10. 2mtr. cable. BRAND NEW £15.00

### GREENPAR

E81500/2 X1, X10 Probe Kit. DC-80MHz. 10M Ohms I/P resistance. Compensation 15-50pF. UNUSED £27.00

## Power Meters

### HEWLETT PACKARD

80C RF Power Meter with 477B Thermistor Mount 10MHz-10GHz. 100mW From £225.00

### MARCONI INSTRUMENTS

E893A AF Power Meter 20Hz-35kHz. 10W-10W I/P Z 2.5 Ohms-20K Ohms £185.00

F1020A Series RF Power Meter DC-50MHz. 50-100W FS or 150-300W FS. I/P Z 75 Ohms on 50-100W model, 50 Ohms on 150-300W model. From £105.00

### E.S.

U964 AF Power Meter 20Hz-50kHz. 10W-10W I/P Z 2.5 Ohms-20K Ohms £175.00

F1020A Series RF Power Meter DC-50MHz. 50-100W FS or 150-300W FS. I/P Z 75 Ohms on 50-100W model, 50 Ohms on 150-300W model. From £105.00

### E.S.

U964 AF Power Meter 20Hz-50kHz. 10W-10W I/P Z 2.5 Ohms-20K Ohms £175.00

### Pulse Generators

ADVANCE 0.5S-10nS/div. With trigger hold off £550.00

7313 Split Screen Bistable Storage scope c/w 7A18 and 7B53A modules. DC-25MHz. 5mV-5V/div. Full delayed sweep. C.R.T. readout. 4.9cm/µS writing speed. Auto erase can be converted to 4 trace unit with addition of another 7A18 module. EXCELLENTLY PRICED AT ONLY £2450.00

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422 Dual Trace Battery Portable Oscilloscope DC-15MHz. 10mV-20V/div. Full delayed sweep £750.00

## HOUSTON INSTRUMENTS

6520 YT Recorder £450.00

### PHILIPS

PM8110 Mini Single Channel Chart Recorder Sensitivity 10mV-10V full span. Chart width 12cm. Chart speed 5 and 20mm/min £250.00

### RECORD

3" Paper Width Recorder with 500 µA sensitivity FS. Left-hand zero. 1 and 6 per hour chart speed £75.00

## Radio Receivers

### EDDYSTONE

730/1A Communication Receiver 480kHz-30MHz in 5 ranges. BFO, noise limiting. AF filter, AVC, RF/gain. S Meter £175.00

730/4 Communications Receiver 480kHz-30MHz. 5 Bands. BFO, noise limiting. AVC, RF gain, AF filter. UNUSED CONDITION £275.00

880 Communications Receiver 500kHz-30.5MHz in 1MHz wavebands. BFO, AGC, RF/IF gain, noise limiting, AF filter, S Meter £325.00

### RAICAL

RA117E Communications Receiver 1-30MHz. MHz and KHz tuned separately. Selectivity 100Hz-13kHz in 6 ranges. BFO, AVC, Noise limiter, RF/IF gain, S meter £375.00

730/4 Communications Receiver 480kHz-30MHz. 5 Bands. BFO, noise limiting. AVC, RF gain, AF filter. UNUSED CONDITION £275.00

880 Communications Receiver 500kHz-30.5MHz in 1MHz wavebands. BFO, AGC, RF/IF gain, noise limiting, AF filter, S Meter £325.00

880 Communications Receiver 500kHz-30.5MHz in 1MHz wavebands. BFO, AGC, RF/IF gain, noise limiting, AF filter, S Meter £325.00

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880 Communications Receiver 500kHz-30.5MHz in 1MHz wavebands. BFO, AGC, RF/IF gain, noise limiting, AF filter, S Meter £325.00

880 Communications Receiver 500kHz-30.5MHz in



# 20Hz to 100MHz ~ £85.37

(U.K. only including VAT + P &amp; P.) Export + 10%



Actual size.

It's an incredible price for a very credible frequency counter... Continental's MAX-100.

It comes to you from a major American corporation and has one operating range, and one only: 20Hz to 100MHz, minimum. (Guaranteed.)

So we've pensioned off the range selector, and fitted the sharpest of LED displays. (Sheer brilliance.)

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But, most importantly, the MAX-100 is totally automatic — and available now. In fact, you could have one tomorrow.

Hesitating? Just take a look at the spec. Then, if you're ready to order immediately, call us on (0799) 21682. And

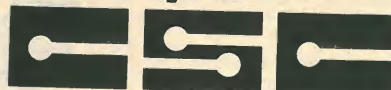
WW-104 FOR FURTHER DETAILS

your MAX-100 could be on its way, today! (Continental are great performers, too.)

For data, please use our enquiry number.

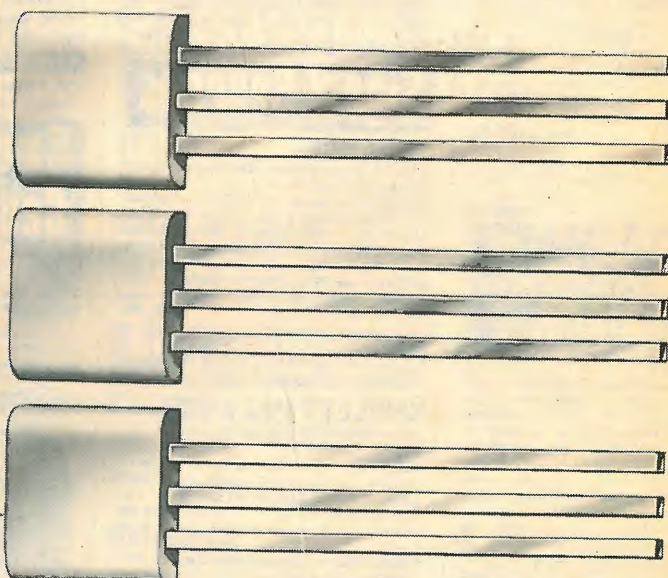
**Specification** \* Frequency range 20Hz to 100MHz \* Input impedance 1 megohm shunted by 56pF \* Sensitivity 30mV to 300mV r.m.s., from 20Hz to 100MHz \* Timebase accuracy 3ppm \* Temperature stability 0.2ppm per °C \* Max. ageing rate 10ppm per year \* Overfrequency indication \* Low battery power alarm \* Operates from a.c. mains, dry or rechargeable cells, or 12Vd.c. auto battery \* Dimensions 45 x 187 x 143 mm. \* Options: 12V auto. cigar lighter adaptor; battery eliminator/charger; r.f. antenna; low-loss r.f. tap; and carrying case.

## Quality. At a low, low, price.



Continental Specialties Corporation (U.K.) Ltd.  
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## Double your diodes with Ferranti and save on system costs.



If you're using glass diodes, here's a gentle warning. Pre-forming and insertion costs can make them more expensive than you think on the assembly line.

Ferranti ZDX double diodes save you all that. Each package contains two connected diodes so you save both on insertion cost and component counts.

The ZDX range includes both switching and general purpose diodes. Common anode, common cathode or series pair.

Ready to plug into your PCB — another saving. And they're competitively priced. Saving again.

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Discrete Components Marketing, Ferranti Electronics Limited,  
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## FERRANTI Semiconductors

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# Electronic Brokers

## The Computer People

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Programme boards for switching and interconnecting input/output circuits. 11 x 20 XY matrix. Interconnection is by means of shorting, skip and component holding pins (not supplied). Dimensions: 7 1/2" x 5 1/4" x 1"

**PRICE: £12.50**  
(mail order total £14.58)

#### NEW KEYTOP/KEYSWITCH KITS

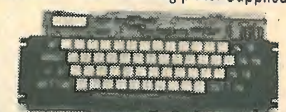
**BRAND NEW SURPLUS**  
Pack of 56 Keyswitches plus 56 keytops comprising standard ASCII character set. Ideal as basis for self-build keyboard projects

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#### SURPLUS ASCII KEYBOARDS BY CLARE-PENDAR

KB3 ROM-encoded ASCII keyboard with 63-push-button key stations. Selectable mode — either full ASCII or TTY. Selectable parity. TTL-compatible. Power requirements, +5V, 12V. Constructed on rugged PCB with metal mounting plate. Supplied with full technical data. Manufacturers surplus.

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**SCREEN CAPACITY** — 960 characters 80 per line x 12 lines.  
**CURSOR** — underline.

**CHARACTER GENERATION** — 5 x 7 dot matrix 625 line raster.

**CHARACTER SET** — 64 ASCII alphanumeric and symbols.

**CHARACTER SIZE** — 1/8 inch (.32cm) nominal height. 3/32 inch (.24cm) nominal width.

**TUBE PHOSPHOR** — P4 (white on black)

**REFRESH RATE** — 50 fields per second.

**KEYBOARDS** — TTY format attached.

**INDICATORS** — Power On, Parity Error, Dataset ready.

**PARITY** — Parity error indicated by Parity light and question mark (?) displayed in character position.

**TRANSMISSION** — Asynchronous. Switch-selectable for any two standard rates up to 9600 baud.

**OPERATING MODES** — Full/half Duplex.

**MEMORY** — High speed MOS refresh.

**STANDARD INTERFACE** — CCITT V-24, (EIA RS-232 B/C).

**REMOTE COMMANDS** — Home Cursor, Clear Screen, Carriage Return, Line Feed.

#### HAZELTINE H2000 SPECIFICATION

**SCREEN SIZE** — 12" diagonal. 1998 characters; 74 per line x 27 lines.

**CHARACTER GENERATION** — 5 x 7 dot matrix 625 line raster.

**CHARACTER SET** — 64 alphanumeric and symbols. 32 ASCII control codes.

**KEYBOARD** — Detachable, solid state, teletypewriter design. 10-key numeric cluster plus editing and cursor control keys.

**TRANSMISSION** — Asynchronous. Switch-selectable for combinations of 5 standard rates, 75 to 9600 baud.

**OPERATING MODES** — Switch-selectable, full duplex, half duplex or batch (buffered).

**MEMORY TYPE** — 2048 x 8 RAM.

**EDITING FEATURES** — Full Cursor controls plus Insert/Delete Character, Insert/Delete Line, Clear Screen, Clear Foreground Data Only, Tab.

**STANDARD INTERFACE** — CCITT V-24 (EIA RS-232 B/C) or 202C Compatible.

**REMOTE COMMANDS** — Insert/Delete Line, Clear Screen, Clear Foreground Data Only, Home Cursor, Address Cursor, Set Background Intensity, Set Foreground Intensity, Carriage Return, Backspace, Ring Bell, Transmit, Print.



H-2000A ONLY £495  
Also available:  
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**AUXILIARY OUTPUT** — Standard printer interfaces; standard cassette interface; remote monitor interface.  
**TUBE PHOSPHOR** — P39 (green on black).

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- \* ASCII Keyboard
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132 column printer with 64 ASCII character set, 165 cps operation.  
Model 101-5x7 dot matrix.  
Model 101A-7x9 dot matrix. £750.00

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NE556-14	0.82	7805-UC	0.81	8C207A	0.11
T84810	1.30	7812-UC	0.81	8C208	0.10
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72702	0.79				
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7410	0.12	74126	0.44	7470	0.28
7411	0.18	74132	0.67	7473	0.25
7413	0.25	74136	0.72	7474	0.25
7420	0.12	74141	0.58	7480	0.45
7421	0.27	74142	1.96	7484	0.88
7422	0.17	74160	0.78	74147	1.25
7427	0.24	74161	0.78	74148	1.15
7428	0.32	74162	0.78	74150	0.96
7430	0.13	74163	0.78	74151A	0.60
7440	0.13	74164	0.87	74153	0.60
7422A	0.92	74165	0.87	74154	1.03
7485	0.83	74173	1.16	74157	0.63
7490	0.33	74174	0.87	74159	1.65
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7494	0.75	74180	0.87	74198	1.45
7495	0.51	74184A	1.18	74393	1.94

WW—076 FOR FURTHER DETAILS

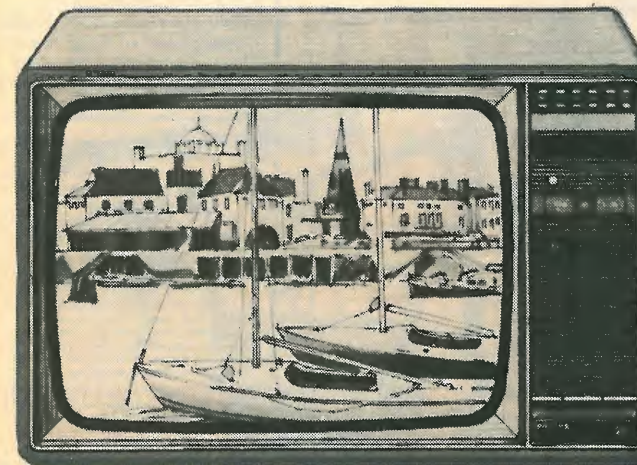
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## Test Systems Engineers

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Lowestoft: a pleasant seaside town of some character, with a busy fishing quarter and narrow lanes running down to the beach. Close to the extensive leisure and recreational facilities offered by Oulton Broad, and the Suffolk Wild Life and Country Park, with housing offering really excellent value for money.

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We're currently looking for additional men and women with a sound knowledge of television systems and measurement techniques to join a team involved in the design and construction of specialist test equipment for use within the production facility. The work is extremely varied, covering monochrome and colour receivers, teletext and viewdata systems, and remote control units for the UK and European markets.

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If you're qualified to degree level or equivalent and are looking for a new way of life, contact Richard McMullan, Staff Personnel Officer, Pye Ltd., Manufacturing Division, Oulton Broad, Lowestoft. (0502) 62222.

8923



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FOR VIDEO AND AV FACILITIES  
£3933-£4761 p.a. plus £312 p.a. salary supplement

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Further details and application forms from Personnel Officer, Brighton Polytechnic, Moulsecomb, Brighton BN2 4GJ. Tel: Brighton 693655 Ext. 2536. Closing date: 2nd February, 1979.

(8846)

## TECHNICAL WRITER

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(8879)



Opportunities exist for Engineers (male or female) to contribute to the planning and maintenance of the Independent Television and Local Radio Networks.

## Transmitter Engineer

Ref: WW/247CC

To join our specialist team responsible for the planning, installation and commissioning of television and radio transmitting stations. The work involves liaison with equipment manufacturers, installing equipment on site and preparation of technical specifications.

Applicants should be qualified to degree level (or equivalent) and have several years' experience either related to the manufacture of transmitter equipment or in the operations and maintenance of broadcasting equipment.

## Transmitter Engineer

Ref: WW/249CC

An experienced Broadcast Engineer familiar with radio and television transmission equipment to assist in the provision of a back-up service to our Station Engineers. The work involves the investigation of problem areas in the transmission equipment associated with our VHF and UHF television relay stations and VHF and MF radio stations, the maintenance of adequate documentation and spares; liaising with component suppliers; and assisting with the processing of modifications.

Applicants should be qualified to HNC/HND level in electrical/electronic engineering, or equivalent, and have had several years' experience in the maintenance, testing or commissioning of television transposing equipment and radio broadcast transmitters.

## Telemetry Engineer

Ref: WW/251CC

The Engineer will assist with the provision of a maintenance support service for computer based remote control and supervisory equipment used to monitor and control the IBA's network of transmitters. The work involves maintenance of operational software and the provision of test equipment, diagnostic procedures and spares for field maintenance staff.

Applicants should be qualified to HNC/HND level in electrical/electronic engineering, or equivalent, and have had several years' experience in the maintenance or installation of digital equipment.

All three posts will be based at the Authority's Engineering Headquarters in Hampshire, and since travelling throughout the U.K. is involved, a current driving licence is essential.

Starting salaries will be on a range which rises to £6,066 per annum (inclusive of a productivity supplement). Generous travel and subsistence allowances are payable and relocation expenses will be paid, where appropriate.

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BROADCASTING  
AUTHORITY

Please write with full details of experience, quoting the appropriate reference nos., or telephone for an application form to: Glynis Powell, IBA, Crawley Court, Winchester, Hampshire, SO21 2QA. Telephone: Winchester 822270.

### University of Leicester Department of Engineering **ELECTRONICS TECHNICIAN**

Applications are invited for a post of Electronics Technician (Grade 5) in the Electrical and Electronics Laboratories. The successful applicant will work as a member of a small group of technicians who provide a service to the Department that includes construction, installation, maintenance and repair of electrical and electronic equipment used for teaching and research. Applicants should have a sound knowledge of electronic circuits and experience in digital techniques. Work will include manufacture and calibration of specialised equipment for teaching and research, including microprocessor systems.

The post is pensionable. Salary scale (under review) from £3186 to £3720 (Grade 5).

Apply in writing to the **Head of Department of Engineering, The University, Leicester, LE1 7AH**, quoting reference 56/T. (8878)

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## TECHNICAL OFFICER

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in its Audio Visual Aids Unit

#### Duties:

Supervising the work of the electronic technicians in the Unit including maintenance and studio services. Advising on specifications and plans for audio visual systems for lecture theatres, laboratories, TV installations, etc., in the University.

#### Requirements:

1. A very good technical knowledge of all types of audio and video equipment.
2. Wide experience of maintaining and developing all types of CCTV and video recording equipment including colour equipment.
3. Experience in technical supervision of professional television/sound studio.

The following qualifications are essential:

- a) Experience — At least 7 years in a position of responsibility looking after technical staff.
- b) Education — A degree in Electronics or Electrical Engineering or HNC or HND (Electrical Engineering or Applied Physics).

#### Reward:

Salary in the range £4,631 by 6 increments to £6,080.

Further details:

If you would like to apply for this appointment please write to the Director, Cambridge University AVA Unit, Old Exam Hall, Free School Lane, Cambridge CB2 3RH.

Closing date for applications is 15th February, 1979.

(8848)

### LONDON COLLEGE OF PRINTING

ELEPHANT AND CASTLE  
LONDON SE1 6SB

#### DEPARTMENT OF PHOTOGRAPHY, FILM AND TELEVISION

## LECTURER GRADE I

Applications are invited for the post of lecturer in the School of Film and Television.

The post offers scope to a person with television engineering and some production experience to teach students on a three-year honours degree course in Visual Communications while being responsible for the television facilities in the department. Experience of installation and maintenance in the Broadcast or CCTV industries is an essential qualification.

The successful applicant would be expected to work closely with the Head of School on future planning and development of the technical facilities within an area where there are limitations both in resources and finance.

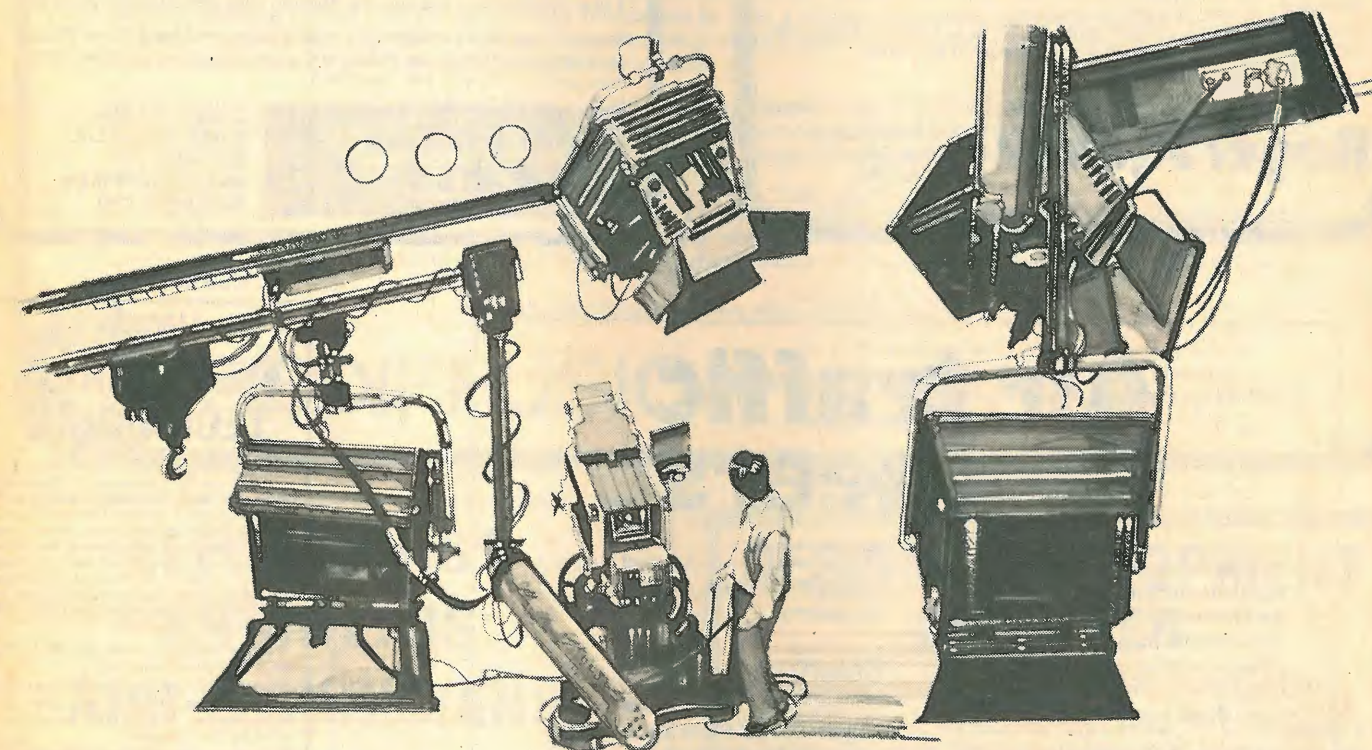
Salary scale in accordance with the Burnham (FE) Report: on an incremental scale within the range £3192-£5334 (plus £474 London Allowance, subject to formal approval) starting point depending upon qualifications, training and experience.

Application forms, returnable by 31 January, 1979 and further details, may be obtained from the Acting Senior Administrative Officer at the college. Telephone 735 8484, Ext. 227.

(8853)

## INSTALLATION TECHNICIANS

# Behind the scenes, behind the screens.



Behind the people who are behind the scenes at all the BBC's film, radio or television studios you will find the expertise, professional competence and innovative skill of the Studio Capital Projects Department. This department is responsible for the planning, installation and commissioning of entire sound and vision systems for all the BBC's film, radio and television studios and so is responsible for the outstanding technical reputation which we enjoy worldwide.

To help us maintain this excellent tradition, we are looking for Installation Technicians to assist our engineers in the preparation and requisitioning of schedules of equipment, on-site liaison, supervision, testing and commissioning of installations. The work is exciting and varied.

To succeed in this project work, you should be trained to ONC or equivalent level, with experience in testing

electronic equipment and workshop or office drawing. Further experience in the manufacture, operation or maintenance of broadcasting equipment in a relevant field would be an advantage.

Although you will be based in London, you must be prepared to work throughout the U.K. Salaries, ranging from £3775 to £4590 rising to £5130 (under review) will reflect your experience and qualifications and relocation expenses will be considered.

If you could fit in behind the screens, we would like to hear from you. Just call **George Boston** on 01-580 4468 ext 2426 or write to: **Engineering Recruitment Officer, BBC, Broadcasting House, London W1A 1AA** quoting reference no 78.E.4093/WW.

**BBC**

(8850)



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(8890)

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The Company is pleasantly situated on the Avon/Somerset border just a few minutes from the sea, the Mendip Hills and Junction 21 of M5.

Salary is negotiable according to qualifications and experience. Please write giving brief personal details and home telephone number, stating position in which you are interested.

BRISTOL AEROJET

Contact Ron Moir,  
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Banwell,  
Weston-Super-Mare,  
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(8872)

## Air traffic Engineers

The Civil Aviation Authority has vacancies for men and women as Air Traffic Engineers Grade 2 in its Telecommunications Division offering a variety of work on a wide range of electronic systems and specialised equipments.

Air Traffic Engineers Grade 2 are involved in the installation and maintenance of radio, radar, air navigational and landing aids, and data processing systems. Staff are employed at some Civil Airports, Air Traffic Control Centres and Radar Stations and other locations throughout the U.K. but at present most of the vacancies are likely to be in the South of England with a few vacancies at locations in Scotland and Shetlands.

### Qualifications and Experience

You should be at least 20 years of age and have obtained either the ONC(ENG) with an electronic bias or C & G Telecommunications Technician T3 Certificates or other similar technical qualifications. You

should also have had skilled working experience in radio, radar or data processing.

### Salary

Salaries are on an incremental scale £3890-£5763. Posts in the London area attract an additional allowance (Inner London £495-Outer London £293). Grade 1 posts (maximum salary £6957) are normally filled by promotion from Grade 2.

For full details and an application form complete and send the coupon to:—CAA Tels Staff, Management (ATE2), Room K206, CAA House, 45/59 Kingsway, London WC2B 6TE.

Name .....

Address .....

(WW)

## UMIST ELECTRONICS TECHNICIAN

Applications are invited from candidates of either sex for a vacancy of Electronics Technician Grade 5 in the Department of Pure and Applied Physics. The technician is required for the development, construction and maintenance of specialised electronic equipment for research and teaching using the full range of electronics workshop procedure. Applicants should be at least 24 years of age and preferably hold ONC, OND or equivalent qualification.

Salary within the scale: £3186-£3720 per annum (review pending).

Application forms may be obtained from the Registrar, UMIST, P.O. Box 88, Manchester, M60 1QD, by quoting reference PH/119/AU. (8844)

## UNIVERSITY OF SURREY LINGUISTIC & INTERNATIONAL STUDIES TECHNICIAN

Grade 5  
£3186-£3720 (under review)

A vacancy exists in this rapidly expanding Department. The successful candidate would take a prominent part in the day-to-day running of the Department's language laboratories.

Technical experience with audio and video tape recording apparatus and associated equipment, experience of film, slide or filmstrip overhead projection, and reprographic equipment are essential skills.

Educational requirement C & G. Radio and T.V. Technicians Cert.

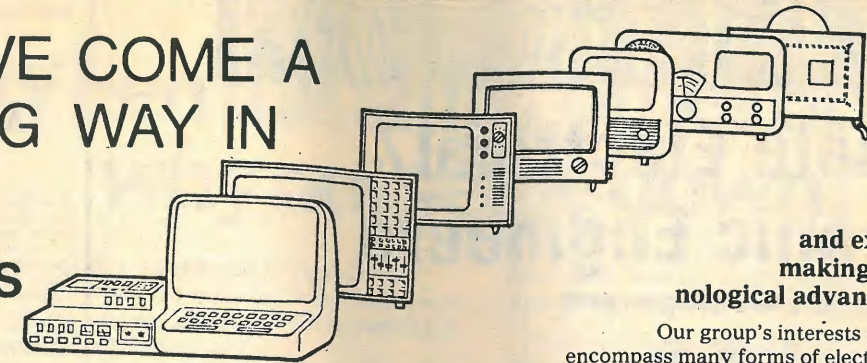
Application forms and further details may be obtained from the Staff Officer, at the University of Surrey, telephone 71281, Ext. 452. (8877)

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567 9705 evenings (8893)

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Our group's interests are wide ranging, and encompass many forms of electronics and communications, including computers, flight simulators, marine radio, satellite navigation and all aspects of television from receiver design and manufacture to cable systems and the supply of complete TV studios and outside broadcast vehicles.

Our Post Design Service on TV receivers offers technical assistance to our factories and rental companies throughout the UK, aimed at securing the best return on the huge investment in over 1 million operational television receivers. We require additional engineers for this important section, who are able to tackle a wide variety of re-design projects working very much autonomously and seeing their projects through to completion.

Formal qualifications in electronic engineering would be an asset, but most importantly we are looking for men and women with sound technical experience in some of the following areas: — colour and monochrome receiver re-design, detailed circuit fault investigation, safety investigations and approvals to BS415, component evaluations and receiver quality assurance and evaluation.

Career opportunities within this large and successful group are excellent, offering you the possibility of eventually designing a new generation of colour TVs or entering an allied field. You can expect an attractive salary, in line with your experience, together with generous benefits and relocation assistance to the Chessington area of Surrey, ideally situated for both London and the South Coast.

For further information please write or telephone:—

## REDIFFUSION

Mr J. Sinclair,  
Rediffusion Consumer Electronics Limited,  
Fullers Way Sth., Chessington, Surrey KT9 1HJ  
Telephone No. 01-397 5411

(8891)

## ELECTRONICS ENGINEERS AND TECHNICIANS

### THE COMPANY:

We are a young company experiencing vigorous growth, full of good ideas and successful in putting these ideas into practical uses.

We are now the dominant force in our original market area and have expanded into others.

### THE PRODUCTS:

Our products are state of the art, well conceived and built with care. To back this up we pride ourselves on the service our customers receive. Our products include Traffic Monitoring Equipment, Data Loggers, through to OEM Single Board Microcomputers.

### THE JOB:

We are looking for both Engineers and Technicians. Successful candidates would be involved in the design, development and debugging of microprocessors based products. The ability to work in an inventive and practical manner is essential. Knowledge of programming would also be an advantage.

### THE APPLICANTS:

These persons will be qualified and hold relevant degrees, HNC or HND, although the emphasis is on ability rather than qualifications.

### THE PAY:

For the Technician, circa £5,500 and for the Engineer circa £6,500. In addition a yearly bonus is paid depending on merit and company performance, up to 20% of yearly salary.

IT'S YOU? Then for an interview write or phone Roy Tuthill (Technical Director).



Golden River Company Limited  
Telford Road  
Bicester (086-92) 44551  
Oxfordshire

(8858)

## DESIGN/DEVELOPMENT ENGINEER

To work on the next generation of VHF/UHF ground to air communication equipment.

We are building up our design team to undertake all aspects of design and development from product conception through to production, including product definition, initial design, prototype development and the generation of full manufacturing information. This includes subsequent liaison with the Production Department.

The successful applicant will be a self-motivated engineer capable of working alone or as part of a team. A relevant Degree/HNC is desirable but this should not deter less qualified engineers with VHF/UHF experience.

The Company is located in a rural environment and offers an attractive salary package which includes a Contributory Pension Scheme, subsidised Canteen and concessional fare rebates on holiday air fares.

Applicants who are keen to join a small stable Company with excellent growth prospects should apply in the first instance for an Application Form to:

**Mrs M. R. Jennings, Park Air Electronics Limited, Northfields, Market Deeping, Peterborough PE6 8LG. Telephone: Market Deeping (0778) 345434.**



**PARK AIR ELECTRONICS LIMITED**  
(A Member of the IAL Group)

(8880)

**CAA**  
Civil Aviation Authority





## Graduate Electrical/ Electronic Engineers

### Research and Development in telecommunications

The Directorate of Telecommunications, London, is responsible for the extensive and sophisticated facilities used by the police, fire, prison and associated services. The role of the Research and Development Section is to ensure that maximum benefit is derived from the use of modern techniques.

The training and experience given to Graduate Engineers covers the training requirements of the IEE — ranging from the initial interpretation of a non-technical statement of requirement through to the management of design, development and contract — and is carefully planned by a senior engineer.

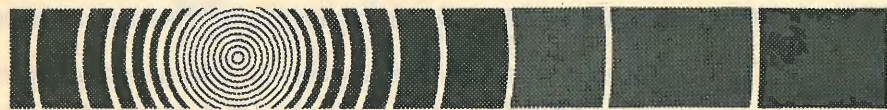
You should preferably be aged under 26 and must have (or obtain in 1979) a good honours degree in electronics or electrical engineering or an allied subject.

Your starting salary will be at least £4150. Completion of training (usually one or two years) leads to a salary rising to £6200. Further promotion prospects to £9190 and above. Non-contributory pension scheme.

For further details and an application form (to be returned by 15 February 1979) write to Civil Service Commission, Alencon Link, Basingstoke, Hants, RG21 1JB, or telephone Basingstoke (0256) 68551 (answering service operates outside office hours). Please quote ref. T/9998/2.

### HOME OFFICE

(8847)



## TELEVISION SOUND TECHNICIANS

Thames Television Limited has several vacancies for experienced Television Sound Technicians. These vacancies are based at our Riverside Studios in Teddington, Middlesex.

The posts involve the operation and maintenance of all television studio or outside broadcast equipment and entails both work on the studio floor or on location and in post production sound dubbing. Salary will be up to £5596 per annum (for a 38-hour week) depending on experience.

We also offer 21 days holiday, Company Pension Scheme, Subsidised Canteen and Restaurant and a Social Club and Bar.

For application form please write or telephone: Mike Allen, Staff Relations Officer, Thames Television Limited, Broom Road, Teddington Lock, Teddington, Middlesex. 01-977 3252 ext. 325.



(8849)

## Television Engineer

We are based in Buckinghamshire and operate a broadcast quality colour mobile unit and studio equipped with Link hand-held and studio cameras, Cintel Mark III telecine, VPR I recorders and a wide range of other facilities.

An experienced television engineer is now required for operational and maintenance work with our small team producing training programmes for the Services at base and on location.

You should have worked on professional colour equipment and some training could be provided, where necessary.

Good starting salary. Assisted travel allowance when applicable. Free canteen. Four weeks' annual leave.

Pleasant rural environment. Pension and Life Assurance Scheme.

For further information telephone or write to:

Personnel Officer  
The Services Kinema Corporation  
Chalfont Grove, Narcot Lane  
Gerrards Cross, Bucks SL9 8TN  
Chalfont St. Giles (02407) 4461

(8851)



The University of Liverpool  
Department of Electrical Engineering  
and Electronics

## SENIOR EXPERIMENTAL OFFICER

A vacancy exists in the Department of Electrical Engineering and Electronics for a Senior Experimental Officer to join a small group giving technical support across a wide range of research activities. These include power electronics, mass spectrometry, high current arc investigations, microprocessors and other digital systems as well as general electronic instrumentation.

Applicants should be of graduate or equivalent status with a minimum of five years' experience of electronic engineering and will be expected to have the ability to translate basic ideas into practical working systems with the minimum of supervision.

Salary within a range from £3883 to £6555 per annum according to qualifications and experience.

Application forms may be obtained from The Registrar, The University, P.O. Box 147, Liverpool L69 3BX.  
Quote Ref: RV/412/WW. (8870)

UNIVERSITY OF BATH  
SCHOOL OF MATHEMATICS

## TRAINEE TECHNICIAN

A vacancy exists for a trainee technician, aged about 18, to assist mainly in servicing and developing computing devices, including working with mini- and micro-computers. Candidates should have at least four relevant 'O' Levels and provide evidence of a keen interest in electronics.

Salary at 18 £1824 per annum (under review).

Application forms from the Personnel Officer, University of Bath, Bath, BA2 7AY, quoting reference number 78/268/WW. Closing date will be 31.1.79. (8889)

## Radio Officers

# When the ship comes home, why not settle down?

We're the Post Office Maritime Service and we have everything in a job that you'd want: the kind of work you're trained to do, good pay, job security and all the comforts of home where they really count — at home!

Vacancies exist at several coast stations for qualified Radio Officers to carry out a variety of duties that range from Morse and teleprinter operating to traffic circulation and radio telephone operating. And for those with ambition, the prospects of promotion to senior management are excellent.

You must have a United Kingdom Maritime Radio Communication Operator's General Certificate or First Class Certificate of proficiency in Radio-telegraphy or an equivalent certificate issued by a Commonwealth

Administration or the Irish Republic. And, ideally, you should have some sea-going experience.

At 25 or over, salary starts around £4093, rising after three years to about £5093. (Starting salary for those between 19-24 varies between £3222-£3732.) Overtime is additional, and there is a good pension scheme, sick-pay benefits and at least 4 weeks' holiday a year.

For further information, please telephone Andree Trionfi on Freefone 2281 or write to her at the following address: ETE Maritime Radio Services Division (WW/C/8) ETE17.1.1.2, Room 643, Union House, St. Martins-le-Grand, London EC1A 1AR.

## Post Office Telecommunications

(7141)



## ROYAL COLLEGE OF ART AUDIO TECHNICIAN

required in the Department of Environmental Media, to assist and advise students in the creative use of equipment, in the design and construction of "one off" pieces of equipment; to interface or modulate the standard departmental facilities; to maintain equipment. Applicants must have a thorough working knowledge of:—Sony video recording and editing facilities (1/2" open reel and U-Matic); sound recording and synthesizing equipment; multi-screen encoder / decoder and cross-fade unit; film and slide cameras and projectors.

The starting salary will be in the scale £3651-£4185 according to qualifications and experience.

Please write giving full details of age, qualification and previous experience to: **Assistant Registrar (Staff), Royal College of Art, Kensington Gore, SW7 2EU.** (8856)

## THE POLYTECHNIC OF CENTRAL LONDON Educational Development Unit Audio Visual Services TECHNICIAN Grade 4 / Language Laboratories Salary £3441-£3891

Duties will involve responsibility for day-to-day servicing and maintenance of language Laboratories in PCL, and will also include some work with basic Audio Visual equipment. Good working knowledge of tape recorders and preferably experience of language laboratory work is required.

Qualifications: C & G or equivalent and 7-9 years' experience.

Application form and job description from the Establishment Officer, PCL, 309 Regent Street, London, W1R 8AL (01-580 2020, ext. 212). (8871)

# We've variety and interest to offer you as a service and test engineer in Stanmore

It's the variety that comes with working on a wide range of equipment. And the interest of knowing that your skills and experience are playing a vital role in maintaining the critical standards demanded by major airlines and Air Forces for their highly sophisticated avionics equipment.

Working either in aircraft or in our well equipped and pleasantly situated workshops in Stanmore, Middlesex, you will be involved in the repair, maintenance and overhaul of a variety of advanced airborne electronics equipment, both British and American.

It's work for which you'll need

to have sound practical experience of radio and electronics theory, ranging from audio to microwave. You should also have experience of using advanced test equipment for fault diagnosis, although training can be given where necessary.

We can offer you an excellent salary and benefits together with really first-class working conditions and subsidised staff restaurant, so if it's variety and interest you're looking for write now with details of your experience to: Mrs. E. Wagg, Marconi Avionics Limited, 22-26 Dalston Gardens, Stanmore, Middlesex HA7 1BZ. Telephone: 01-2043322.

**MARCONI  
AVIONICS**

A GEC-Marconi Electronics Company

(8763)



## Opportunities in Test Engineering Digital and HF/VHF

Our range of equipment has an international reputation for its reliability under the most demanding operational conditions and it is the responsibility of our Test Departments to ensure this standard is maintained by fault finding and aligning the equipment, from sub units to complete systems, using sophisticated test equipment.

At our Chelmsford establishment we are now looking for additional men and women to test a wide range of HF/VHF systems and digital equipment.

Applicants should be qualified to Final C & G or H.N.C. in electronics and ideally have experience of radio or digital communications. A few vacancies also exist for those with intermediate C & G or with good practical experience of digital systems.

These positions carry attractive salaries, benefits and conditions of service and, in approved cases, relocation expenses will be met.

Write giving details of your experience to Mr. R. Humphries, Marconi Communication Systems Limited, New Street, Chelmsford, Essex CM1 1PL or telephone Chelmsford 53221 ext. 474.

A GEC-Marconi Electronics Company

**MARCONI**  
COMMUNICATION  
SYSTEMS

(8883)

### KING'S COLLEGE HOSPITAL MEDICAL PHYSICS TECHNICIAN III FOR ELECTRONIC EQUIPMENT SERVICING

Minimum qualifications an appropriate HNC or equivalent plus three years' relevant experience. The successful candidate would be responsible primarily for the maintenance under supervision of nuclear medicine imaging and counting equipment. Experience of similar work with sophisticated electronic systems essential though training on particular equipment will be given.

Salary in range: £4098-£5142 according to qualifications and experience.

Job description and further details from the Sector Administrator's Office, King's College Hospital, Denmark Hill, London SE5 9RS. Tel. 01-274 6222 ext. 2408.

(8910)

### BRISTOL POLYTECHNIC Faculty of Technology: Department of Engineering

#### RESEARCH ASSISTANT IN MICROPROCESSOR APPLICATIONS

Ref: R52/130

Salary: £3192 per annum.

The project is concerned with the development of microprocessor controlled data acquisition and processing systems.

Candidates should have a degree or equivalent in a scientific subject and some experience either with minicomputers or microprocessors.

Further details and application forms, to be returned by the 30 January 1979, from the Personnel Office, Bristol Polytechnic, Coldharbour Lane, Frenchay, Bristol BS16 1QY.

Please quote Post Reference Number R52/130 in all communications. 8914

## ELECTRONIC FIELD ENGINEERS GET INTO COMPUTERS IN GERMANY & U.K.

SMS is a successful Computer Maintenance Company, well-established in Europe with offices in London and Frankfurt and engineers in most EEC Countries.

Business is growing fast and SMS needs additional Electronic Engineers to train for Computer Maintenance in Germany and the U.K.

An HNC/ONC/FTC or equivalent in Computers, Electronics, Telecommunications, Radio or Radar together with 2-3 years' practical experience could be your chance to get into Computers now!

Let SMS train you as a **Computer Field Engineer**. If you are less qualified but still eager to work in Computers you could train as an **Associate Field Engineer**. The right applicants will be between 20 and 30 years of age with the ability to work with the minimum of supervision.

Salaries and benefits package will reflect the important role played by the engineer within S.M.S.

Please send particulars of your career to date to:

**ROY BARLING**  
Training Manager  
**SYSTEMS MAINTENANCE  
AND SERVICES LTD.**  
P.O. Box 13, Unitair Centre  
Great South West Road  
Feltham, Middlesex TW14 8NT  
England

(8866)

## MAINTENANCE SUPERVISOR

Rediffusion Reditune Limited, the World's leading music service, requires a MAINTENANCE SUPERVISOR to take charge of a small team responsible for the maintenance of professional recording and high speed duplicating equipment in a busy Dubbing Studio Complex at Orpington.

A formal qualification to H.N.C. Standard is preferred but not essential, but applicants must have a thorough knowledge of studio equipment and techniques.

Previous experience of control of staff will be an advantage.

The salary is negotiable around £4,200 per annum.

If you are fed up with shift work and irregular hours this could be the job for you!

Please write giving details of age and experience to:

**Chief Engineer**  
**Rediffusion Reditune Limited**  
Cray Avenue, Orpington, Kent



**REDIFFUSION REDITUNE**

THE WORLD'S LEADING COMMERCIAL MUSIC SERVICE.

(8867)

## Listening-in at 75 fathoms needs your kind of engineering experience



(8875)

Modern anti-submarine warfare relies heavily on detection devices such as the sono-buoys manufactured by UEL Electronic Communications Ltd that, after drop from an aircraft flying at up to 10,000 ft, deploy themselves automatically, lowering hydrophones to a pre-selected depth and raising a radio aerial so as to listen for tell-tale engine noises, amplify them and transmit the information back to submarine hunting aircraft.

The company, part of the international Dowty Group, also manufactures and develops communication control systems and intercom units for civil and military aircraft, airborne emergency radios, and beacons for homing and rescue applications. Our latest project is in the area of VHF radio where we are providing British Rail with a communication system between signal boxes and trains. Many of these systems need a high degree of ingenuity and the kind of engineering experience that maybe you can offer. In particular we are looking for the following men or women:

### Electronic Development Engineers

We are looking for men or women to join a small team of Engineers and Technicians working on the design of analogue systems and circuits. Visits to trials may be necessary

We are offering attractive salaries, negotiable according to qualifications and experience, plus a wide range of attractive large company benefits. There are good promotion prospects and generous relocation package is available where necessary covering all legal and estate agency fees, Building Society survey fees, viewing expenses, and a disturbance allowance.

and opportunities might arise for visits to clients and suppliers. Applicants should be qualified to HND or preferably degree level with several years design experience.

### Senior Development Technicians

We require men or women to join project teams working on the design and development of analogue systems and circuits for prototype equipment. Will be responsible for building, testing and evaluating experimental equipment and for assisting with the development of analogue circuitry.

Applicants, aged between 25 and 45, should hold City & Guilds Electronics, Radio & TV, or Telecommunications Certificates up to Part 1 or Intermediate level and have at least 5 years' development experience, preferably involving government contracts.

### Test Technicians

Our production department require additional male or female Testers with experience of radio or analogue circuits and test equipment. Candidates should have several years practical experience in this area with or without qualifications.

For further information and an application form phone or write to:-

**Mr Gavin Rendall, Personnel Manager,**  
**Uel Electronic Communications**  
**Limited, 419 Bridport Road,**  
**Greenford,**  
**Middlesex UB8 8AU.**  
Tel: 01-578 0081.

**UEL**

Electronic Communications Limited

## AIRWORK SERVICES LIMITED now require

### FIELD SERVICE ENGINEERS

for  
Installation and Commissioning  
of

**Ground Radar, Comms and Navigational Aids at various overseas tax-free locations**

also required

**DIESEL GENERATOR ENGINEERS and GENERAL FITTER/WIREMAN experienced with Stand-by and No-Break Diesel Generators and associated A.C. Power Distribution systems as used with Airfield Navigational Aids.**

Applicants should have ONC or similar qualifications and have at least 5 years' experience. Ex-Service Personnel would be particularly suitable.

You are invited to write in the first instance to the Senior Personnel Manager (Ref.).



**AIRWORK SERVICES LIMITED**  
Bournemouth (Hurn) Airport  
Christchurch, Dorset BH23 6EB

(8968)

## LONDON COLLEGE OF PRINTING

ELEPHANT & CASTLE, LONDON SE1

Department of Photography, Film and Television

## Television Technician/ Engineer (ST1/2)

Applications are invited for the above post in the School of Film and Television, Department of Photography, Film and Television. Candidates should be conversant with 1/2", 3/4" and 1" black and white and colour equipment and be capable of electronic maintenance. Experience in professional broadcasting would be an advantage, as well as an interest in experimental video work. The successful applicant will be expected to assist in running studio productions, and video tape editing.

Salary scale within: £3,640-£5,353 inclusive.

Application forms returnable within 14 days, obtainable from the Senior Administrative Officer at the College. Tel. No. 735 8484 Ext. 227.

**ilea**

INNER LONDON  
EDUCATION AUTHORITY



## Broadcast Transmission Engineers Train for a future in Broadcasting

We require Engineers qualified, or about to qualify, at least to HNC, HTC or equivalent level and possibly with a few years' experience, who will be trained to operate and maintain the advanced electronic equipment at our Transmitting Stations throughout the country bringing Independent Television and Radio into millions of homes. Our Engineers fulfil a mobile role, being required to maintain equipment anywhere within a specified area. You will, therefore, need a driving licence.

### Paid while you train

You must be prepared to undergo our demanding eighteen months training course, which combines theoretical study with practical familiarisation training on stations, which will give you a comprehensive knowledge of broadcasting operations and maintenance techniques. You will be paid a training salary on a scale up to £3,792 per annum (inclusive of a productivity supplement), depending upon experience.

### The Future

On completion of your training, you will be a full-time engineer on a salary range of £4,715 to £5,688 per annum, (inclusive of a productivity supplement). There are opportunities for advancement to higher grades.

Write or telephone for full details and an application form quoting Reference No. WVW/55ER to Mike Wright, Independent Broadcasting Authority, Crawley Court, Nr. Winchester, Hants. Telephone: Winchester 822574.



INDEPENDENT  
BROADCASTING  
AUTHORITY

8860

## APPOINTMENTS IN ELECTRONICS £5 - £10,000

Take your pick of the permanent posts in:

MISSILES — MEDICAL  
COMPUTERS  
RADAR — COMMS  
MICROPROCESSOR  
HARDWARE — SOFTWARE

For free expert advice and immediate action on salary and career improvement, phone or write to, Mike Gernat BSC.

**Technomark**

11 Westbourne Grove  
London W2. 01-229 9239



**CAPITAL  
APPOINTMENTS LTD.**

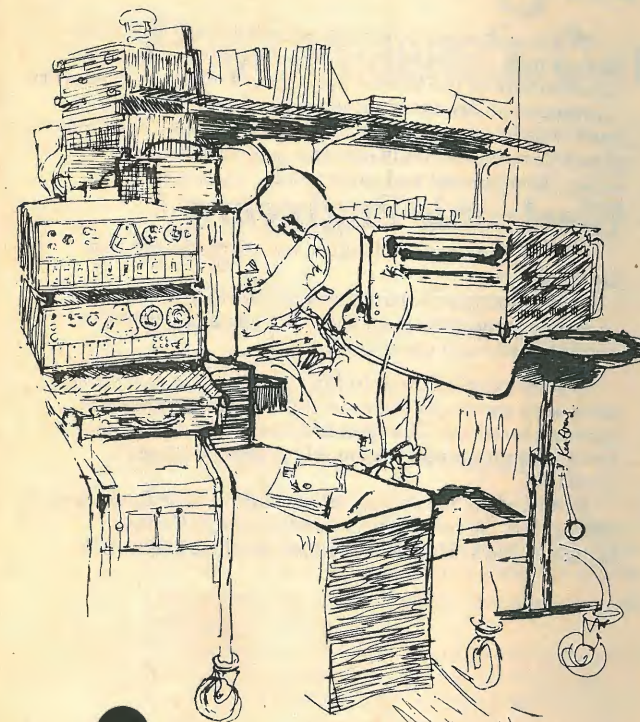
## FREE JOBS LIST

for  
**FIELD SERVICE ENGINEERS**  
BASIC SALARIES TO  
£7,000 + CAR

(8781)

30 Windmill Street, London, W1  
01-637 5551

# Marconi Instruments



## ELECTRONIC TECHNICIANS

**Opportunities for the experienced and sometimes inexperienced in St. Albans and Luton.** Work situations range from fault finding on PCB's and components, to batch product testing of equipment that utilise very advanced techniques including microprocessors and the repair/calibration of all manner and types of test instruments.

Attractive salaries and, where appropriate, relocation are offered for the right candidates. Further information may be obtained in confidence from John Prodger

**Marconi Instruments Limited,**  
Longacres, St. Albans, Herts. tel: St. Albans, 59292



A GEC-MARCONI ELECTRONICS COMPANY

(8638)

Experienced in  
television engineering?  
Join Middlesex Polytechnic as our

## Technical Operations Manager

Centre for Television and Film  
Salary scale rising to c. £7000 pa inc.

To be responsible for the operation, management and development of the Polytechnic's new television and film centre at Cockfosters, North London.

As the most senior technical member of staff, you will play a major role in the overall development of the centre, its personnel, and the services offered to students and staff.

Considerable experience in professional television engineering and film operations, a lively interest in research and development work, and appropriate qualifications are expected. An enthusiastic and outgoing attitude, together with a willing 'hands on' approach whenever needed, are also looked for.

Write quoting ref. 136.6F for further details and an application form, posting first-class to: Appointments Officer, Middlesex Polytechnic, Bounds Green Road, London N11 2NQ. Closing date 25 January.

(8863)

## Middlesex Polytechnic

## AUDIO + VIDEO LTD.

**We require top grade Engineers** capable of servicing and maintaining to a high standard, all types of video tape recorders. Experience of all U-matic, VCR, VHS, Betamax of 2" Quad machines is essential. Highest rates paid.

Other equipment in-house includes Vidicon and Flying Spot Mk III Telecine, DICE Standards Converter, TBCs etc.

Please phone Cliff Carroll on 01-580 7161

8711

## Quality Assurance

## Test Gear and Test Engineers WELLS, SOMERSET

EMI Electronics Limited, a Company within the EMI Group, is involved in the research, design, development and manufacture of advanced electronic equipment. The Company invites applications for the following posts.

### PROJECT QUALITY ENGINEERS

To work with development teams, with responsibility to ensure compliance with customers requirements, on a wide range of technologically advanced projects.

### TEST GEAR ENGINEERS

To maintain and calibrate a wide range of modern test equipment to Def. Std. 05-26.

### TEST ENGINEERS

To test and diagnose faults on: Analogue, digital and R.F. Circuits, or - pulse and R.F. Transformers.

We are looking for men and women of at least ONC (or equivalent standard) and with proven quality assurance experience.

Competitive salaries are offered together with (where appropriate) assistance with relocation expenses.

Please write or phone for an application form (quoting Ref. WW 104) to: D. K. Shires, Personnel Manager, EMI Electronics Limited, Wells, Somerset. (0749) 72081.

(8885)



**EMI Electronics Limited**  
Wells, Somerset BA5 1AA. Tel: (0749) 72081

A member of the EMI group. International leaders in music, electronics and leisure.

## How to find a better job without leaving your armchair.

Don't for a single moment question your motives. Striving for a higher income is a philosophy practised by people in all walks of life.

Perhaps though, you cannot get a fair picture of the opportunities available from the standard, limited sources of job information.

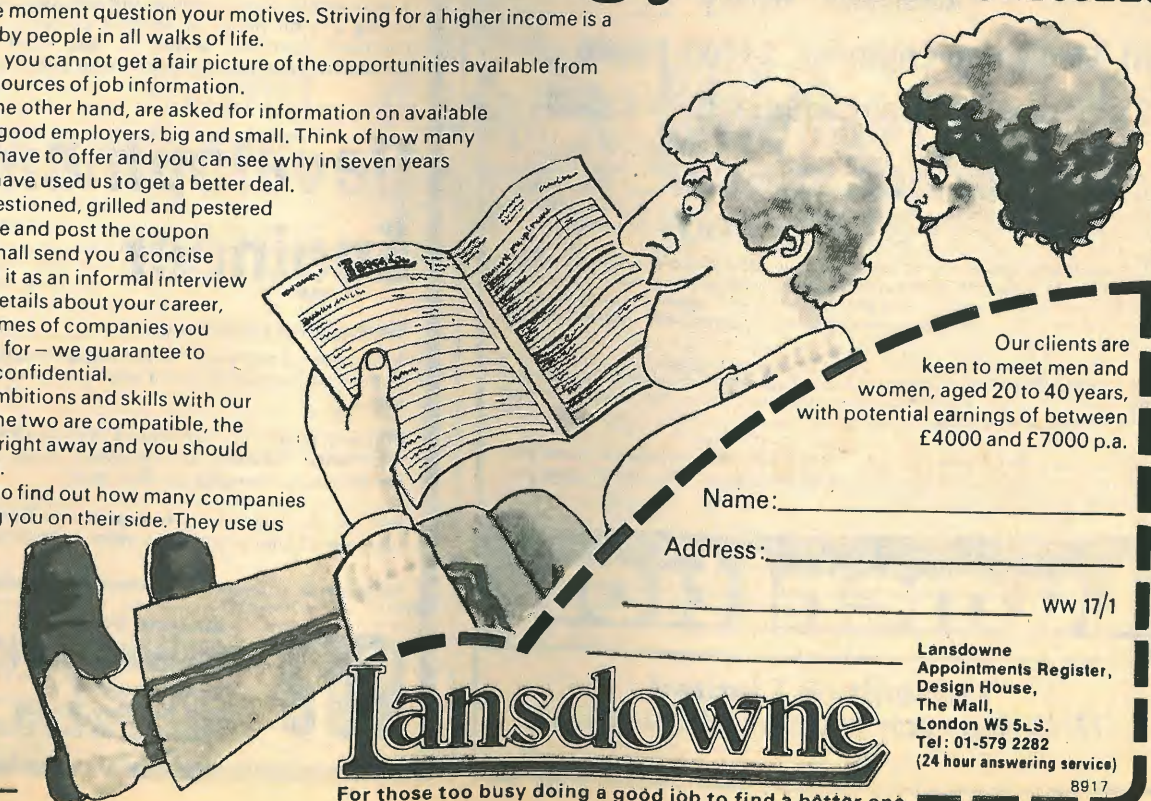
Lansdowne on the other hand, are asked for information on available people by over 3,000 good employers, big and small. Think of how many different careers they have to offer and you can see why in seven years thousands of people have used us to get a better deal.

You won't be questioned, grilled and pestered by us. Simply complete and post the coupon below. By return we shall send you a concise application form—treat it as an informal interview giving us all relevant details about your career, aspirations and the names of companies you would not like to work for—we guarantee to keep this information confidential.

We match your ambitions and skills with our clients' needs. When the two are compatible, the clients hear about you right away and you should get an invitation to talk.

Take this chance to find out how many companies are interested in having you on their side. They use us because our method is simple, quick, efficient.

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House, The Mall,  
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Tel: 01-579 2282  
(24 hour answering  
service).



Our clients are  
keen to meet men and  
women, aged 20 to 40 years,  
with potential earnings of between  
£4000 and £7000 p.a.

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Address: \_\_\_\_\_

WW 17/1

**Lansdowne**

For those too busy doing a good job to find a better one.

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Design House,  
The Mall,  
London W5 5LS.  
Tel: 01-579 2282  
(24 hour answering service)

8917



## Electronic Engineers

Make Hanslope Park  
your next stop

The Government Communications Centre at Milton Keynes now requires several of the above qualified staff to contribute to the Centre's growing reputation. Our work is often novel, always challenging and requiring a high level of dedication and application to the task. The fields of work are increasingly offering new opportunities for career development and experience.

### Electronic Engineers

VHF, HF general and digital circuitry, design and development.

Minimum qualification needed is HNC.

Salaries ranging up to £5448 per annum, depending on qualifications and experience.

We are situated close to Milton Keynes, a fast growing town with many modern entertainment, shopping and sporting facilities. The area is crossed by several main travelling routes and reasonably priced housing is available.

Please apply for an application form to the Recruitment Officer, HM Government Communications Centre, Hanslope Park, Hanslope, Milton Keynes, MK19 7BH.

Hanslope Park — is vital to your career. 8911

## ELECTRONIC ENGINEERING

### Automotive Industry

A) Electronic Engineer, £4000-£5000

B) Electronic Technician, £3000-£3500

Lumenition, an expanding company engaged in the automotive electronic ignition and fuel injection field wish to appoint additional staff to their research and development team. The company forms part of the successful Autocar Equipment Limited group. Lumenition automotive ignition systems are sold to customers throughout the United Kingdom and in countries overseas.

The appointees will be part of an engineering team which is dedicated to the development of new, advanced systems. They will need to have an engineering background and for position (A) a degree or HNC level qualification ideally with two or three years' experience in analog or digital systems and opto-electronics will be required.

For position (B) an ONC in electronics would be useful but not essential as the duties will be biased towards practice rather than theory and will involve the interpretation of diagrams.

Please write to the General Manager, stating age, experience, current salary and stating for which position you are making application. Indicate also how you consider you meet the requirements.

# Lumenition

Lumenition Limited  
77/85 Newington Causeway, London S.E.1

(8882)

## Electronics Engineering

Sound experience as important as qualifications.

A well established, growing organisation, manufacturing and marketing a wide range of synchro-conversion products and micro-computer systems, wishes to hear from experienced personnel particularly concerning the following appointment.

## Development Engineer

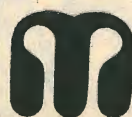
Starting circa £5,500 p.a.

A minimum of two years' sound practical experience in Analog/Digital techniques is essential and if you have an HNC or Degree so much the better.

Starting salary for the above appointment depends largely on experience. Salaries are reviewed twice yearly and our working conditions and benefits are first class.

We are an expanding operation in Surrey and provide ample opportunities for progress for men and women of ability.

Kindly write or telephone in the first instance:



Margaret Holland  
**MEMORY DEVICES LTD.,**  
Central Avenue  
East Molesey, Surrey  
Telephone 01-941 1066

8913

## Electronics Engineers and Technicians Electromagnetic Compatibility

The Underwater Weapons Division of Marconi Space and Defence Systems at Portsmouth needs Professional ELECTRONIC DESIGN ENGINEERS AND TEST TECHNICIANS, men and women, with experience in at least one of the following fields to join the Division for work on an important new weapon.

### RADHAZ

Radio Interference Suppression  
Electronic System Compatibility

EMC Testing

Filter Design

Shielding Design

EMC Analysis and Prediction

The specialist group in which you would work forms part of a large team. As a member of that team you would have the opportunity of making a significant contribution to the successful attainment of required weapon performance.

Engineers can expect to be involved in some of the following activities:

Critical Analysis of Weapon and Test Equipment Design.  
Preparation of EMC Test Plans and Procedures.  
Support of RADHAZ, EMP and MAGNETIC tests.

Investigation of EMC problems.

Specification of EMC designs and performance requirements.

Analysis of Test Results.

and Technicians will be involved in:

Operation of EMC Test Equipment.

Analysis of Test Results.

Design and construction of test aids.

Support of engineers.

We can offer you a salary that reflects the true value of your qualifications and experience and an extensive and worthwhile benefits package. If you are not familiar with the attractions of living and working on the South Coast, you are in for a pleasant surprise.

All you have to do now is to decide which of these jobs suits your experience and your ambitions.

Please telephone or write to Ken Hoxey, Marconi Space and Defence Systems Limited, Browns Lane, The Airport, Portsmouth, PO3 5PH. Tel: Portsmouth 64966 Ext. 305 quoting reference CDA.

## Marconi Space & Defence Systems (Portsmouth)

A GEC-Marconi Electronics Company

8862

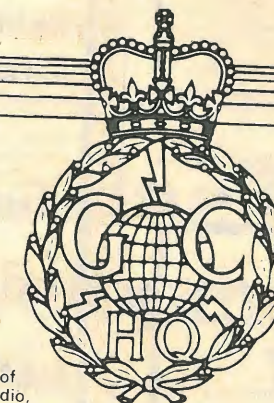
## RADIO TECHNICIANS

At the Government Communications Headquarters we carry out research and development in radio communications and their security, including related computer applications. Practically every type of system is under investigation, including long-range radio, satellite, microwave and telephony.

Your job as a Radio Technician will concern you in developing, constructing, installing, commissioning, testing, and maintaining our equipment. In performing these tasks you will become familiar with a wide range of processing equipment in the audio to microwave range, involving modern logic techniques, microprocessors, and computer systems. Such work will take you to the frontiers of technology on a broad front and widen your area of expertise — positive career assets whatever the future brings.

Training is comprehensive special courses, both in-house and with manufacturers, will develop particular aspects of your knowledge and you will be encouraged to take advantage of appropriate day release facilities.

You could travel — we are based in Cheltenham but we have other centres in the UK, most of which, like Cheltenham are situated in environmentally attractive locations. All our centres require resident Radio Technicians and can call for others to make working visits. There will also be some opportunities for short trips abroad, or for longer periods of service overseas.



## WORK IN COMMUNICATIONS R&D AND ADD TO YOUR SKILLS

You should be at least 19 years of age, hold, or expect to obtain shortly, the City and Guilds Telecommunications Technician Certificate Part 1 (Intermediate), or its equivalent, and have a sound knowledge of the principles of telecommunications and radio, together with experience of maintenance and the use of test equipment. If you are or have been in HM Forces your Service trade may allow us to dispense with the need for formal qualifications.

You start on £2927 at 19, up to £3700 if you are 25 or over, rising to £4252, and promotion will put you on the road to posts carrying substantially more. There are also opportunities for overtime and on-call work paying good rates.

Get full details from our Recruitment Officer, Robby Robinson, on Cheltenham (0242) 21491, Ext. 2269, or write to him at GCHQ, Oakley, Priors Road, Cheltenham, Glos. GL52 5AJ. If you seem suitable, we'll invite you to interview in Cheltenham — at our expense, of course.

(8508)



## PERSONNEL REQUIRED FOR SAUDI ARABIAN TELEVISION NETWORK

### Three Chief Engineers

with considerable experience in Television transmission maintenance. 8200/00 SR/month.

### Four of Class A Engineers

with experience in maintaining telecine, studio cameras and associated studio video and audio mixers and processors preferably with transmission experience also. 7400/SR/month.

### Two of Microwave and Transmitter Engineers

with experience in maintenance of microwave and TV transmitters. 7400/00 SR/month.

### Two of Class A Engineers

with experience in maintenance of VTR and studio video and sound equipment. 7400/00 SR/month.

### Six Utility and Powerhouse Supervisors

with maintenance experience Diesel Generators up to 250 KVA and Cubical Switchboards up to 750 KVA 400 Volt systems, with control circuitry and switching for mains/diesel systems, three phase wiring and distribution systems, single and three phase motors and pumps, wiring of breaker boards, lighting and heating distribution, maintenance of self-contained and central air-conditioning systems up to 80 tons of both radiator cooling and chilled water types, including associated handling units, cooling tower maintenance, switch gear, relays and controls. 6000/00 SR/month.

Furnished accommodation and other benefits for all positions as per the Contract of Employment which includes return air tickets, 30 days annual vacation, medical treatment, etc.

All applicants should have B.Sc. in Electrical or Electronic Engineering with a minimum of five to seven years' experience, or a two years' High Technical Certificate from a recognised institution or its equivalent with nine to twelve years' maintenance.

All salaries mentioned above are negotiable and dependent upon qualifications and experience.

Applicants should be ready for interview in two or three weeks from now and should be prepared to move to Saudi Arabia immediately after their interview.

Applications to:

**RICHARD JOB LIMITED**  
41 Whitehall, London SW1

Approx. 6.5 Riyals = 1 Sterling.

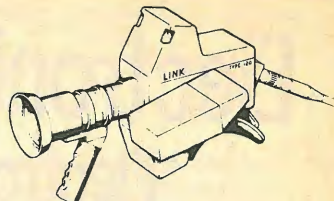
(8908)

## SOUTHAMPTON AND SOUTH WEST HAMPSHIRE HEALTH DISTRICT Senior Electronics Technician

To be responsible for servicing, maintenance, repair and testing of electro-medical and electronic equipment, including some communication equipment, in an established and expanding team based on Southampton General Hospital site. Candidates should possess Ordinary National Certificate or preferably Higher National Certificate or equivalent in Electronic Engineering. Salary £3,744-£4,788. This post offers wide experience and practical training in medical electronics. Application form and job description from Personnel Department, Tremona Court, Tremona Road, Southampton, telephone Southampton 777222, Ext. 3556. Closing date 9th February, 1979.

(8857)

# LINK



# TELEVISION

### WE ARE

Link Electronics Ltd., a successful expanding company with room for individual ability to make itself felt.

### WE MAKE

A full range of TV studio broadcast equipment, including colour cameras.

### WE NEED

DEVELOPMENT ENGINEERS at senior and junior levels, to put good theoretical knowledge into practice in circuit design of all types of broadcast equipment, employing the latest techniques. You must have 2-3 years' relevant experience in industry, preferably obtained in a TV or similar environment.

SENIOR TEST ENGINEER to undertake test and commission of advanced and complex TV cameras and associated equipment. This appointment is at a senior level and so direct experience of similar equipment is a must.

### WE OFFER:

#### SALARY

Above average, according to ability and not a rigid grade structure.

#### BENEFITS

Generous holidays, free life and health insurance, pension scheme, staff restaurant, relocation expenses.

#### LOCATION

A modern factory in a very pleasant part of Hampshire with no traffic problems and easy access to London, the South Coast and many major towns.

#### HOUSING

A wide choice. Prices from about £15K upwards if you want to buy.

#### TO APPLY:

Either phone Jean Smith at Andover (0264) 61345 and ask for an application form or write with enough information to make form unnecessary.

**LINK**  
ELECTRONICS

North Way, Andover  
Hampshire, England Telephone Andover (0264) 61345

(8701)

## COVENTRY AREA HEALTH AUTHORITY WALSgrave HOSPITAL MEDICAL PHYSICS TECHNICIANS GRADE II AND III

are required by the Clinical Physics and Bio-engineering Department to join a team involved in the maintenance and development of a wide range of physiological measurement equipment, nuclear equipment and laboratory equipment. A knowledge of diagnostic maintenance of instrumentation and/or mini computers would be desirable.

Candidates for the posts should hold an ONC, HNC or equivalent qualification. Salary scales: Medical Physics Technician, Grade III, within the range £3744 to £4788 per annum. Medical Physics Technician, Grade II, within the range £4470 to £5610 per annum. Further details can be obtained from Chief Physicist, Walsgrave Hospital, Tel. Coventry 613232, ext. 482. Application forms (quoting ref: WW) obtained from the Sector Administrator, Walsgrave Hospital, Clifford Bridge Road, Walsgrave, Coventry, CV2 2DX. (8886)

## HATFIELD POLYTECHNIC SCHOOL OF HUMANITIES LANGUAGE LABORATORIES REF 146

### TECHNICIAN (degree level)

for the servicing and repair of electronic equipment within Laboratories and Departments. This will also involve the repair of Tape and Cassette Recorders, Radio Tuners, Amplifiers, etc.

Work includes the use of the Recording Studio, and the editing and copying of tapes and cassettes.

Candidates, male or female, must be ambitious and able to work on their own initiative.

#### Salary Scale:

T1: rises to a maximum of £3399 inclusive.  
T2: £3399-£3771 inclusive, according to age and qualifications.

Informal enquiries to Mr. A. Wellington, Extension 337.

Application forms for the above post are available from the Staffing Office, The Hatfield Polytechnic, P.O. Box 109, Hatfield, Herts, AL10 9AB. Telephone Hatfield 68100, extension 309. Please quote reference number. (8887)

## BORDER TELEVISION LIMITED

have vacancies at their Carlisle Studios for

## TELEVISION BROADCAST ENGINEERS

Applicants should be experienced in telecine, VTR and/or CAR operations and maintenance. An academic qualification would be an advantage. Commencing salary up to £6024 per annum (including supplement) depending on qualifications and experience. Applications, in writing, to:

Chief Engineer  
BORDER TELEVISION LIMITED  
Carlisle, CA1 3NT

(8884)

## TOP JOBS IN ELECTRONICS

Posts in Computers, Medical, Comms, etc. ONC to Ph.D. Free service.

Phone or write: BUREAUTECH  
AGY, 46 SELVAGE LANE,  
LONDON, NW7. 01-959  
3517.

(8912)

## Can we bring your interest in technical publications to light?

For a company which designs and builds complex electrical and electronic equipment, the technical handbooks on which service engineers and customer operators rely are of vital importance.

At Rank Strand Electric, we need a man or woman with an ONC or sound general electronics knowledge to look after our technical publications section.

We develop and install professional sound and lighting systems, some computer-controlled, and it will be your responsibility to originate or edit technical copy, prepare layouts and indexes for all handbooks, arrange economical printing and then ensure that the information can be accessed when required.

Your experience to date may have been gained as an electronics technician, service engineer or technical information officer. The only real requirements are familiarity with the technology organisational ability and a flair for communicating technical data verbally as well as in writing.

Please contact Valerie Hutchings, Personnel and Training Officer,  
Rank Audio Visual, P.O. Box 70, Great West Road, Brentford,  
Middlesex. Telephone: (01) 568 9222.

**RANK AUDIO VISUAL**

8920

## Marine Transmitter Development Engineer

In Marine Division we are responsible for the design and development of advanced HF Communication systems for the Merchant Navy at both low and high powers using solid state techniques as well as valves.

The two engineers we are seeking will have a professional qualification and some years experience of one of the following: Transmitter RF design, solid state drives and frequency generation equipment in general. You should also have the ability to progress your design through the production phase.

Write or phone for an application form to Mr. G. Short, Marconi Communication Systems Ltd., New Street, Chelmsford, Essex. Telephone Chelmsford 53221.

A GEC-Marconi Electronics Company

**MARCONI  
COMMUNICATION  
SYSTEMS**

(8859)



## WESSEX EDUCATIONAL TELEVISION CONSORTIUM

### CCTV ENGINEER

An engineer is required for the Consortium's mobile CCTV production unit which makes video programmes for use in initial and in-service teacher education in this area.

Applicants should have a graduate or full professional qualification.

Inclusive salary ranges from £4773 to £6060, depending on age, qualifications and experience.

Further details from The Director, Wessex Consortium, King Alfred's College, Winchester, SO22 4NR. Phone (0962) 66359.

Closing date for applications is February 2. (8873)

## CAPITAL APPTS.

### FREE LISTS

101 Design/Development and Test Jobs

Permanent and Contract

To £6,000

(8782)

637 5551 day, 636 9659 eve.

## Broadcasting Engineer

enjoy the benefits of working at the fore of broadcasting technology

We require an Engineer, male or female, qualified to HNC or equivalent level, with at least three years' experience in a broadcasting discipline to maintain the advanced electronic equipment at our main transmitting stations and relay stations bringing Independent Television and Independent Local Radio into millions of homes.

The successful applicant will become a member of a team based at our Black Mountain Transmitting Station, Hannahstown, Dunmurry, Northern Ireland and will be required to travel throughout Northern Ireland. A current driving licence is required. The post will involve some week-end working. Equipment familiarisation training will be provided.

The salary will be on a scale up to £5,688 per annum, with provision for movement on to a higher scale rising to £6,399 per annum. These rates are inclusive of a productivity supplement.

Relocation expenses will be paid where applicable.



INDEPENDENT  
BROADCASTING  
AUTHORITY

Act now! Write or telephone for an application form quoting Ref. No. WW/35ER to: Mr M. Wright, Personnel Officer - Engineering Regions, Independent Broadcasting Authority, Crawley Court, WINCHESTER, Hants. SO21 2QA. Telephone No: Winchester 822574.

8861

## Oxfordshire AREA HEALTH AUTHORITY (TEACHING)

### ELECTRONICS TECHNICIANS

M.P.T. III £3,744-£4,788

M.P.T. IV £3,069-£4,143

For the Nuffield Department of Anaesthetics, a busy University Teaching Hospital Department. These new posts are due to the imminent opening of the John Radcliffe Hospital. Duties will include the maintenance and repair of electronic patient monitoring and laboratory apparatus, as well as modification and development of clinical equipment. The Department offers an electronics service to many other departments and possesses a strong technical structure.

Applicants for the Grade IV post will have 2 years' experience and 5 years' experience for the Grade III post and will be qualified to ONC/HNC (Electronics).

Application forms and further particulars from Mr. J. B. Thompson, Principal Chief Technician, Nuffield Department of Anaesthetics, Radcliffe Infirmary, Oxford.

(8855)

## SALES ENGINEER

Zettler require an enthusiastic Sales Engineer for their rapidly expanding relay sales programme.

The person appointed, aged 20/40, should have an electrical engineering background and preferably some sales experience. Applicants without previous sales experience will also be considered.

A good salary, commensurate with experience and ability, will be paid, plus company car. Assistance given with re-location to Harrow area.

Applications in writing please, giving brief career details, to The Divisional Director, Zettler UK Division, Brember Road, Harrow, Middlesex HA2 8AS. Tel. No: 01-422 0061.

(8888)

## Design / Development Engineer

Labgear Limited, a member of the Pye of Cambridge Group of Companies involved in the development and manufacture of Television RF distribution, Service Test and Teletext equipment, requires an Engineer, male or female, of at least HNC or equivalent standard with a minimum of two years' experience in electronics, preferably with the television industry.

The post could, alternatively, offer an excellent opportunity to a university graduate to enter an industry which is continually expanding in both analogue and digital techniques.

This pensionable post carries an attractive salary, 20 days' holiday and where appropriate assistance with relocation expenses.

Please write giving details of your experience to:  
Mr. C. Houghton, Personnel Manager



Labgear Ltd  
Abbey Walk, Cambridge CB1 2RQ  
Tel: Cambridge 66521

(8912)

## THE UNIVERSITY OF LEEDS DEPARTMENT OF CARDIOVASCULAR STUDIES

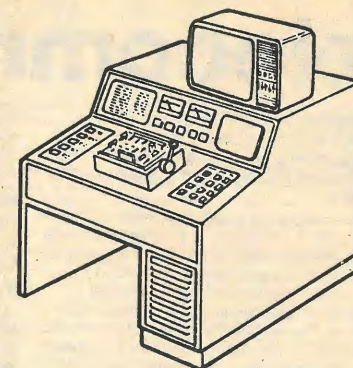
Applications are invited for the post of ENGINEER OFFICER (Electronics) in the above Department which operates in the New Medical School and two hospitals. Responsibilities will include the design and development of electronic equipment for use in biological and physiological instrumentation and the successful candidate will be expected to collaborate in research into new applications for physiological investigations. He/she will also be expected to provide effective liaison with the electronics technicians in the University department and the hospitals. Applicants should have an honours degree in Electronic Engineering together with proven experience in the application and development of electronic instrumentation equipment preferably in the medical field. A knowledge of digital techniques including the use of microprocessors would be an advantage.

Salary at an appropriate point on either the grade IA or II scale for Other Related Staff (£3883-£6555 or £6317-£7754) according to age, qualifications and experience.

Application forms and further particulars may be obtained from the Registrar, The University, Leeds LS2 9JT, quoting reference number 105/9/C1. Closing date for applications: 5th February, 1979.

(8854)

## TEST EQUIPMENT ENGINEERS



Come and join Rediffusion and help us design a brand new range of test equipment for production testing our future colour receivers. Our present equipment is a combination of sophisticated manual and automatic test consoles employing both analogue and digital circuitry. We are currently looking at the application of microprocessors to the testing of television modules and are confident that our future equipment will take advantage of these latest techniques.

Applications are invited from well qualified and experienced test equipment engineers to join our energetic young team at Chessington, Surrey, working in a congenial and stimulating environment. We have vacancies at both senior and intermediate levels, offering excellent salaries and opportunities for career advancement in this exciting field.

Since engineers are expected to see their projects through to completion, some travelling to our factories in the North East is expected to assist on installation and commissioning and to give back-up service where necessary.

Additionally, an intermediate engineer is required to be based at Bishop Auckland in our Engineering Department on the main factory site. This position involves some design and development work interspersed with the planned maintenance of complex production test equipment.

If you are interested in these challenging positions and would like more details or wish to discuss the matter in depth, please write to or telephone:



## REDIFFUSION

Mr H. Brearley,  
Rediffusion Consumer Electronics Limited,  
Fullers Way Sth., Chessington, Surrey KT9 1HJ  
Telephone: 01-397 5411

8892

## ELECTRONIC BLOKES!

### YOU ARE NEEDED AND APPRECIATED!

Electronic Blokes have evolved from the dedicated electronic engineers of the 1960s. Past masters at doing their job with the minimum of effort, their unequivocal zeal is directed into — designing their own hi-fi, reading scientific journals (particularly Sits Vac), stocking the bit box and creating an amicable working environment — their hole.

Our client, a manufacturer of computer peripheral equipment needs electronic blokes who will desist from these activities for a few hours and help their employer try to:

1. Improve track location and reduce access times on head positioners.
2. Devise methods to identify and eliminate data errors on disc files and floppys.
3. Get rid of bugs and beetles in heads, discs and data channels.
4. Design test equipment to measure nano-second wide pulses.
5. Evaluate methods for testing — guess what? — the microprocessor itself.

CASH REWARD — Excellent for the bloke who wants to get off his backside, display some enthusiasm, and develop his financial acumen.

Location 20 miles due west of London.

For further details, contact:

## Charles Airey Associates

"PROBABLY THE BEST KNOWN SUPPLIER OF ELECTRONICS ENGINEERS IN THE COUNTRY" — FINANCIAL TIMES

155 KNIGHTSBRIDGE, LONDON, SW1. TEL: 01-581 0286

(8864)

## Electronic Engineers

£4815 — £6643

There are vacancies for electronic engineers with experience in the design and specification of audio frequency, CCTV, CATV and mobile radio systems. These vacancies are in a number of grades with differing responsibilities and requirements.

Qualifications required for the higher positions are degree or chartered engineer status (IEE or IERE).

HNC, ONC, City & Guilds or equivalent are required for the junior posts.

Appointments will be made depending upon qualifications and experience. We offer excellent conditions of employment including 4 to 5 weeks leave and a pension scheme.

Write or telephone for an application form, returnable by 2 February to the Establishment Officer (ME/Estab/931/AC/021) Room 510, 1 Queen Anne's Gate Buildings, Dartmouth Street, London SW1H 9BS. Tel: 01-633 5700.

## GLC Mechanical & Electrical Engineering

(8865)



## SITUATIONS VACANT

# Telecommunications

We require staff, male or female, to prepare and maintain the latest in communications equipment used by the Police and Fire Brigade in England and Wales.

You will need to be qualified at least to City and Guilds Intermediate Telecommunications standard and be able to demonstrate practical skills in locating and diagnosing faults in a wide range of equipment from computer based data transmission to FM and AM radio systems. You would live near to and work from one of our service centres located throughout England and Wales or our Headquarters in the London area. Specialised courses of training are run to assist staff to keep up to date with developments and new equipment, and there are opportunities for day release to gain higher qualifications. Applications from registered disabled persons will be considered.

Promotion prospects are good and the

work represents a secure future with generous leave allowances and a non-contributory pension scheme.

Possession of a driving licence is essential since some travelling will normally be involved.

The salary is £2627 (at 17), £3176 (at 21) and £3700 (at 25), rising to a maximum of £4252.

If you are interested in working with us, then write for further details and an application form stating where you are interested in working, to:

Mr C B Constable Directorate of Telecommunications Horseferry House Dean Ryle Street LONDON SW1P 2AW Telephone: 01-211 6420



HOME OFFICE

8894



**Dublin Corporation**  
Bárdas Átha Cliath

## ASSISTANT SUPERINTENDENT (Waterworks)

Dublin Corporation has a vacancy for an Assistant Superintendent in its Waterworks Section. The successful candidate will be based at Ballymore Eustace Reservoir.

**QUALIFICATIONS:** Applicants must:

- hold a National Diploma in Engineering (Electronics, Industrial Electronics or Telecommunications Electronics) or an equivalent qualification in Electronics or Instrumentation AND
- have had at least three years' satisfactory experience in Sanitary Services Works (water supply and/or sewerage or in Industrial engineering).

**AGE LIMIT:** Not less than 23 years and not more than 45 years on 1st February, 1979.

**SALARY SCALE:** £5,597 x £121 - £5,839 x £120 - £6,559 per annum.

For application forms apply to: Personnel Department, 55 Angler Street, Dublin 2.

Latest date for receipt of completed application forms: Friday, 16th February, 1979.

(8876)

**UNIVERSITY OF LEEDS.** Electronics Technician - Grade 5 required in the Department of Physiology. The person appointed will be responsible to the Supervisor of the Electronics Workshop for the development, construction and maintenance of electronics equipment associated with research and teaching of biological studies. Candidates should hold ONC or equivalent qualification in relevant subjects and have at least, 5 years appropriate experience. Salary on the scale £3,186 to £3,720 (under review) according to qualifications and experience. Applications stating age, qualifications and full experience, together with the names and addresses of two referees, should be addressed to Mr E. French, Departmental Superintendent, Department of Physiology, Medical and Dental Building, Leeds LS2 9NQ. (8783)

**RADIO TELEPHONE SERVICE ENGINEER** required in Croydon. Proven ability to repair equipment more important than formal qualifications. Salary commensurate with ability. Contact LONDON CAR TELEPHONES on 01-680 1010. (8822)

### ARTICLES FOR SALE

**LOW COST LOGIC ANALYSER** Paratronics Model 100A plus Model 10 Trigger Expander and full instructions. Use with scope for memory mapping etc. Also an A.T.E. Systems Ferret Model E56200 circuit tester plus frequency to voltage plug in unit, plus add-on p.c.b. so that programs can be stored in 2716 PROMS. Both in first-class as new condition. Offers phone Biceter (086 92) 44551, ask for R. Tuthill. (8894)

### A TELEVISION STUDIO ENGINEER

is required to assist in the operation and maintenance of the University's television unit and other audio-visual production.

Applicants should have a wide experience of television studio techniques including lighting, camera operation and vision control, and a good knowledge of sound recording.

Experience of helical scan video recording systems would be an advantage.

Salary scale (under review) £4107-£4749 according to age and experience.

Write giving details of age, qualifications and experience to: The Director, University of London Audio Visual Centre, 11 Bedford Square, London WC1B 3RA. (8918)

**ELECTRONICS TECHNICIAN** (Grade 5) required in Department of Psychology, University of Reading. Candidates should have, or be completing, Final C & G in electronic servicing or equivalent. Salary in scale £3186-3720 p.a. Apply with full details and names of 2 referees, quoting Ref. T.W.W. 66A, to Assistant Bursar (Personal), University of Reading, Whiteknights, Reading RG6 2AH. (8901)

## AMPEX

the world's leading manufacturers of magnetic recording equipment have the following vacancies:

### BROADCAST ENGINEERS resident in Africa and the Middle East

We are looking for Broadcast Engineers with VTR experience to take up resident engineering positions throughout Africa and the Middle East.

We offer:

- ★ Attractive salaries
- ★ Home leave every three to four months
- ★ Free accommodation
- ★ Free transport
- ★ Free medical insurance

If you have had several years' experience in broadcasting and are looking for something different, interesting and demanding, why not give us the opportunity of discussing the above contracts with you.

### BROADCAST ENGINEERS based in Athens

How would you like to join a team of professional engineers based in Athens, to travel the Middle East and Africa and to receive product training on the world's leading video tape machines? Attractive salaries and conditions are offered for these positions.

Please send your curriculum vitae to **Mr. Don Cameron, Engineering Manager, Ampex World Operations S.A., P.O. Box 45, Halandri (1), Athens, Greece.** (8874)

### ARTICLES FOR SALE

**LAB CLEARANCE:** Signal Generators; Bridges; Waveform, transistor analysers; calibrators; standards; millivoltmeters; dynamometers; KW meters; oscilloscopes; recorders; Thermal, sweep low distortion true RMS, audio FR. deviation. Tel. 040-376236. (8250)

**G.W.M. RADIO LTD.,** 40/42 Portland Road, Worthing, Sussex. Tel. 34897. Pneumatic masts 40ft. By Scam Clark. 300 watt radar calorimeters, noise generators, type CT410, Eddy-stone communication receivers 730/4, v.g.c. £185.00 inc. Many bargains for callers, surplus always wanted. (8832)

**PATCH BOARD V.G.C.** 60 x 60 single XY bussing on 6mm matrix. Also pins and plugs, suit big synth, rom programmer, mixing desk, etc., £370. Mike Boyd, 0234 750111, ext 273 or 751104 evenings. (8902)

**TEKTRONIX OSCILLOSCOPES** type 545A DC/24MHz with type CA D/trace plug-ins, handbooks, in good clean condition, £250 each inclusive. 14 Braidley Road, Bourne-mouth 0202 29148. (8906)

**WIDE selection** unused discrete and logic (TTL/ECL) components; pre-aged 5MHz 3rd. overtone crystals; 7 segment 0.63" LED displays; muffin fans; transformers, etc. Full stock list sent on request. Send S.A.E. to Coope Durrant Ltd., 66/67 High Street, Lymington, Hampshire, England SO4 9YZ. Tel. Lymington (0590) 77971. (8835)

**MARCONI SINE SQUARE PULSE** bar generator TF 2905/8, £400 mint; TF 791D deviation meters £100; 1 Henry standard inductance tinsley 4109C £100; AV08, £65; AVO 7D £50. Please add 8% VAT. G. Services Electronic Camberly Ltd, 0252 871048. (8900)

## ARTICLES FOR SALE

HP 8551 Spectrum Analyser (0V 42 GHz). R&S Polyskop I (0.5-400 MHz). Tek 531, 532, 545, 585, 575, 581 Scopes (inc. 453). Solartron LM 1420 D.V.M. Telegraph Distortion Measuring Equip. Nagard, E-H Pulse Generators. Telonic LH-2M Sweep Oscillator. A.F. & R.F. Signal Generators. Marconi TF 1041 V.T.V.M. Advance VM 77D A.C. Millivoltmeter. R&S Sweep Oscillator. Sandoz SG313 Sig. Gen. (4-7GHz). Marconi TF1313 1/4 Universal Bridge. Marconi T.V. Test Equip. Siemens Level Meter (10KHz-17MHz). Wandel & Goltermann Noise Receiver. MP 250B RX Meter (0.5-250 MHz). Marconi TF801B Sig. Gen. (12-425 MHz). Advance Timer Counters TC6, TC4. Farnell Stats P.S.V. 30T/2/EC. MP185B Sampling Scope (>10ns/cm). Airmec 365A Sig. Gen. (1-320 MHz). Standards Equip.: Potentiometers, Bridges, S/S Meters, Capacitors, COHU Thermal Transfer Standard Model 666 & Galvo. Marconi TF 867/2 Standard Sig. Gen. (15 KHz-30 MHz). Marconi TF 1066 FM & AM Sig. Gen. (10-420 MHz). Advance V.H.F. Millivoltmeter VM79. BPL 10KV Insulation Tester. Rascal 835 Universal Counter. Paper Tape Readers ★ IBM Printers with Keyboards ★

Compare our prices.

### P&R TEST EQUIPMENT

Salcott Mill, Goldhanger Road, Heybridge Near Maldon, Essex (8748)

Open Saturday: Tel. Maldon (0621) 57440

### SOWTER TRANSFORMERS

WITH 37 YEARS' EXPERIENCE we have the expertise to design and manufacture ANY TYPE OF AUDIO TRANSFORMER AT THE RIGHT PRICE.

Whilst we specialise in every kind of transformer for audio control desks and mixers demands are increasing for **LOUDSPEAKER TRANSFORMERS** and **100 VOLT LINE AUDIO OUTPUT TRANSFORMERS** FOR MOST KINDS OF AMPLIFIERS FROM 30 WATTS TO 500 WATTS OR MORE. We can also supply multi-output transformers for **COLUMN LOUDSPEAKERS**. A recent tendency is the demand for **OUTPUT TRANSFORMERS FOR ULTRA LINEAR AMPLIFIERS** using KT 88 and KT 66 BEAM TETRODES and for these we have standard designs with exceptional performance.

We call your attention to our very successful **MICROPHONE SPLITTER TRANSFORMER** type 4079 with a high impedance 200 ohm primary and two 200 ohm secondaries. It will handle up to 2.3 volts rms at 30 Hz and has a frequency response of plus/minus 1/2dB from 20 Hz to 20 kHz. It is contained in a Mumetal Can 33 mm diam x 37 mm high and WORKING DETAILS OF THE CONSTRUCTION OF A SPLITTER ARE AVAILABLE ON REQUEST.

We will supply single transformers, or any quantity, with short delivery times and, without obligation on your part, will quote price and exact dispatch on receipt of your requirements. **E. A. SOWTER LTD., Manufacturers and Designers, P.O. Box 36 IPSWICH IP1 2EG, ENGLAND. Tel. IPSWICH (0473) 52794 and 219390.** (8289)

### THE FABULOUS D2 MICROPROCESSOR EVALUATION KIT FROM MOTOROLA

Featuring "24 key keyboard" Seven segment display "Cassette" interface "Erom & Ram Expandable" Interface Capability. "Full Documentation" 5 Volt power supply required. "One year's FREE membership of The Amateur Computer Club with every purchase" £176 + £1.50 P&P + 8% VAT.

### ENAMELLED COPPER WIRE

SWG	1lb	8oz	4oz	2oz
14-19	2.80	1.40	.66	.55
20-29	2.60	1.60	.85	.65
30-34	3.00	1.70	.95	.70
35-40	3.35	1.90	1.10	.79
40-43	4.50	2.50	1.90	1.25
44-46	5.00	3.00	2.10	1.65
47	8.00	5.00	3.00	1.76
48	15.00	9.00	6.00	3.30

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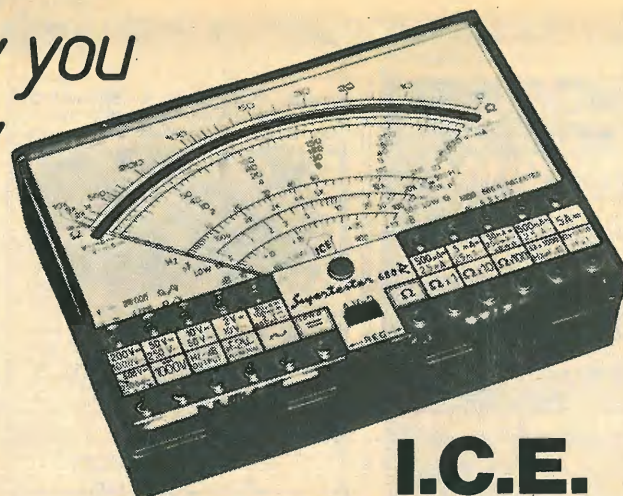
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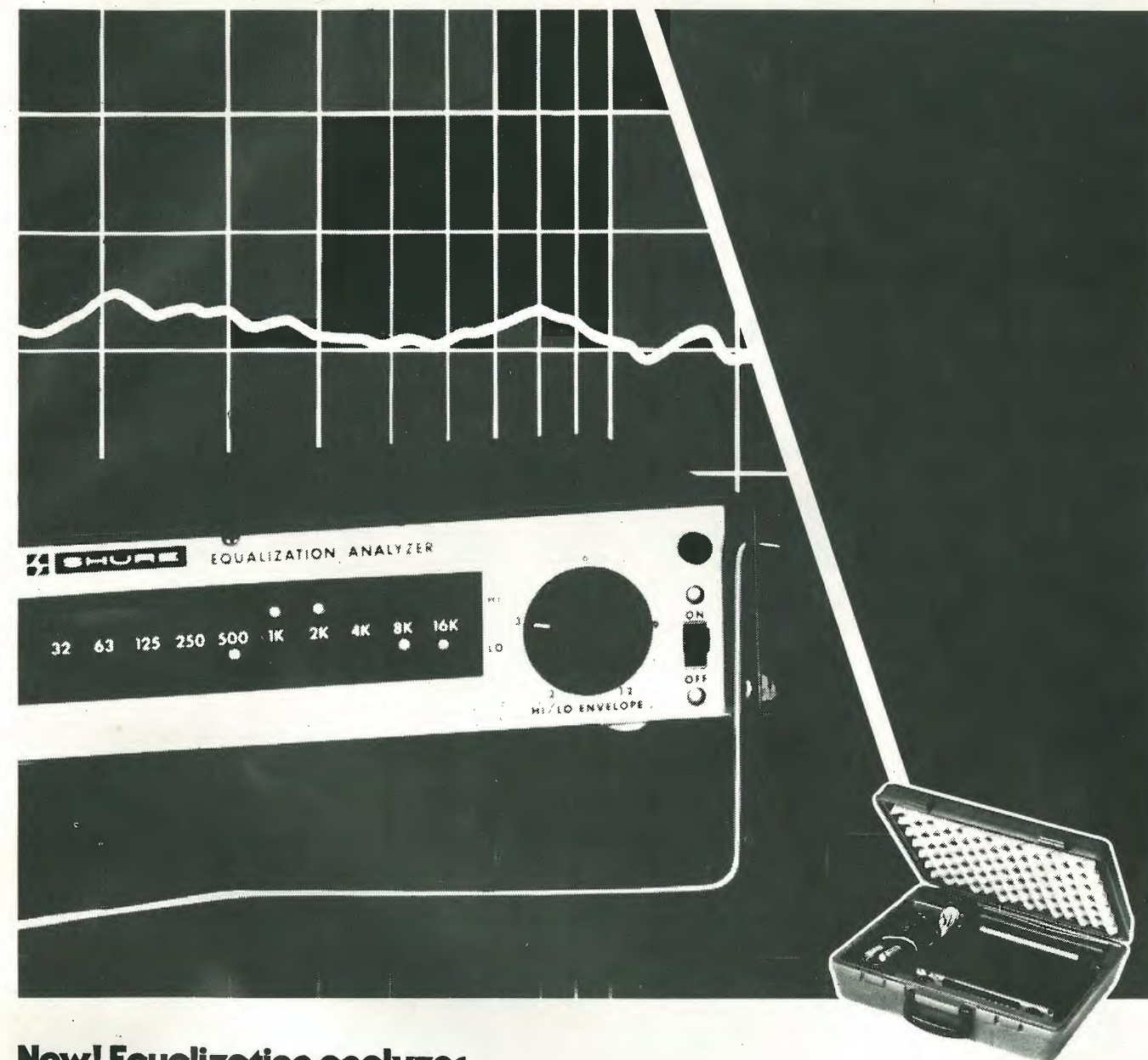
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